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# Ch 4. Combinational Logic Technologies

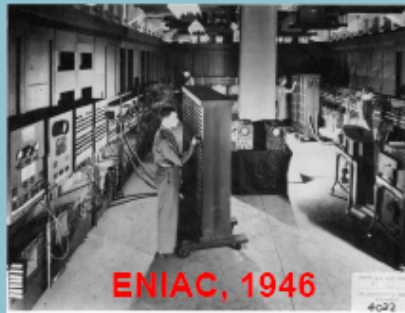
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# History

## - From Switches to Integrated Circuits

- The underlying implementation technologies for digital systems

### Vacuum Tubes



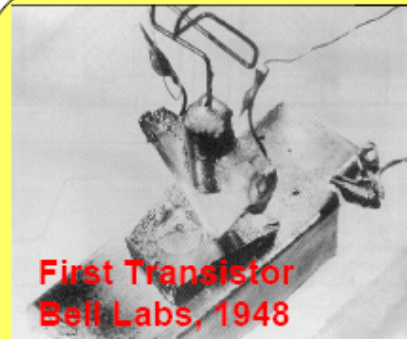
ENIAC, 1946



UNIVAC, 1951

1900 adds/sec

### Transistors



First Transistor  
Bell Labs, 1948



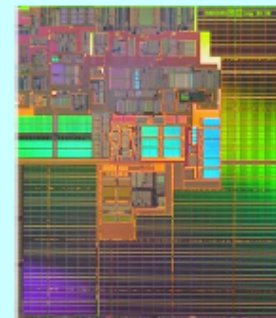
IBM System/360, 1964

500,000 adds/sec

### VLSI Circuits



4004, 1971



Intel Itanium, 2003

2,000,000,000  
adds/sec

# Packaged Logic, Configurability, and Programmable Logic

- Standard parts
  - Integrated circuits that contain small number of simple logic gates widely used
- ROM and PLA/PAL
  - ROM
    - A fixed array of ones and zeros
  - Programmable logic array (PLA), Programmable array logic (PAL)
    - An array of fixed logic gates, arranged in a standard two-level form such as AND/OR
- Application-Specific Integrated Circuit (ASIC)
  - Gate array / Standard cells
- Field-Programmable Gate Array (FPGA)

# Packaged Logic, Configurability, and Programmable Logic (cont'd)

- Historical development of components

<b>TABLE 4.2</b>		<i>Historical Development of Components (Adapted from Oldfield and Dorf, <i>Field Programmable Gate Arrays</i>, John Wiley, New York, 1995)</i>			
	<i>1960s SSI/MSI</i>	<i>1970s LSI</i>	<i>1980s VLSI</i>	<i>1990s Programmable Logic</i>	
<i>Components</i>	Logic, Resistor/ Transistor elements	8-bit $\mu$ processor, Memory, ROM	32-bit $\mu$ processor, Gate arrays	64-bit $\mu$ processor, PALs, FPGAs	
<i>Complexity Level (# of gates)</i>	100s	10,000s	1 million	100,000s to millions	
<i>Pervasive Components</i>	TTL 7400 series	Intel 8008, ROM	Intel 8086, Motorola 68000, Gate arrays, PALs	Pentium I, II, III, FPGAs, Complex PLDs	
<i>Dominant Trend</i>	Standard catalog of components	Larger, General-purpose components, e.g., Micro- processors and Random- access memories	Application-spe- cific integrated circuits	Field- programmable components	

# Technology Metrics

- Gate delay
  - The time delay from an input change to an output change
  - $t(50\% \text{ output final value}) - t(50\% \text{ of input final value})$
- Degree of integration
  - The chip area and number of chip packages required to implement a given function in a technology
- Power dissipation
  - Power consumed as gates perform their logic functions
- Noise margin
  - The maximum voltage that can be added to or subtracted from the logic voltages carried on wires and still have the circuit interpret the voltage as the correct logic value
- Component cost

# Technology Metrics (cont'd)

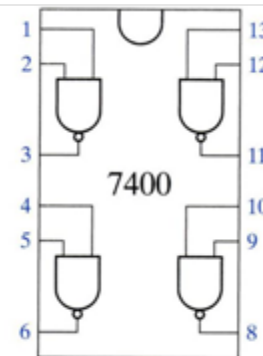
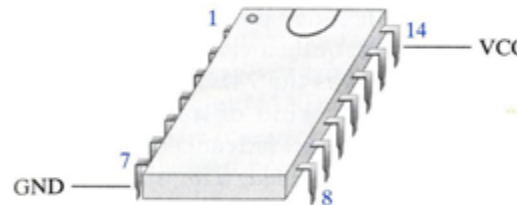
- Fan-out
  - The number of gate inputs to which a given gate output can be connected without exceeding electrical limitations
- Driving capability
  - The speed of communications between packaged components
- Configurability

# Basic Logic Components

## - Fixed Logic

### ■ Cell-based design

- Designers pick and choose gates in a given standard cell library for logic functions they want to implement.
  - NAND, NOR, XOR, ADDER, MUX, ...
- SSI and MSI (Small and Medium Scale Integrated Circuit)
  - TTL: a family of packaged logic components
  - Example: TTL 7400

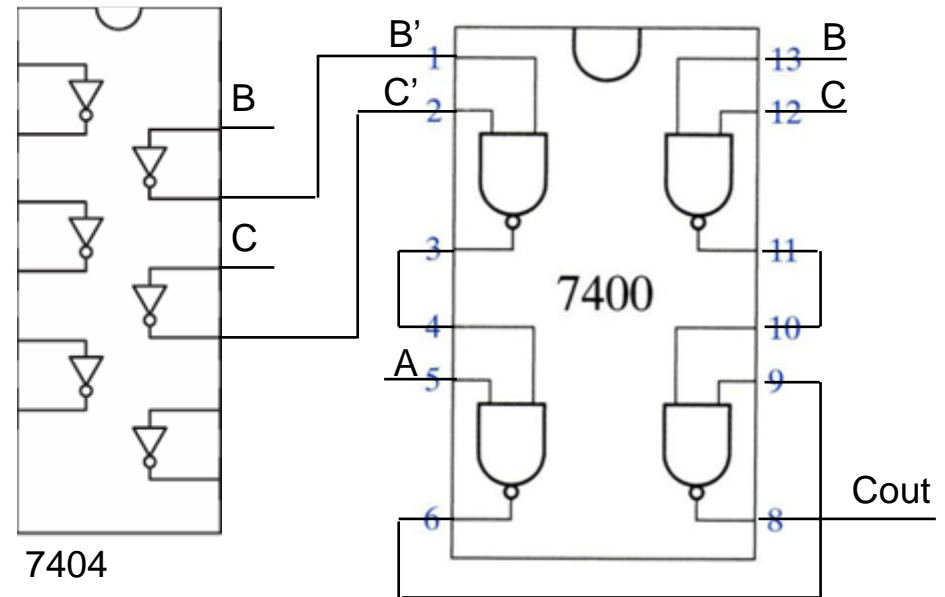
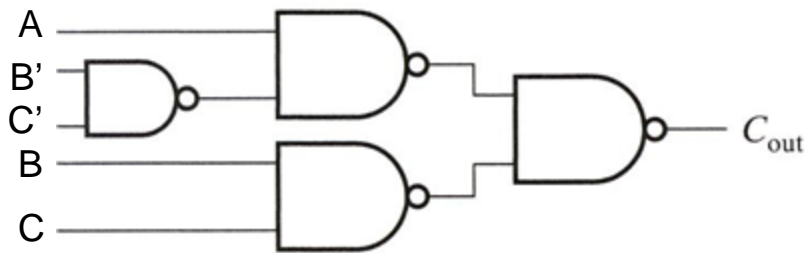


### □ Standard cell library for modern ASIC

- provides the same kinds of logic functions as those TTL standard catalog contains.

# Fixed Logic (cont'd)

- Combinational logic implementation
  - Pick logic packages and wire them
  - Make Package/Cell Counts minimal





# Basic Logic Components (cont'd)

## - Look-Up Tables

- Look-up table approach
  - Storing the output value of a function for each input combination in a table
  - Using the current input value as an index to look-up what the output should be
  - To change the function, simply change the values stored in the table, not change any wiring
  - Less sensitive to the complexity of the function

### ■ Example

- If the input  $A=0$ ,  $B=1$ , and  $C=0$ , the corresponding index (address) of the table = 010 and the value of the entry = 0100.
- Thus, the output  $F0=0$ ,  $F1=1$ ,  $F2=0$ ,  $F3=0$

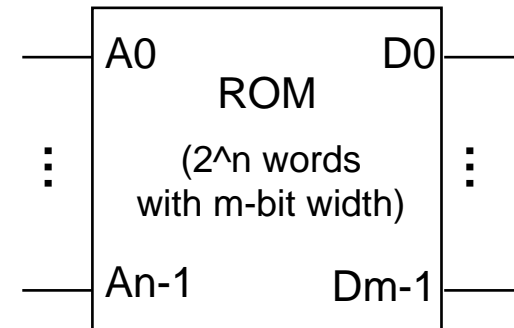
Index			Output			
A	B	C	F0	F1	F2	F3
0	0	0	0	0	1	0
0	0	1	1	1	1	0
0	1	0	0	1	0	0
0	1	1	0	0	0	1
1	0	0	1	0	1	1
1	0	1	1	0	0	0
1	1	0	0	0	0	1
1	1	1	0	1	0	0

**Storing the outputs  
at each entry  
in a table device**

# Look-Up Tables (cont'd)

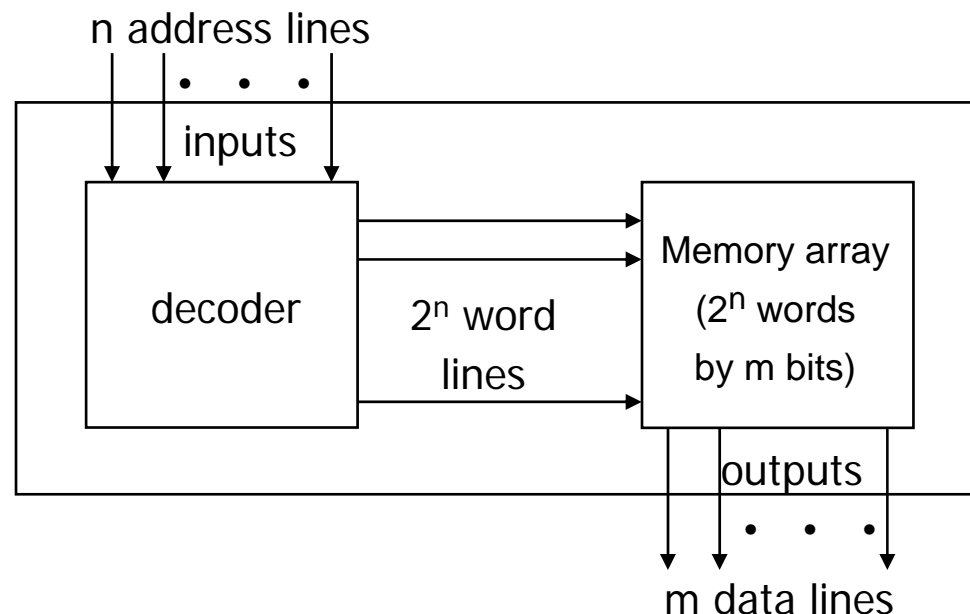
## ■ ROM (Read-Only Memory)

- an array of values intended to be read many times but only written once
- We can program the output of a truth table directly into a ROM
- Address: index of the array
  - Inputs are used to index a row of the array
- The value of the corresponding row are read out as the values of the outputs
- Each additional input bit doubles the size of the memory array
  - Good to implement a complicated function that is difficult to simplify using the standard methods or change often the function.
- Variants
  - Programmable ROM (PROM)
  - Erasable PROM (EPROM)
  - Electrically erasable PROM (EEPROM)



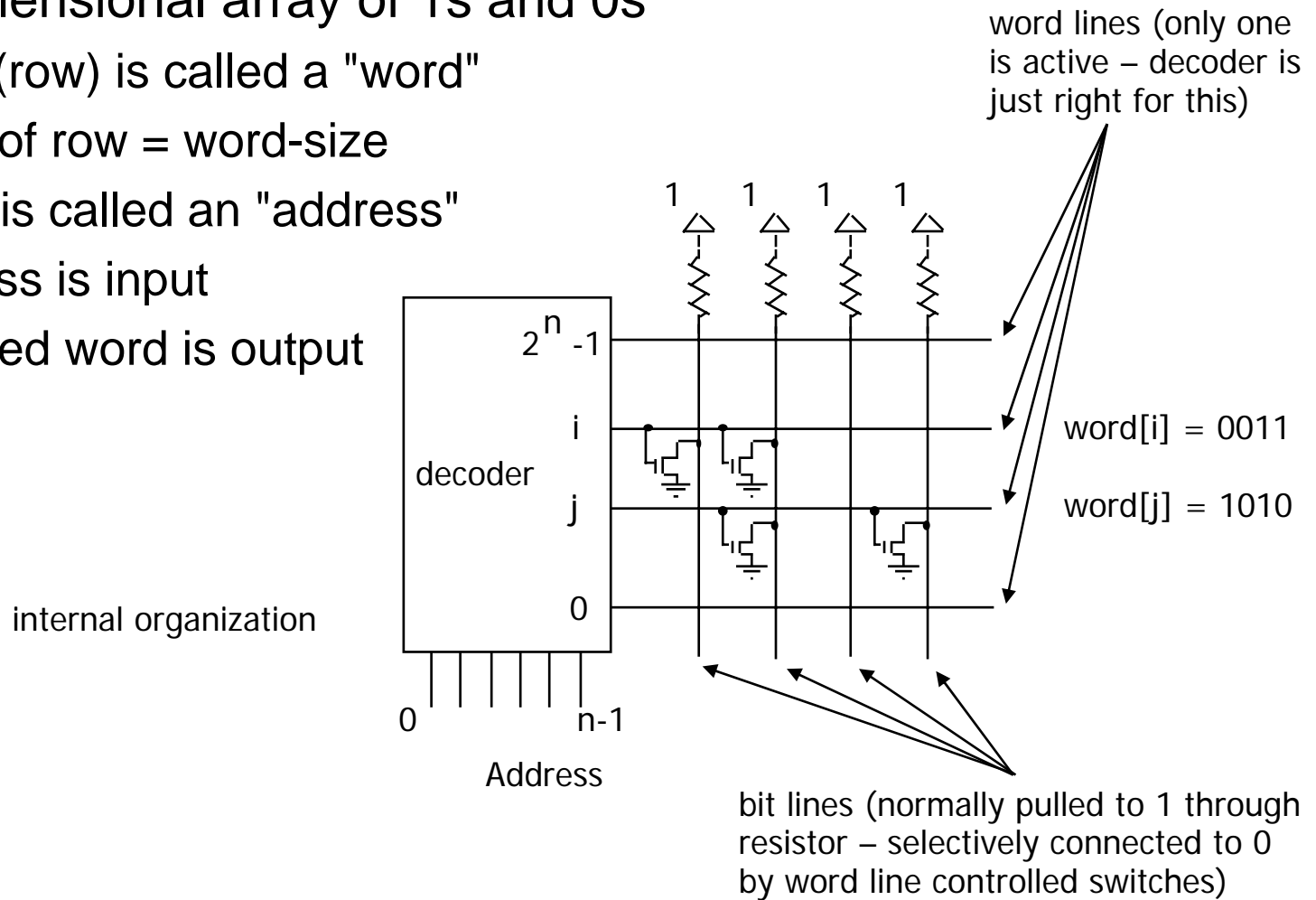
# Look-Up Tables (cont'd)

- Internal organization of a ROM
  - Each row of the array is called a word and is selected by the control inputs, which are called the address.
  - The number of columns in the array is called the bit-width or word size.
  - Similar to a PLA structure but with a fully decoded AND array
    - completely flexible OR array (unlike PAL)



# Look-Up Tables (cont'd)

- Two dimensional array of 1s and 0s
  - entry (row) is called a "word"
  - width of row = word-size
  - index is called an "address"
  - address is input
  - selected word is output



# Look-Up Tables (cont'd)

- Combinational logic implementation (two-level canonical form) using a ROM

$$F0 = A' B' C + A B' C' + A B' C$$

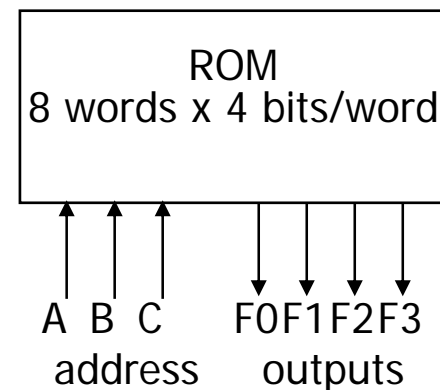
$$F1 = A' B' C + A' B C' + A B C$$

$$F2 = A' B' C' + A' B' C + A B' C'$$

$$F3 = A' B C + A B' C' + A B C'$$

A	B	C	F0	F1	F2	F3
0	0	0	0	0	1	0
0	0	1	1	1	1	0
0	1	0	0	1	0	0
0	1	1	0	0	0	1
1	0	0	1	0	1	1
1	0	1	1	0	0	0
1	1	0	0	0	0	1
1	1	1	0	1	0	0

truth table



block diagram

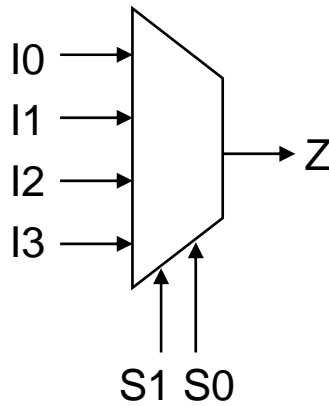
# Basic Logic Components (cont'd)

## - Multiplexer/Selector

### ■ Multiplexer/Selector (MUX)

- Sets its single output to the same value as one of its inputs under the direction of its control inputs.
- $2^n$  data inputs,  $n$  control inputs (called "selects"), 1 output
- used to connect  $2^n$  input points to a single output point
- control signal pattern forms binary index of input connected to output

### ■ Functional description



*Input :*  $I_j = \{0,1\}$  where  $j = 0, 1, \dots, 2^n - 1$

$S_i = \{0,1\}$  where  $i = 0, 1, \dots, n-1$

*Output :*  $Z = \{0,1\}$

*Function :*  $Z = I_j$  where  $j = S = \sum_{i=0}^{n-1} S_i 2^i$

# Multiplexer/Selector (cont'd)

- Truth table of MUX
  - 2:1 MUX

$$Z = A' I_0 + A I_1$$

A	Z
0	$I_0$
1	$I_1$

functional form

logical form

two alternative forms  
for a 2:1 Mux truth table

$I_1$	$I_0$	A	Z
0	0	0	0
0	0	1	0
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	1
1	1	0	1
1	1	1	1

# Multiplexer/Selector (cont'd)

## ■ Boolean equation of MUX

□ 2:1 mux:  $Z = A'I_0 + AI_1$

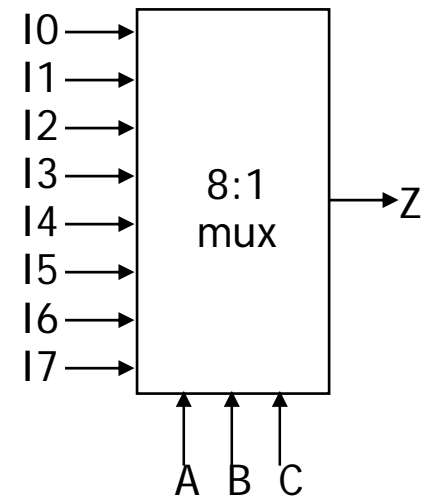
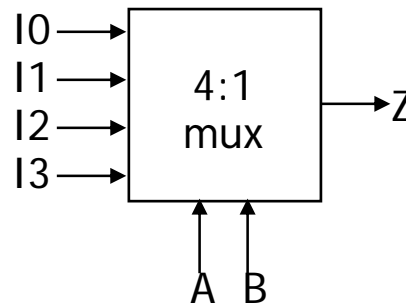
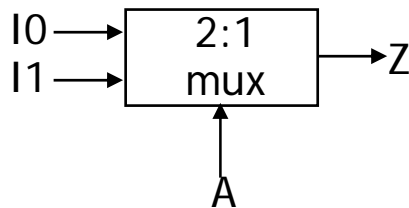
□ 4:1 mux:  $Z = A'B'I_0 + A'BI_1 + AB'I_2 + ABI_3$

□ 8:1 mux:  $Z = A'B'C'I_0 + A'B'CI_1 + A'BC'I_2 + A'BCI_3 + AB'C'I_4 + AB'CI_5 + ABC'I_6 + ABCI_7$

□ In general:

$$Z = \sum_{k=0}^{2^n-1} m_k I_k$$

■ in minterm shorthand form for a  $2^n:1$  Mux

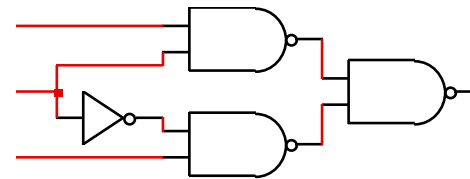
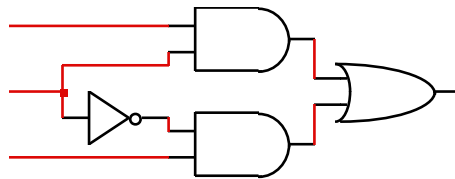




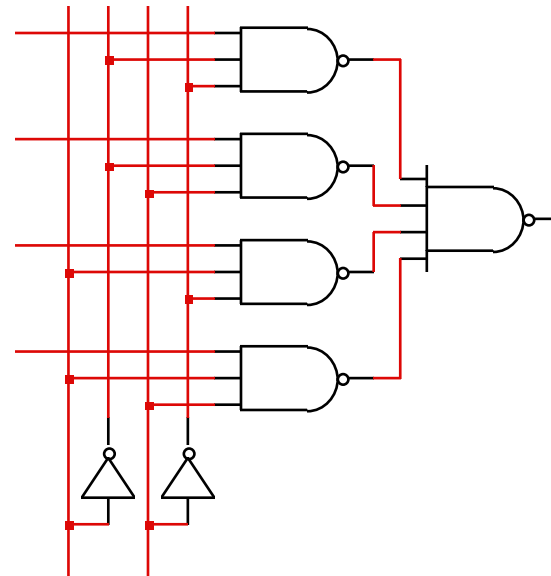
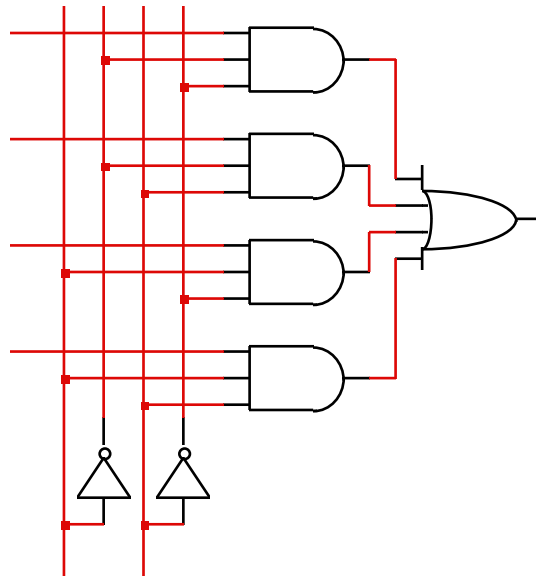
# Multiplexer/Selector (cont'd)

## ■ Gate Level Implementation of MUX

### □ 2:1 mux



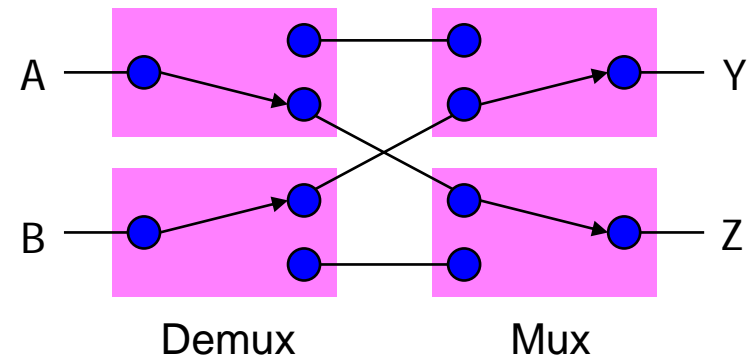
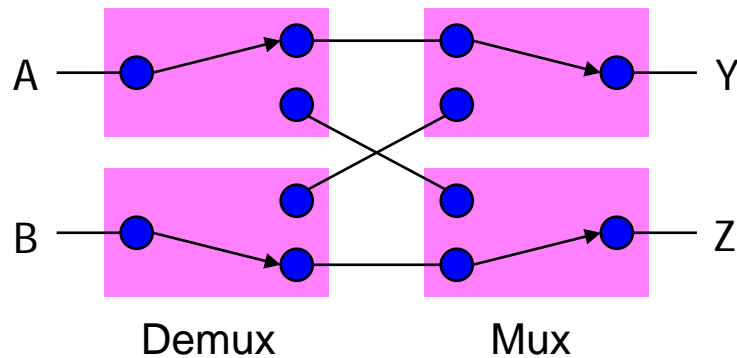
### □ 4:1 mux



# Multiplexer/Selector (cont'd)

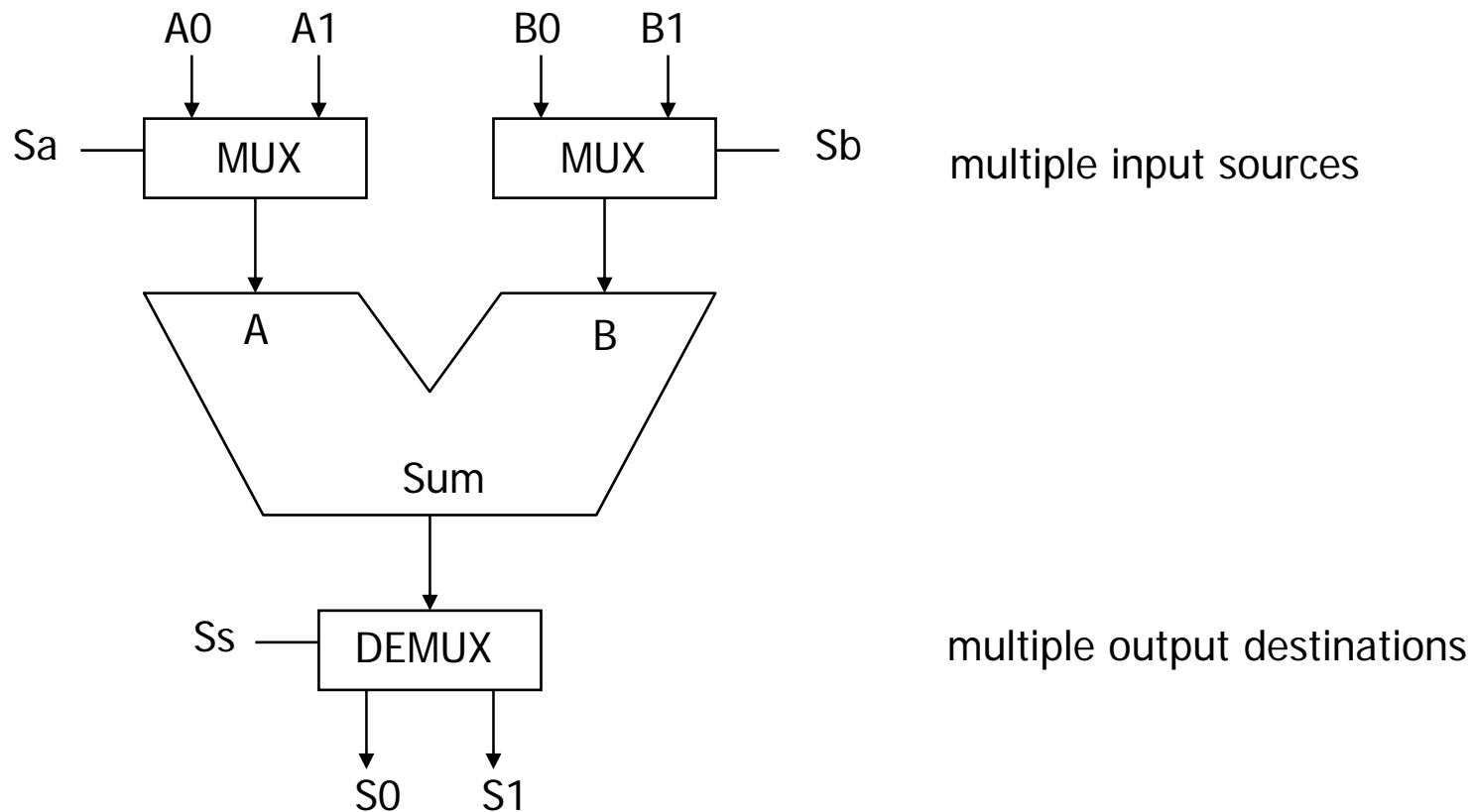
## - Mux and Demux

- Switch implementation of multiplexers and demultiplexers
  - can be composed to make arbitrary size switching networks
  - used to implement multiple-source/multiple-destination interconnections



# Mux and Demux (cont'd)

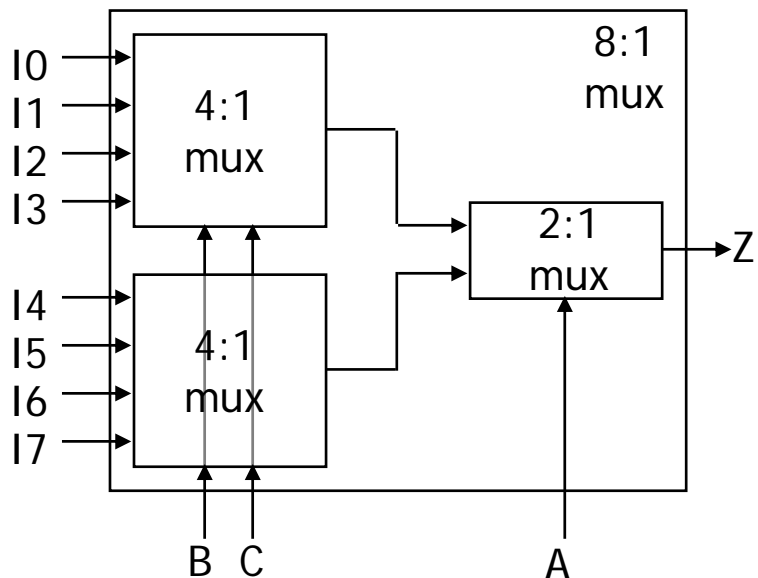
- Uses of multiplexers/demultiplexers in multi-point connections



# Multiplexer/Selector (cont'd)

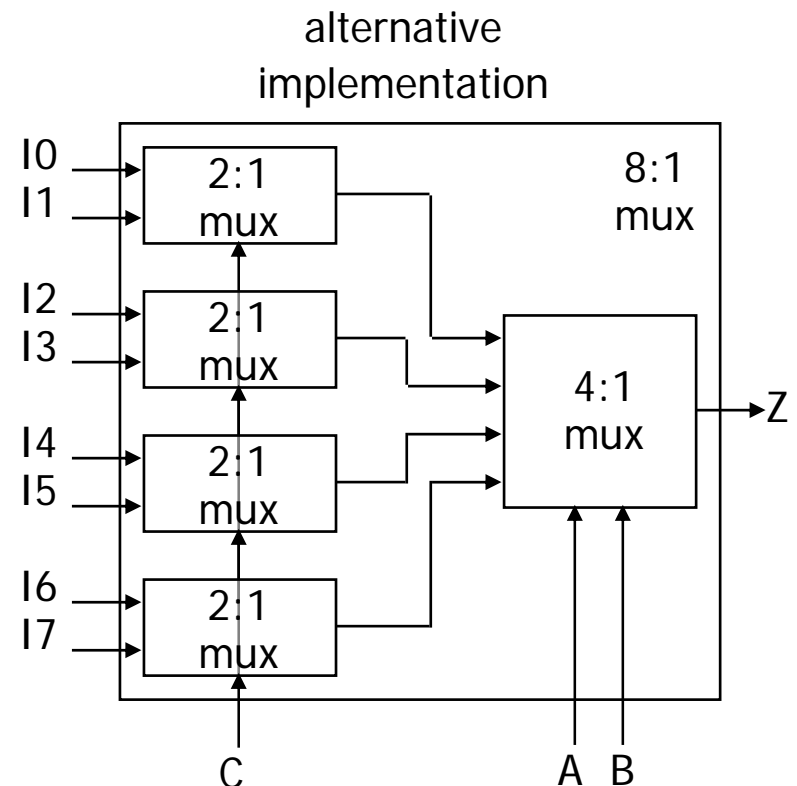
## - Cascading Multiplexers

- Large multiplexers can be made by cascading smaller ones



control signals B and C simultaneously choose one of I0, I1, I2, I3 and one of I4, I5, I6, I7

control signal A chooses which of the upper or lower mux's output to gate to Z



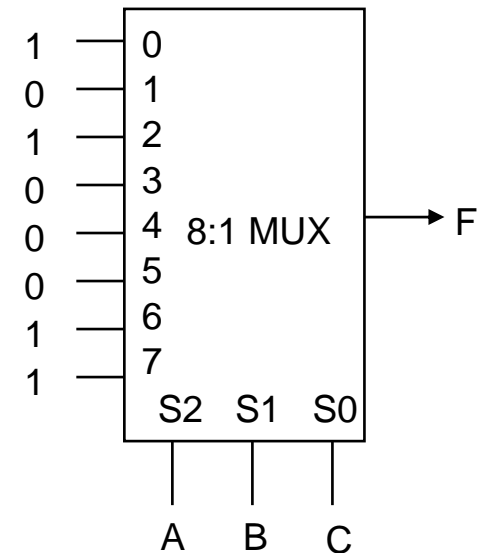
# Multiplexer/Selector (cont'd)

## - Multiplexers as a Logic Building Block

- A  $2^n:1$  multiplexer can implement any function of  $n$  variables
  - with the variables used as control inputs and
  - the data inputs tied to 0 or 1
  - in essence, a lookup table

- Example:

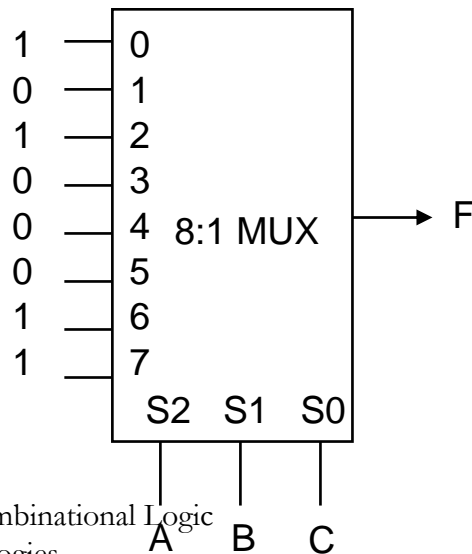
- $$\begin{aligned} F(A,B,C) &= m_0 + m_2 + m_6 + m_7 \\ &= A'B'C' + A'BC' + ABC' + ABC \\ &= A'B'C'(1) + A'B'C(0) \\ &\quad + A'BC'(1) + A'BC(0) \\ &\quad + AB'C'(0) + AB'C(0) \\ &\quad + ABC'(1) + ABC(1) \end{aligned}$$



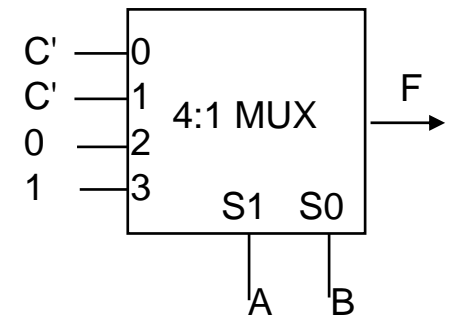
$$F = A'B'C'I_0 + A'B'CI_1 + A'BC'I_2 + A'BCI_3 + AB'C'I_4 + AB'CI_5 + ABC'I_6 + ABCI_7$$

# Multiplexers as a Logic Building Block (cont'd)

- A  $2^{n-1}:1$  multiplexer can implement any function of  $n$  variables
  - with  $n-1$  variables used as control inputs and
  - the data inputs tied to the last variable or its complement
- Example:
  - $F(A,B,C) = m_0 + m_2 + m_6 + m_7$   
 $= A'B'C' + A'BC' + ABC' + ABC$   
 $= A'B'(C') + A'B(C') + AB'(0) + AB(1)$

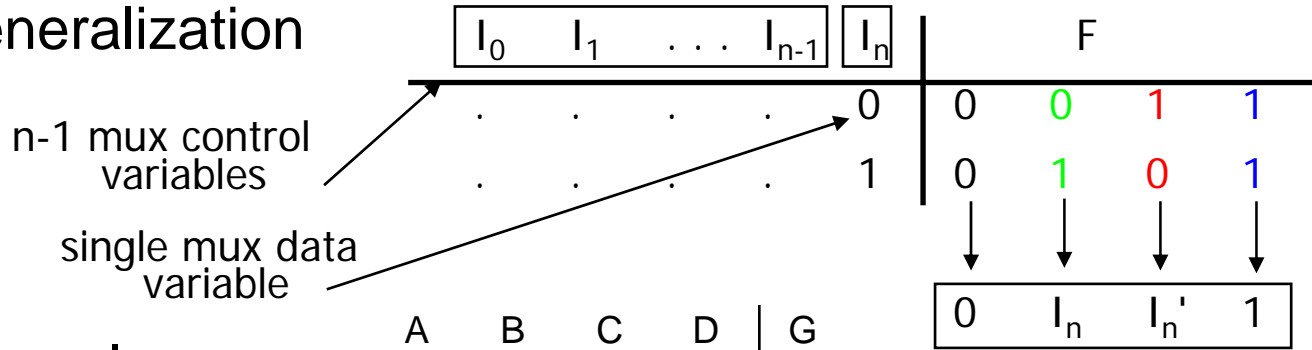


A	B	C	F
0	0	0	1
0	0	1	0
0	1	0	1
0	1	1	0
1	0	0	0
1	0	1	0
1	1	0	1
1	1	1	1



# Multiplexers as a Logic Building Block (cont'd)

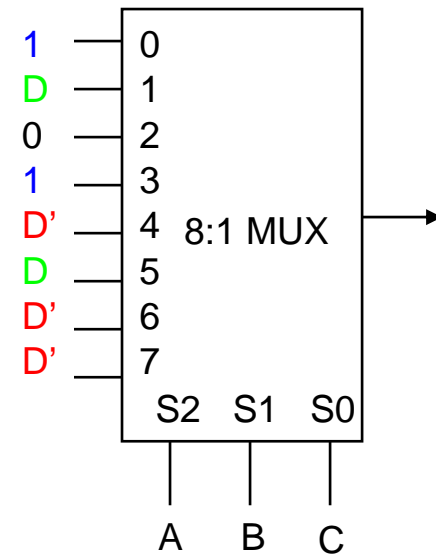
## Generalization



## Example: $G(A,B,C,D)$ can be realized by an 8:1 MUX

A	B	C	D	G
0	0	0	0	1
0	0	0	1	1
0	0	1	0	0
0	0	1	1	1
0	1	0	0	0
0	1	0	1	0
0	1	1	0	1
0	1	1	1	1
1	0	0	0	1
1	0	0	1	0
1	0	1	0	0
1	0	1	1	1
1	1	0	0	1
1	1	0	1	0
1	1	1	0	1
1	1	1	1	0

choose A,B,C as control variables



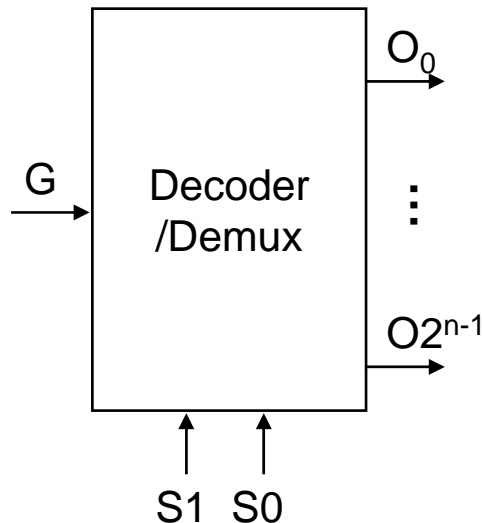
# Basic Logic Components (cont'd)

## - Demultiplexer/Decoder

### ■ Decoders/demultiplexers

- single data input,  $n$  control inputs,  $2^n$  outputs
- control inputs (called “selects” (S)) represent binary index of output to which the input is connected
- data input usually called “enable” (G)

### ■ Functional description



*Input:*  $G = \{0,1\}$

$S_i = \{0,1\}$  where  $i = 0, 1, \dots, n-1$

*Output:*  $O_j = \{0,1\}$  where  $j = 0, 1, \dots, 2^n - 1$

*Function:*  $O_j = \begin{cases} 1 & \text{where } j = S = \sum_{i=0}^{n-1} S_i 2^i \\ 0 & \text{otherwise} \end{cases}$



# Demultiplexer/Decoder (cont'd)

- Boolean equation of Decoder/Demux

- General form for n-selects:

- $m_j$  refers to minterm

$$O_j = G \cdot m_j = \begin{cases} G & \text{where } j = S = \sum_{i=0}^{i=n-1} S_i 2^i \\ 0 & \text{otherwise} \end{cases}$$

- $n = 1$

1:2 Decoder:

$$O0 = G \cdot S'$$

$$O1 = G \cdot S$$

- $n = 3$

3:8 Decoder:

$$O0 = G \cdot S2' \cdot S1' \cdot S0'$$

$$O1 = G \cdot S2' \cdot S1' \cdot S0$$

$$O2 = G \cdot S2' \cdot S1 \cdot S0'$$

$$O3 = G \cdot S2' \cdot S1 \cdot S0$$

$$O4 = G \cdot S2 \cdot S1' \cdot S0'$$

$$O5 = G \cdot S2 \cdot S1' \cdot S0$$

$$O6 = G \cdot S2 \cdot S1 \cdot S0'$$

$$O7 = G \cdot S2 \cdot S1 \cdot S0$$

- $n = 2$

2:4 Decoder:

$$O0 = G \cdot S1' \cdot S0'$$

$$O1 = G \cdot S1' \cdot S0$$

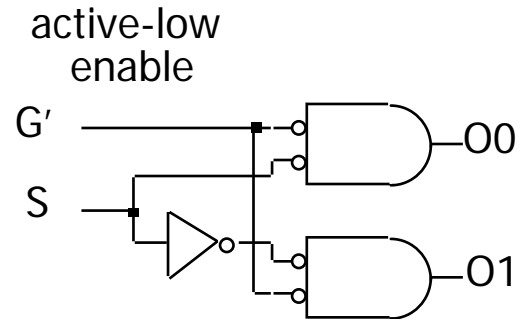
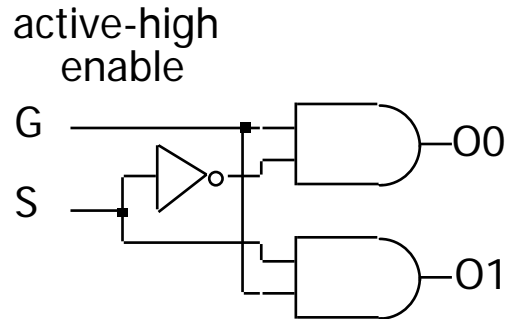
$$O2 = G \cdot S1 \cdot S0'$$

$$O3 = G \cdot S1 \cdot S0$$

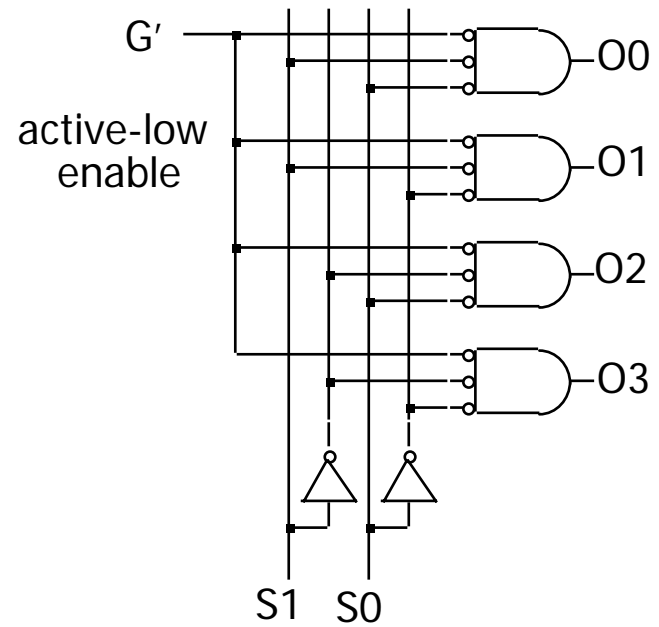
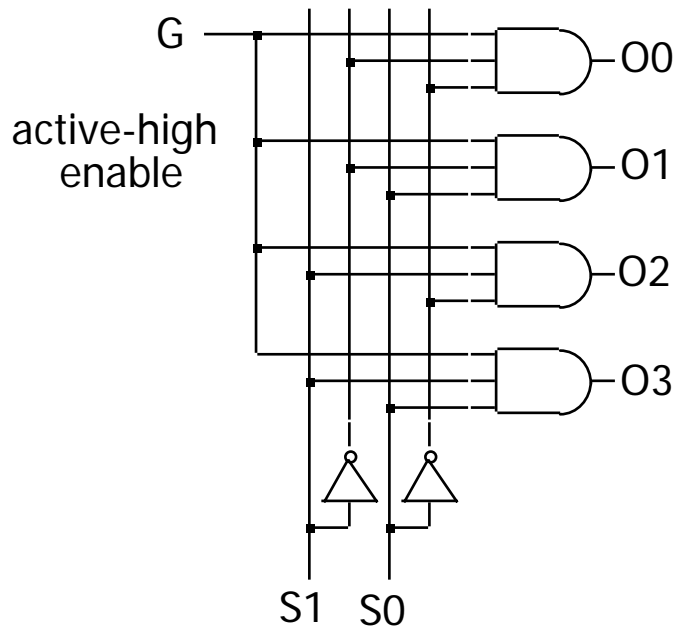
# Demultiplexer/Decoder (cont'd)

- Gate level implementation of Demux

- 1:2 decoders

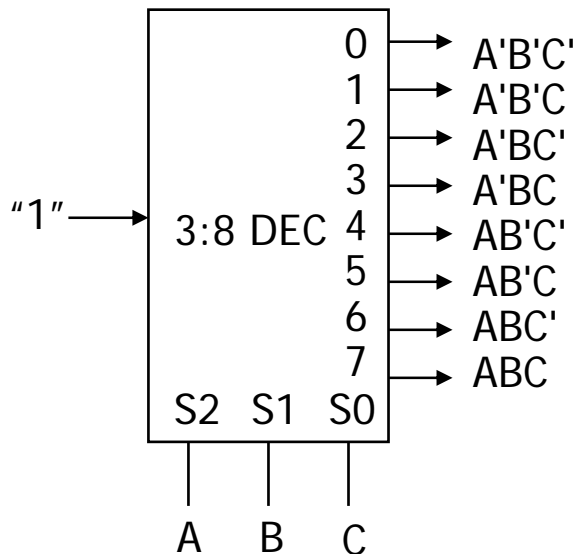


- 2:4 decoders



# Demultiplexer/Decoder (cont'd)

- Demux as a general-purpose building block
  - A  $n:2^n$  decoder can implement any function of  $n$  variables
    - with the variables used as control inputs
    - the enable inputs tied to 1 and
    - the appropriate minterms summed to form the function

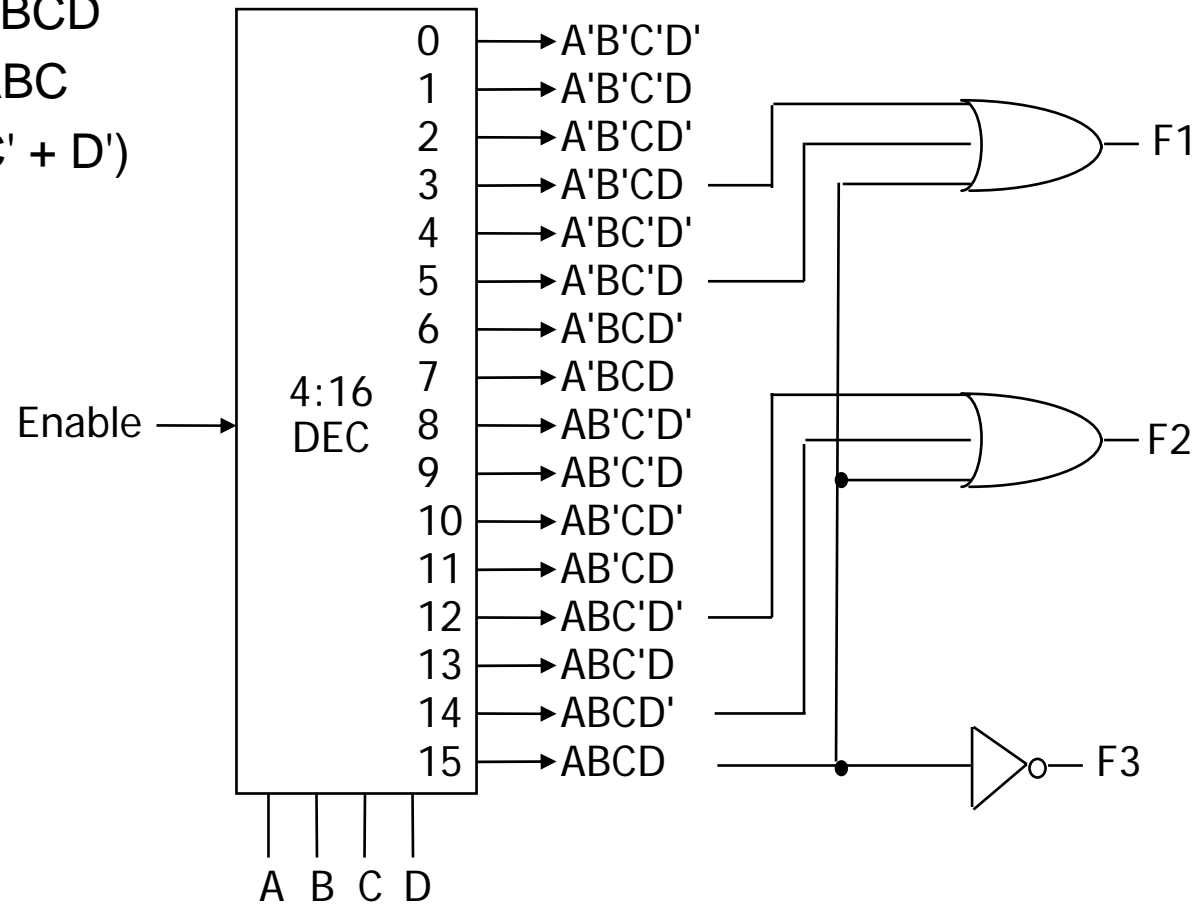


demultiplexer generates appropriate minterm based on control signals (it "decodes" control signals)

# Demultiplexer/Decoder (cont'd)

## ■ Demux as a general-purpose building block

- $F1 = A'BC'D + A'B'CD + ABCD$
- $F2 = ABC'D' + ABC$
- $F3 = (A' + B' + C' + D')$

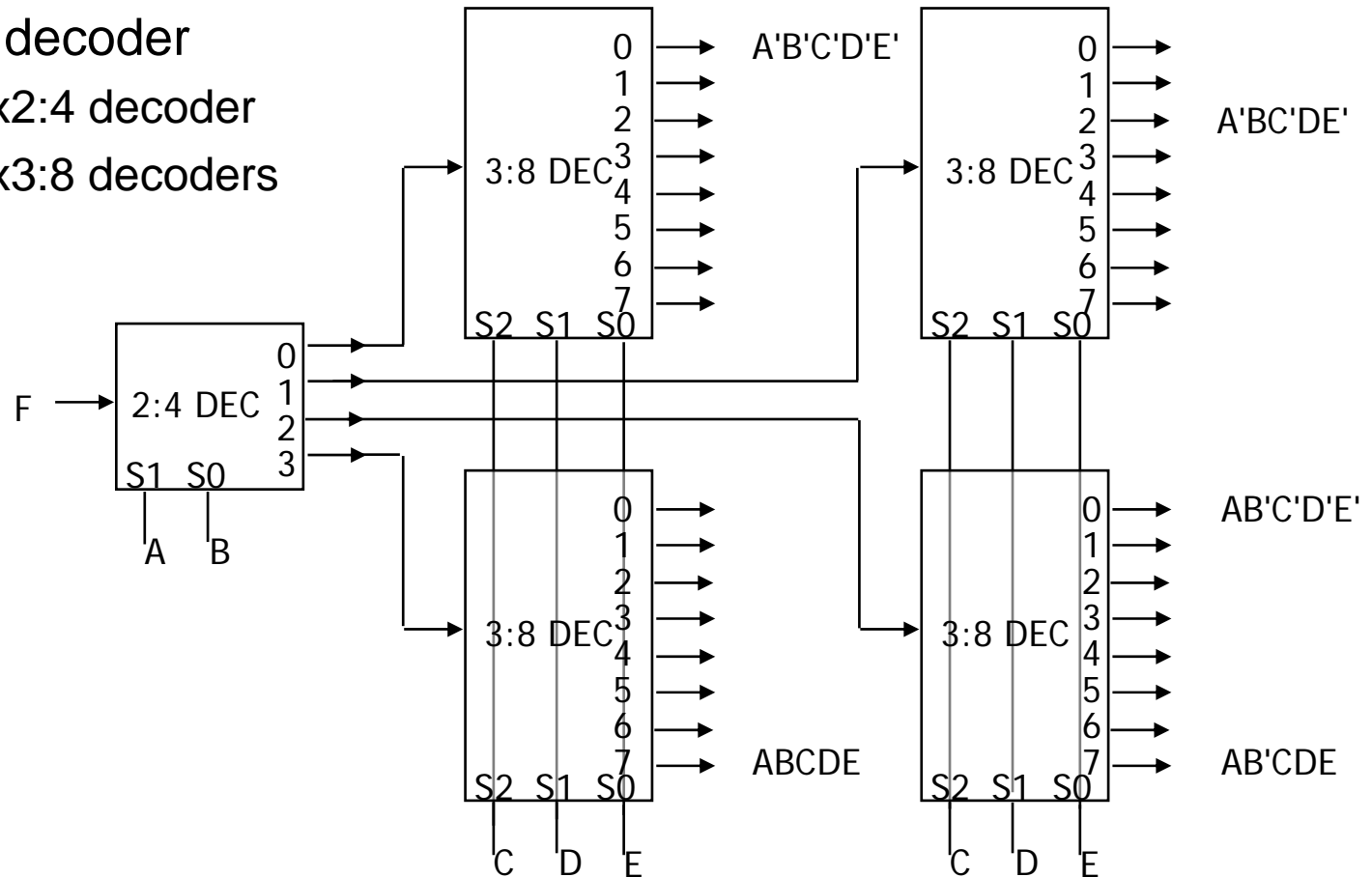


# Demultiplexer/Decoder (cont'd)

## ■ Cascading decoders

### □ 5:32 decoder

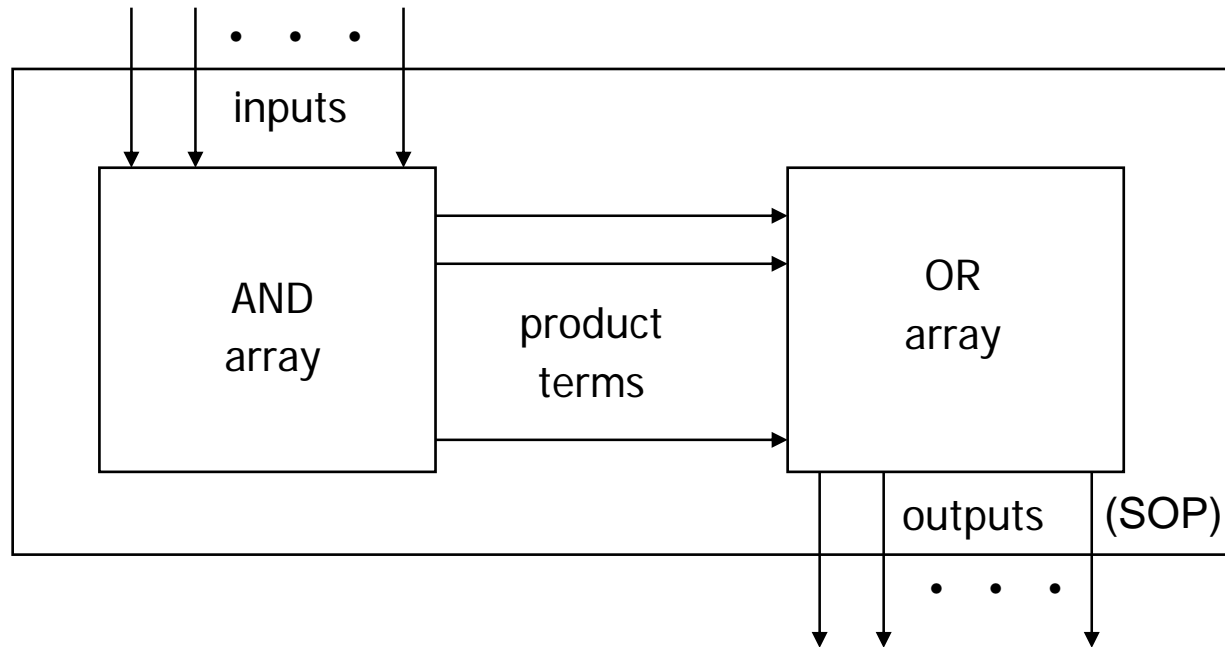
- 1x2:4 decoder
- 4x3:8 decoders



# Basic Logic Components (cont'd)

## - Programmable Logic Array

- Pre-fabricated building block of many AND/OR gates
  - actually NOR or NAND
  - "personalized" by making/breaking connections among the gates
  - programmable array block diagram for sum of products form



# Programmable Logic Array (cont'd)

## - Enabling Concept

- Shared product terms among outputs

example:

$$\begin{aligned}F_0 &= A + B' C' \\F_1 &= A C' + A B \\F_2 &= B' C' + A B \\F_3 &= B' C + A\end{aligned}$$

personality matrix

product term	inputs			outputs			
	A	B	C	F0	F1	F2	F3
AB	1	1	-	0	1	1	0
B'C	-	0	1	0	0	0	1
AC'	1	-	0	0	1	0	0
B'C'	-	0	0	1	0	1	0
A	1	-	-	1	0	0	1

input side:

1 = uncomplemented in term  
0 = complemented in term  
- = does not participate

output side:

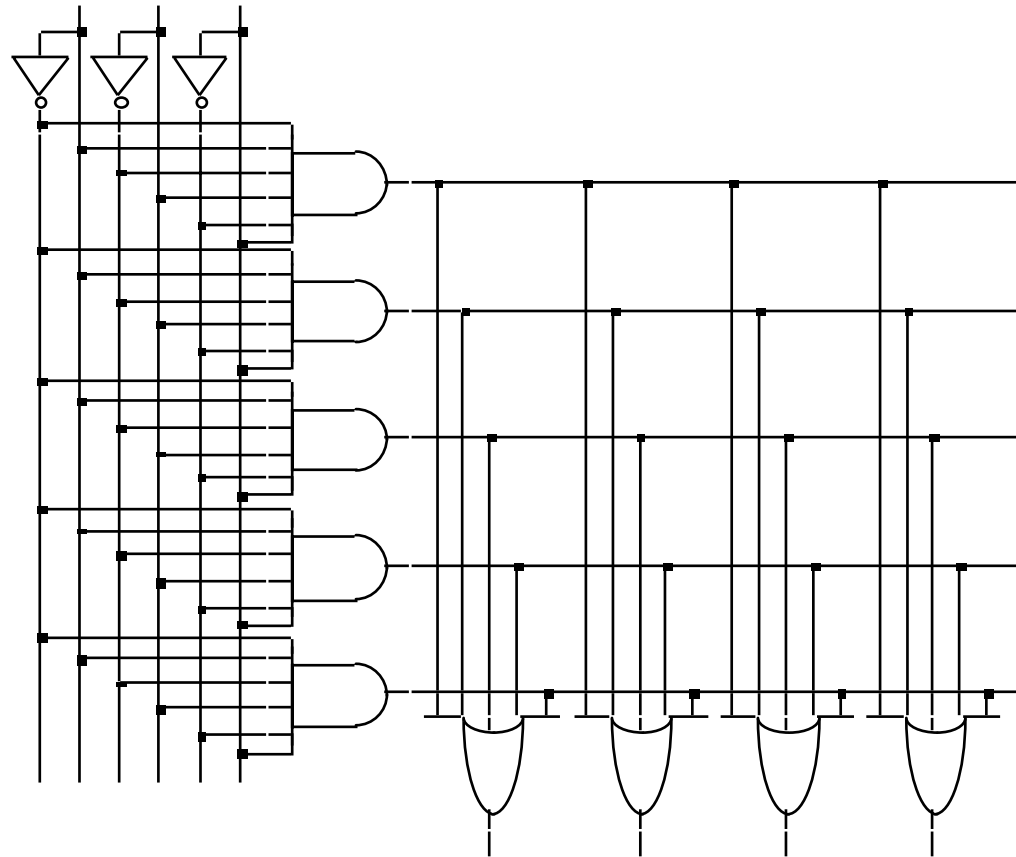
1 = term connected to output  
0 = no connection to output

reuse of terms

# Programmable Logic Array (cont'd)

## - Before Programming

- All possible connections are available before "programming"
  - in reality, all AND and OR gates are NANDs

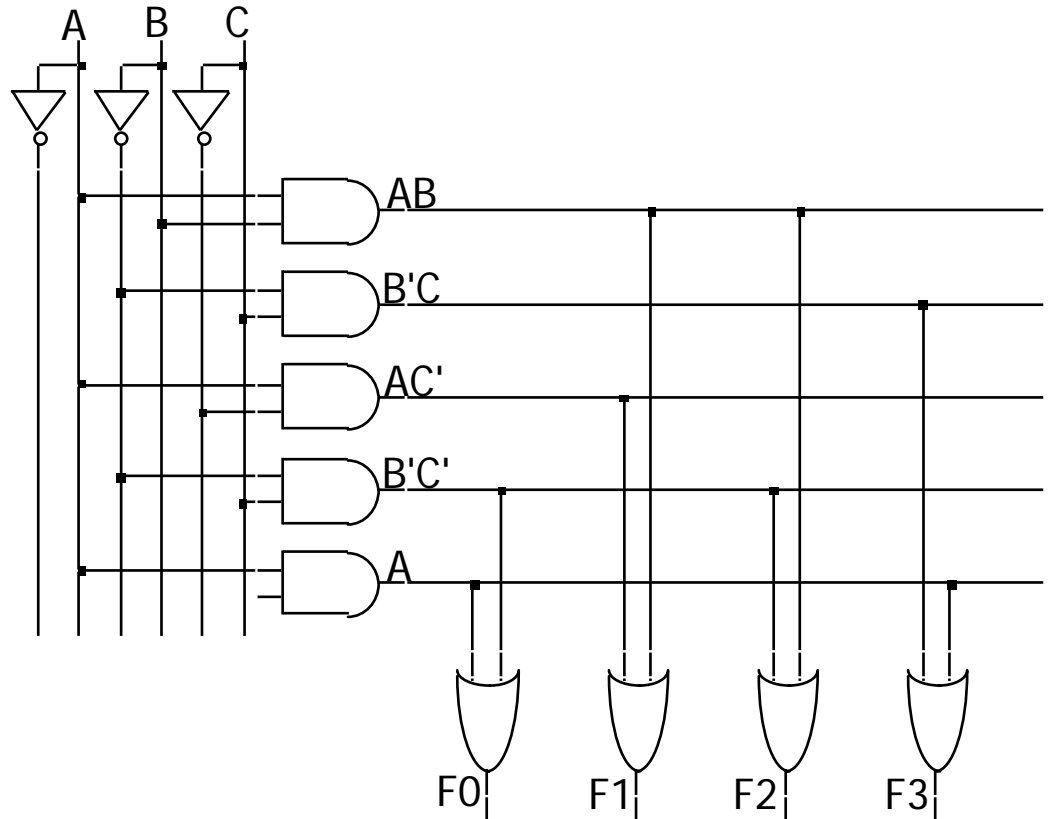




# Programmable Logic Array (cont'd)

## - After Programming

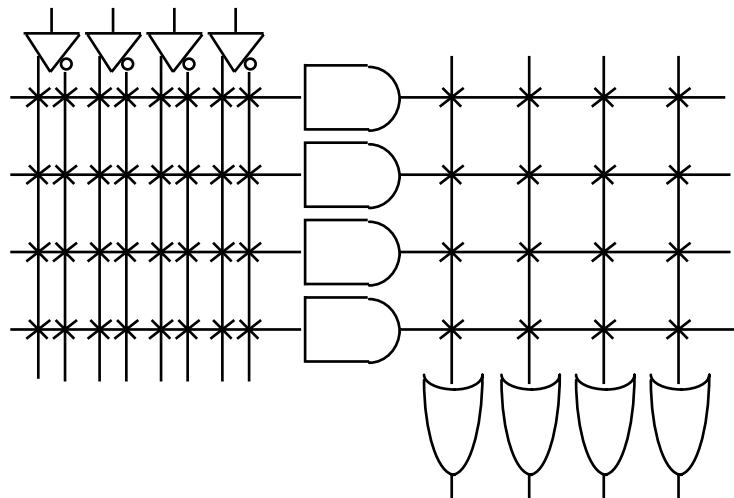
- Unwanted connections are "blown"
  - fuse (normally connected, break unwanted ones)
  - anti-fuse (normally disconnected, make wanted connections)
  - memory cell
    - 0 or 1 can be stored indicating whether there is to be a connection or not.



# Programmable Logic Array (cont'd)

## - Alternate Representation for High Fan-in Structures

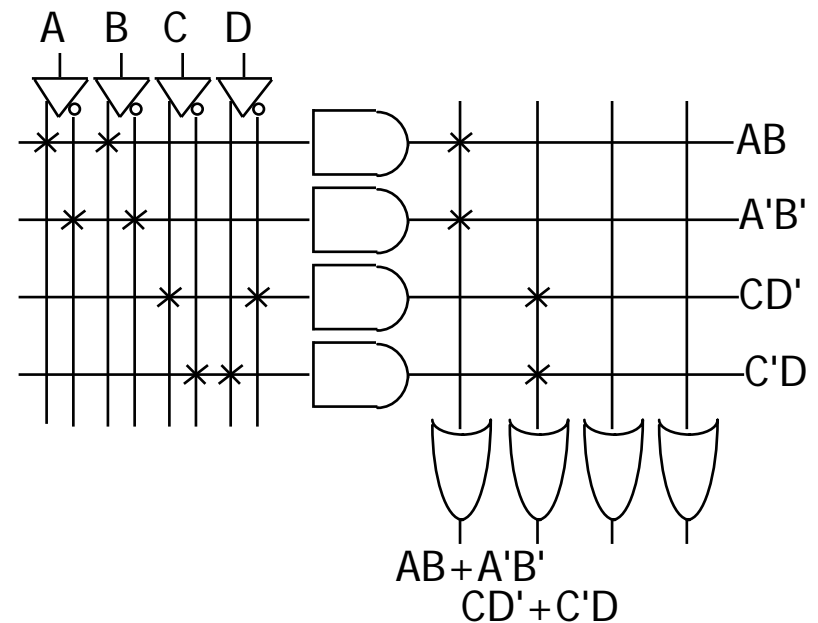
- Short-hand notation so we don't have to draw all the wires
  - × signifies a connection is present and perpendicular signal is an input to gate



notation for implementing

$$F0 = A B + A' B'$$

$$F1 = C D' + C' D$$



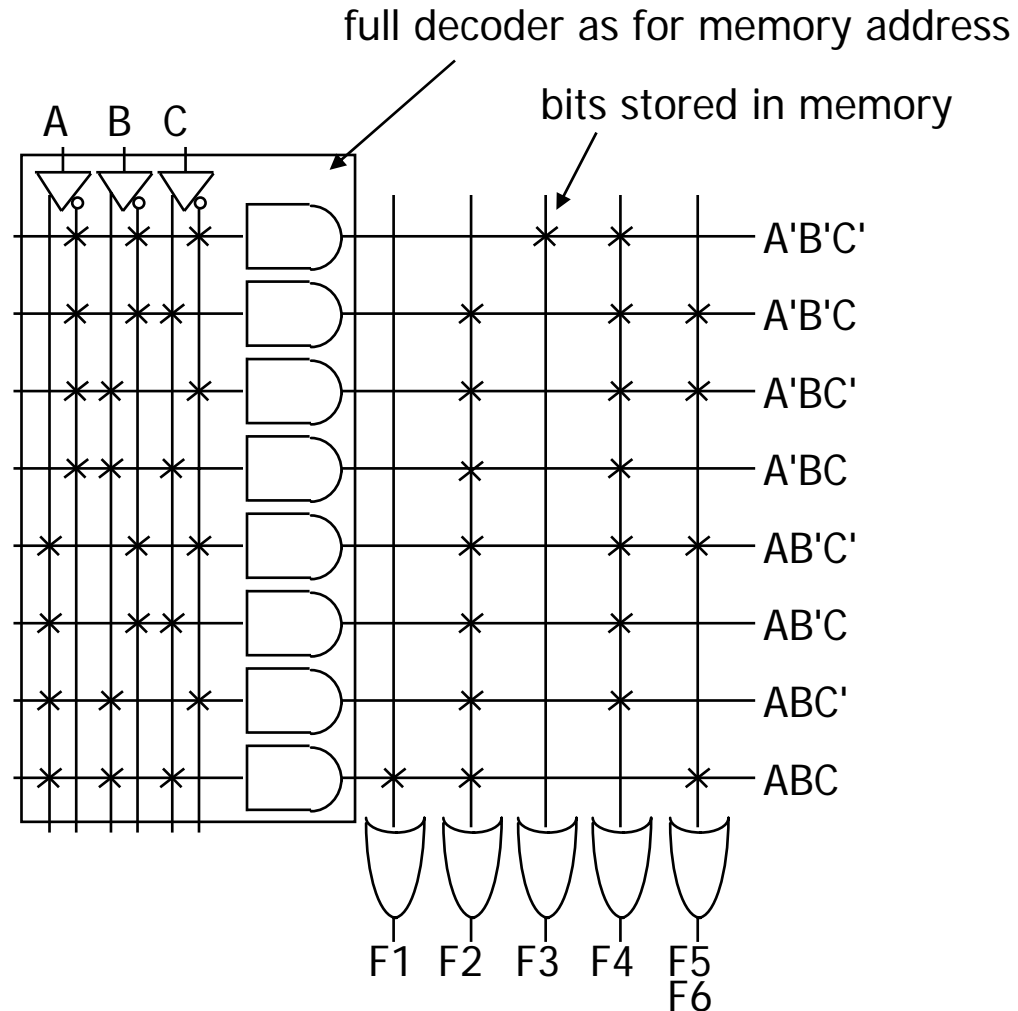
# Programmable Logic Array (cont'd)

## - Example

### ■ Multiple functions of A, B, C

- $F1 = A B C$
- $F2 = A + B + C$
- $F3 = A' B' C'$
- $F4 = A' + B' + C'$
- $F5 = A \text{ xor } B \text{ xor } C$
- $F6 = A \text{ xnor } B \text{ xnor } C$

A	B	C	F1	F2	F3	F4	F5	F6
0	0	0	0	0	1	1	0	0
0	0	1	0	1	0	1	1	1
0	1	0	0	1	0	1	1	1
0	1	1	0	1	0	1	0	0
1	0	0	0	1	0	1	1	1
1	0	1	0	1	0	1	0	0
1	1	0	0	1	0	1	0	0
1	1	1	1	1	0	0	1	1

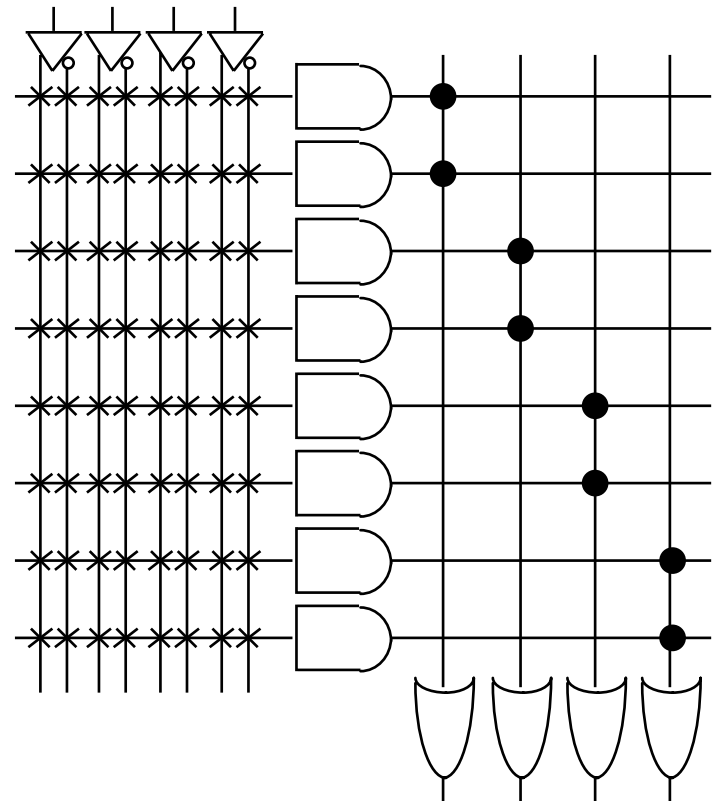


# Basic Logic Components (cont'd)

## - PALs and PLAs

- Programmable logic array (PLA)
  - what we've seen so far
  - unconstrained fully-general AND and OR arrays
- Programmable array logic (PAL)
  - constrained topology of the OR array
  - innovation by Monolithic Memories
  - faster and smaller OR plane

a given column of the OR array  
has access to only a subset of  
the possible product terms



# PALs and PLAs (cont'd)

## - PALs and PLAs: Design Example

### ■ BCD to Gray code converter

A	B	C	D	W	X	Y	Z
0	0	0	0	0	0	0	0
0	0	0	1	0	0	0	1
0	0	1	0	0	0	1	1
0	0	1	1	0	0	1	0
0	1	0	0	0	1	1	0
0	1	0	1	1	1	1	0
0	1	1	0	1	0	1	0
0	1	1	1	1	0	1	1
1	0	0	0	1	0	0	1
1	0	0	1	1	0	0	0
1	0	1	-	-	-	-	-
1	1	-	-	-	-	-	-

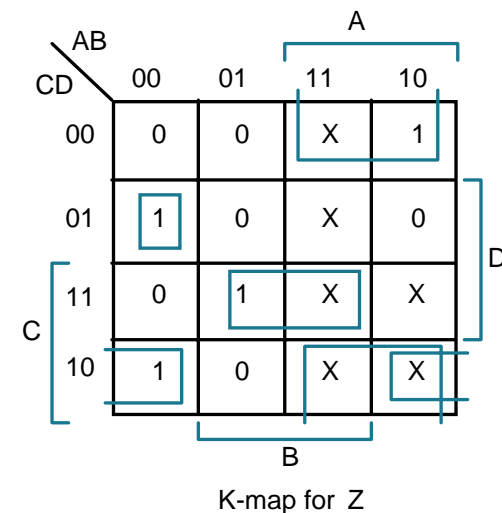
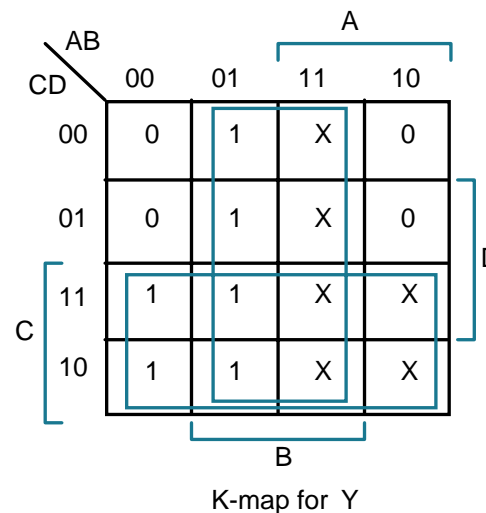
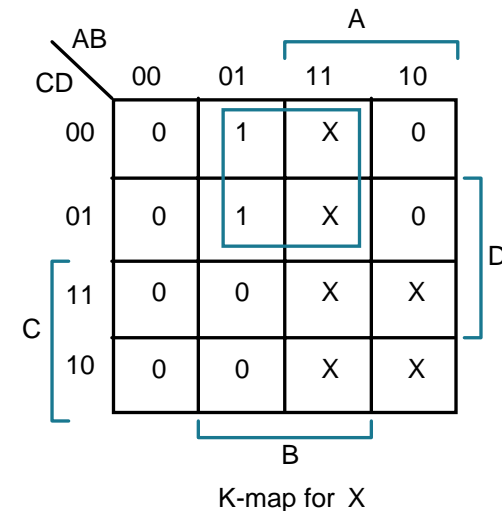
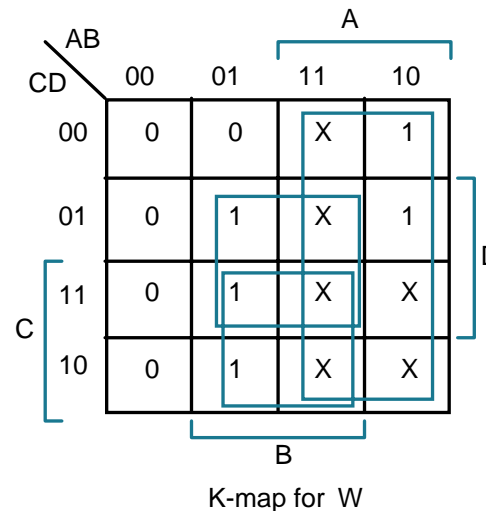
minimized functions:

$$W = A + BD + BC$$

$$X = BC'$$

$$Y = B + C$$

$$Z = A'B'C'D + BCD + AD' + B'CD'$$



# PALs and PLAs: Design Example (cont'd)

## Code converter: programmed PLA

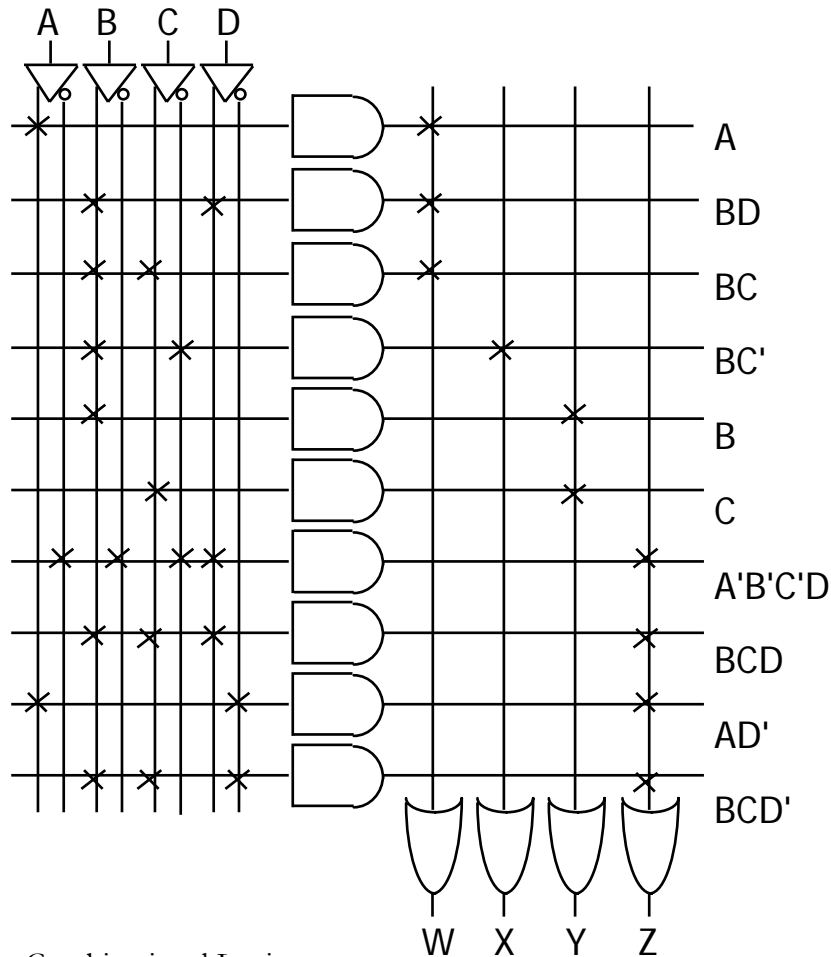
minimized functions:

$$W = A + BD + BC$$

$$X = B C'$$

$$Y = B + C$$

$$Z = A'B'C'D + BCD + AD' + B'CD'$$



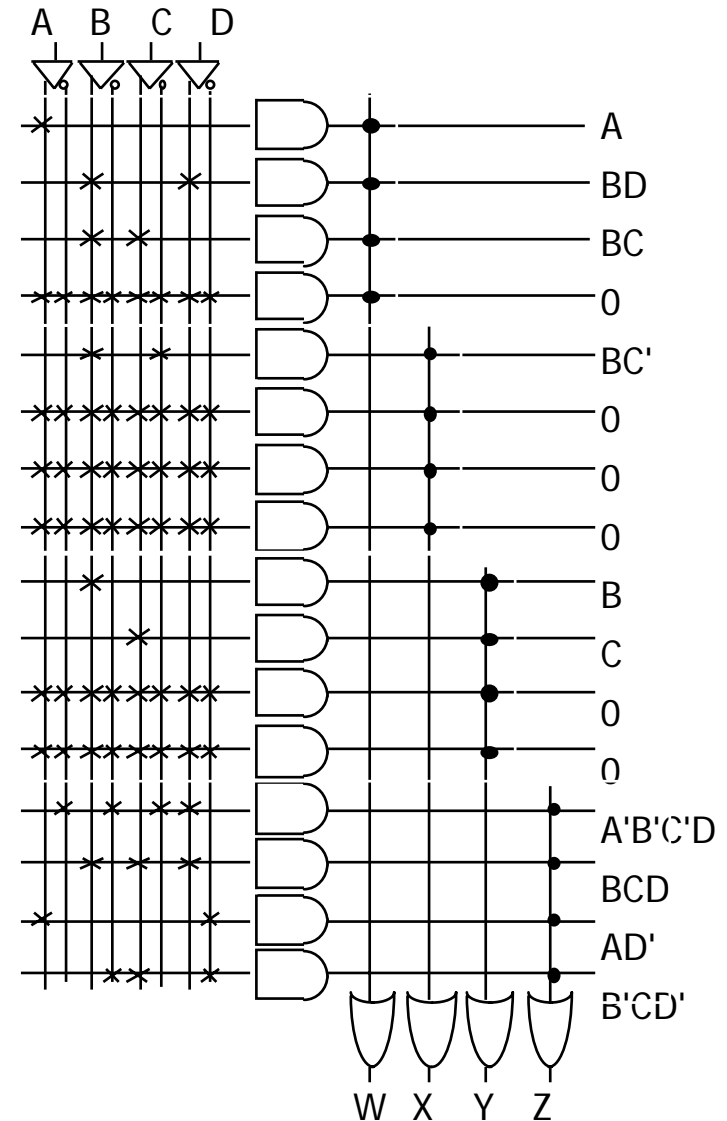
not a particularly good candidate for PLA implementation since no terms are shared among outputs

however, much more compact and regular implementation when compared with discrete AND and OR gates

# PALs and PLAs: Design Example (cont'd)

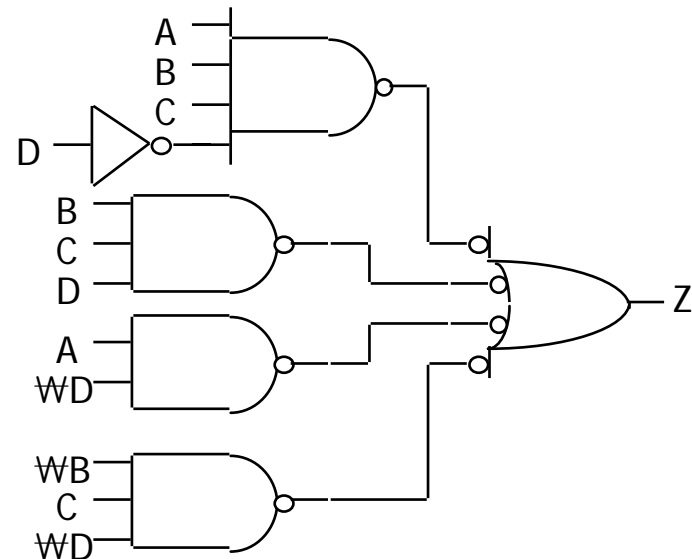
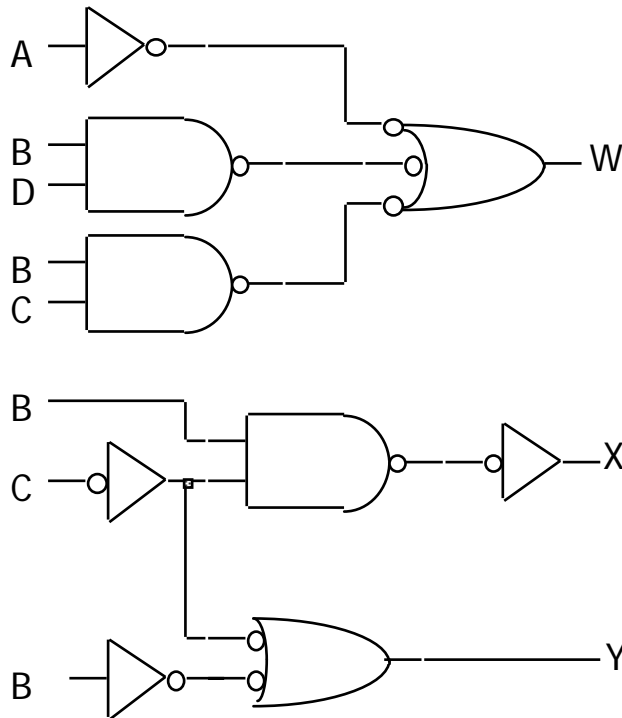
- Code converter: programmed PAL

4 product terms  
per each OR gate



# PALs and PLAs: Design Example (cont'd)

- Code converter: NAND gate implementation
  - loss of regularity, harder to understand
  - harder to make changes





# PALs and PLAs (cont'd)

## - PALs and PLAs: Another Design Example

### ■ Magnitude comparator

*Input:*  $AB, CD$  where  $A, B, C, D = \{0,1\}$

*Output:*  $EQ, NE, LT, GT = \{0,1\}$

*Function:*  $EQ = \begin{cases} 1 & \text{when } AB = CD \\ 0 & \text{otherwise} \end{cases}$

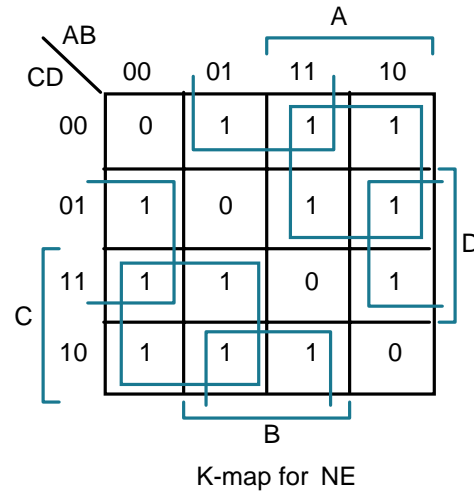
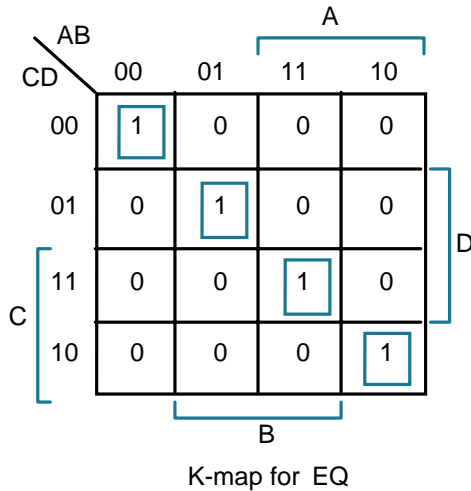
$NE = \begin{cases} 1 & \text{when } AB \neq CD \\ 0 & \text{otherwise} \end{cases}$

$LT = \begin{cases} 1 & \text{when } AB < CD \\ 0 & \text{otherwise} \end{cases}$

$GT = \begin{cases} 1 & \text{when } AB > CD \\ 0 & \text{otherwise} \end{cases}$

A	B	C	D	EQ	NE	LT	GT
0	0	0	0	1	0	0	0
0	0	0	1	0	1	1	0
0	0	1	0	0	1	1	0
0	0	1	1	0	1	1	0
0	1	0	0	0	1	0	1
0	1	0	1	1	0	0	0
0	1	1	0	0	1	1	0
0	1	1	1	0	1	1	0
1	0	0	0	0	1	0	1
1	0	0	1	0	1	0	1
1	0	1	0	1	0	0	0
1	0	1	1	0	1	1	0
1	1	0	0	0	1	0	1
1	1	0	1	0	1	0	1
1	1	1	0	0	1	0	1
1	1	1	1	1	0	0	0

# PALs and PLAs: Another Design Example (cont'd)



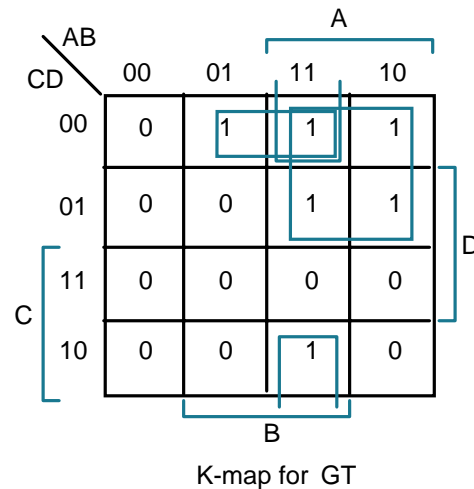
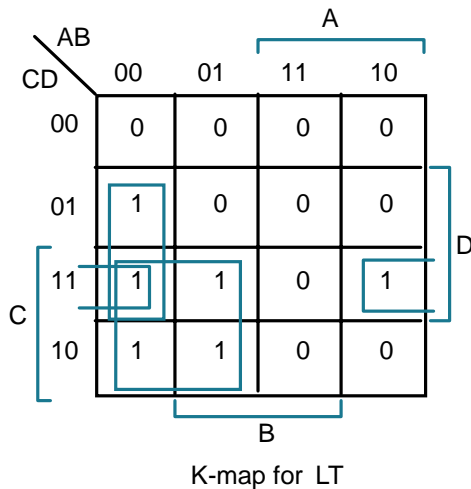
**minimized functions:**

$$EQ = A'B'C'D' + A'BC'D + ABCD + AB'CD'$$

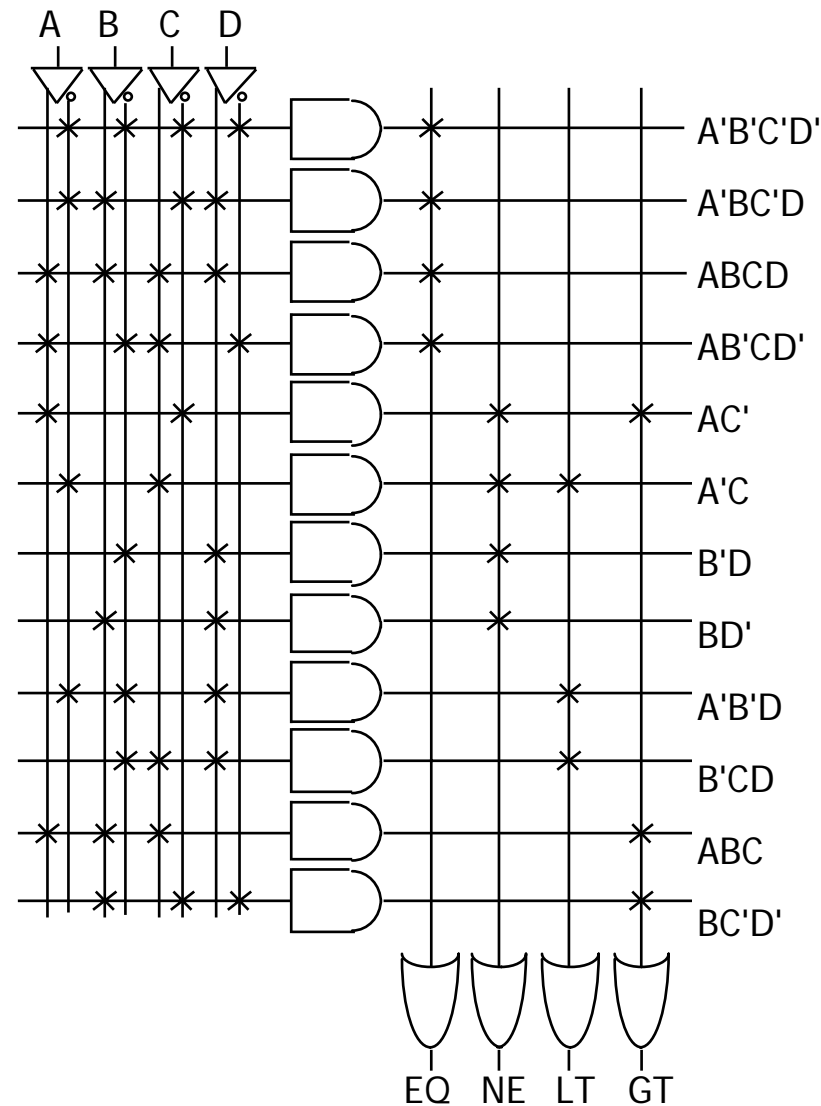
$$NE = AC' + A'C + B'D + BD'$$

$$LT = A'C + A'B'D + B'CD$$

$$GT = AC' + ABC + BC'D'$$



# PALs and PLAs: Another Design Example (cont'd)



# Basic Logic Components (cont'd)

## - Multiplexer-Based FPGA

- Actel logic module

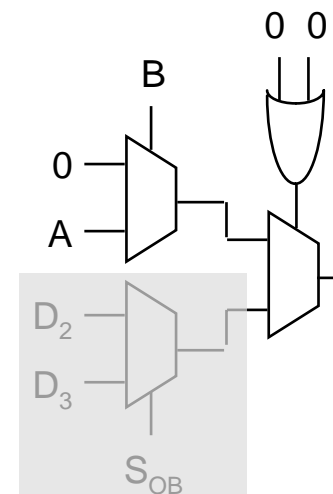
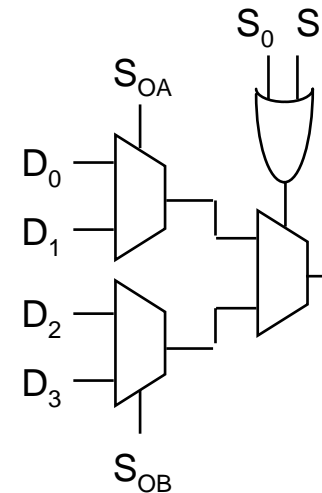
- Example using a logic module

  - AND

- $Y = (S_0 + S_1)'(S'_{OA}D_0 + S_{OA}D_1) + (S_0 + S_1)(S'_{OB}D_2 + S_{OB}D_3)$

- $Y = (0 + 0)'(B'0 + BA) + (0 + 0)(S'_{OB}D_2 + S_{OB}D_3)$

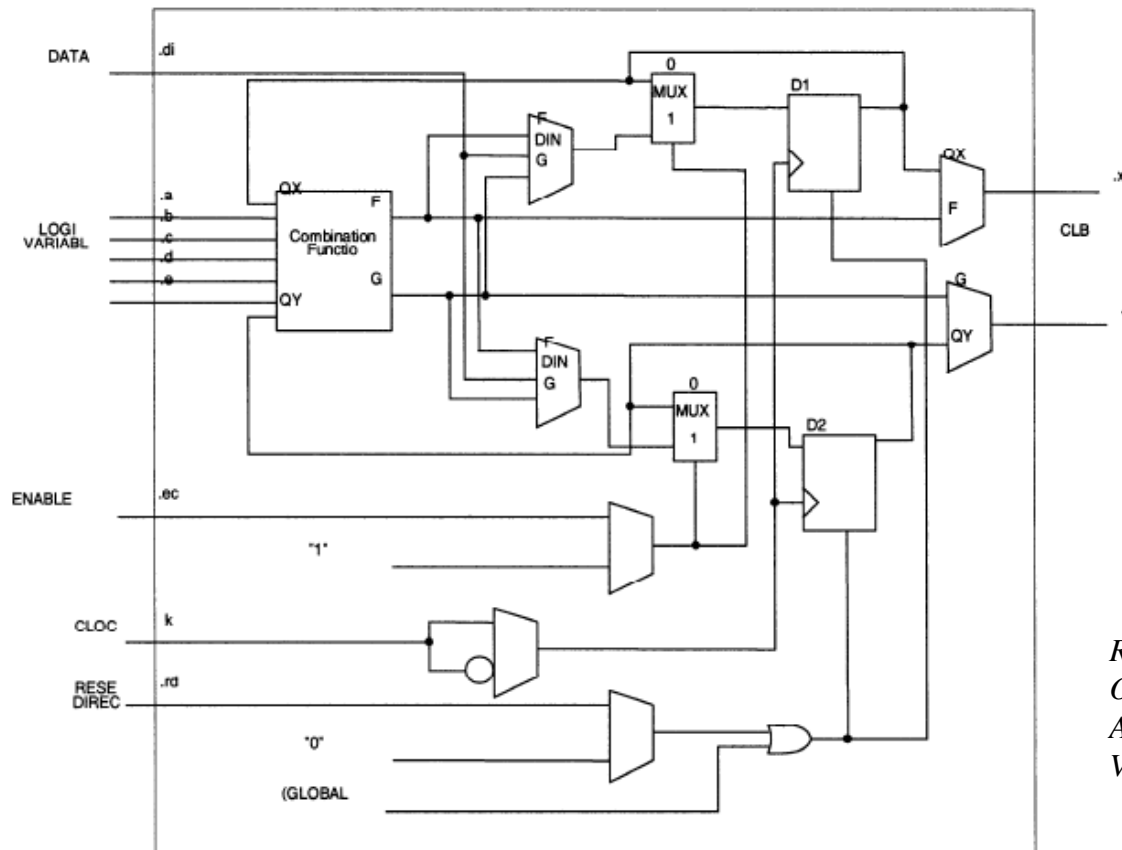
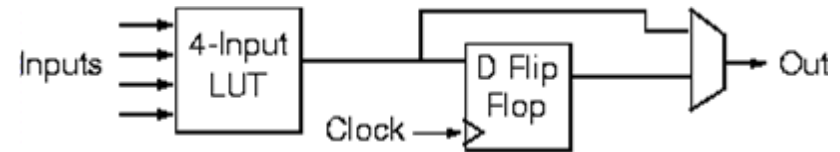
- $Y = AB$



# Basic Logic Components (cont'd)

## - Look-Up Table Based FPGA

- Simplified basic building block of FPGA
- Xilinx Configurable Logic Block (CLB)



*Ref.: IEEE TRANSACTIONS  
ON INSTRUMENTATION  
AND MEASUREMENT,  
VOL. 52, NO. 5, OCTOBER 2003*

# Two-Level and Multilevel Logic

## - Regular Logic Structures for Two-Level Logic

- ROM – full AND plane, general OR plane
  - cheap (high-volume component)
  - can implement any function of n inputs
  - medium speed
- PAL – programmable AND plane, fixed OR plane
  - intermediate cost
  - can implement functions limited by number of terms
  - high speed (only one programmable plane that is much smaller than ROM's decoder)
- PLA – programmable AND and OR planes
  - most expensive (most complex in design, need more sophisticated tools)
  - can implement any function up to a product term limit
  - slow (two programmable planes)

# Regular Logic Structures for Two-Level Logic (cont'd)

## - ROM vs. PLA

- ROM approach advantageous when
  - design time is short (no need to minimize output functions)
  - most input combinations are needed (e.g., code converters)
  - little sharing of product terms among output functions
- ROM problems
  - size doubles for each additional input
  - can't exploit don't cares
- PLA approach advantageous when
  - design tools are available for multi-output minimization
  - there are relatively few unique minterm combinations
  - many minterms are shared among the output functions
- PAL problems
  - constrained fan-ins on OR plane

# Two-Level and Multilevel Logic (cont'd)

## - Regular Logic Structures for Multilevel Logic

- Difficult to devise a regular structure for arbitrary connections between a large set of different types of gates
  - efficiency/speed concerns for such a structure
  - FPGA: just such programmable multi-level structures
    - programmable multiplexers for wiring
    - lookup tables for logic functions (programming fills in the table)
    - multi-purpose cells (utilization is the big issue)



# Regular Logic Structures for Multilevel Logic (cont'd)

## ■ Using multiple levels of PALs/PLAs/ROMs

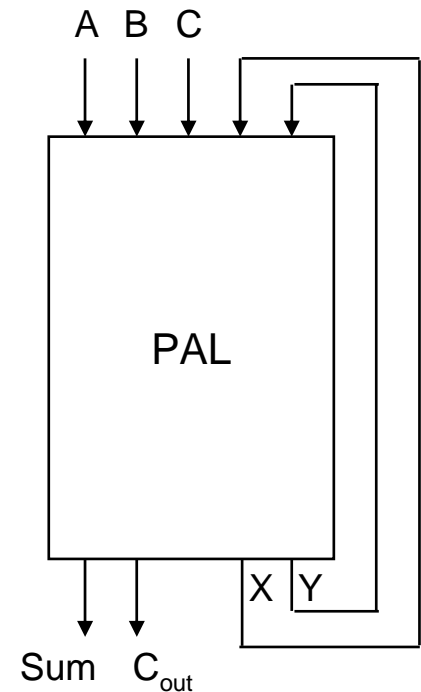
- output intermediate result
- make it an input to be used in further logic
- Example: Full Adder

### ■ Boolean equations

- $\text{Sum} = A'B'C_{in} + A'BC_{in}' + AB'C_{in}' + ABC_{in}$
- $C_{out} = AB + BC_{in} + AC_{in}$

### ■ Boolean equations for feedback PAL implementation

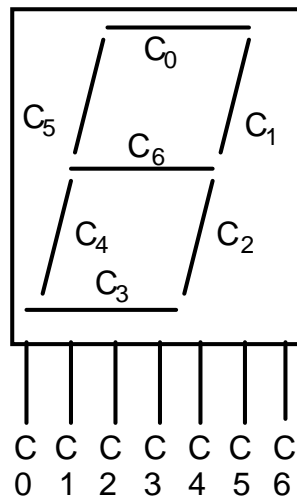
- $X = AB' + A'B$ ,  $Y = AB$
- $\text{Sum} = XC_{in}' + X'C_{in}$
- $C_{out} = XC_{in} + Y$



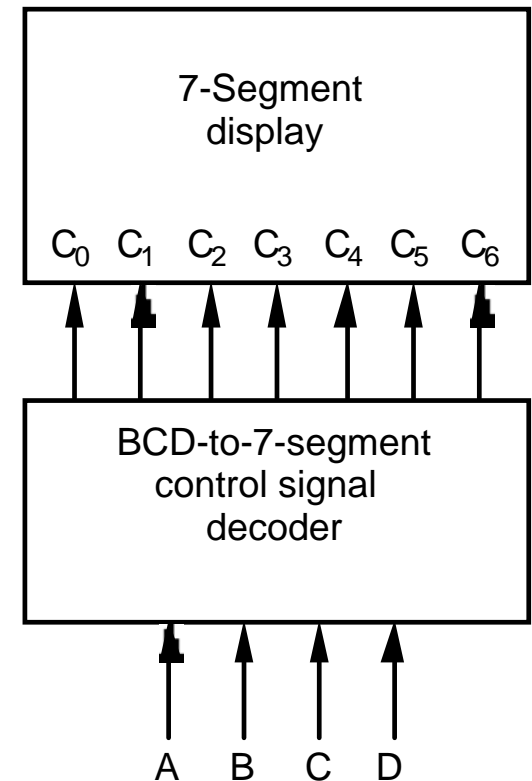
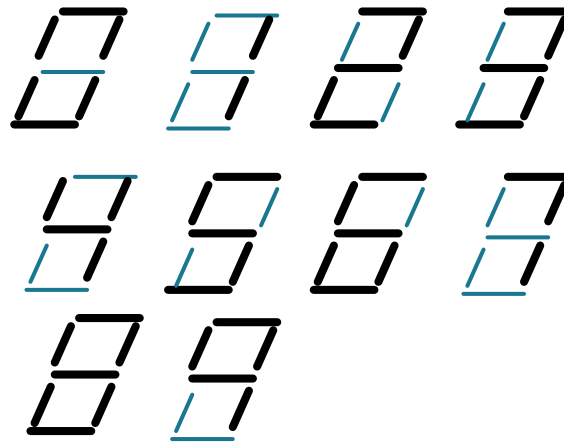
# Two-level and Multilevel Logic (cont'd)

## - 7-Segment Display Decoder

- Understanding the problem:
  - input is a 4 bit BCD digit
  - output is the control signals for the display
  - 4 inputs A, B, C, D
  - 7 outputs C0 ~ C6



**7-segment display**



**Block Diagram**

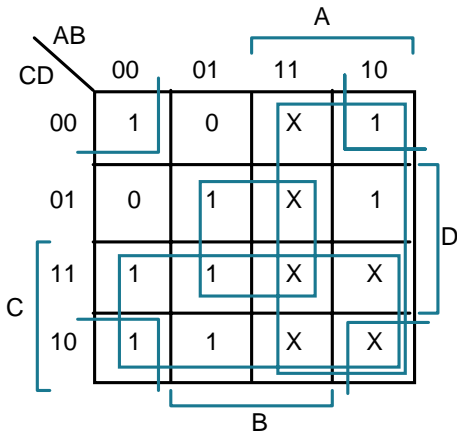
# 7-Segment Display Decoder (cont'd)

- Truth table

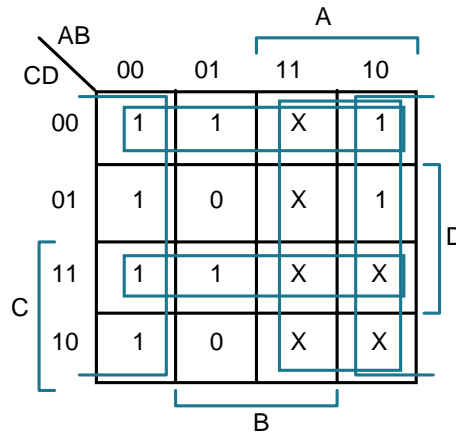
ABCD	C0	C1	C2	C3	C4	C5	C6
0000	1	1	1	1	1	1	0
0001	0	1	1	0	0	0	0
0010	1	1	0	1	1	0	1
0011	1	1	1	1	0	0	1
0100	0	1	1	0	0	1	1
0101	1	0	1	1	0	1	1
0110	1	0	1	1	1	1	1
0111	1	1	1	0	0	0	0
1000	1	1	1	1	1	1	1
1001	1	1	1	0	0	1	1
101x	x	x	x	x	x	x	x
11xx	x	x	x	x	x	x	x

- Formulate the problem in terms of a truth table
- Choose implementation target:
  - if ROM, we are done
  - don't cares imply PAL/PLA may be attractive
- Follow implementation procedure:
  - hand reduced K-maps
  - espresso

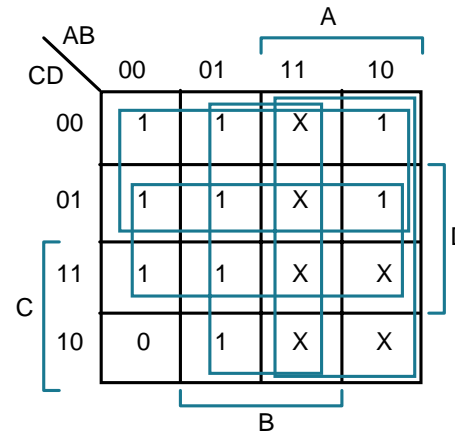
# 7-Segment Display Decoder (cont'd)



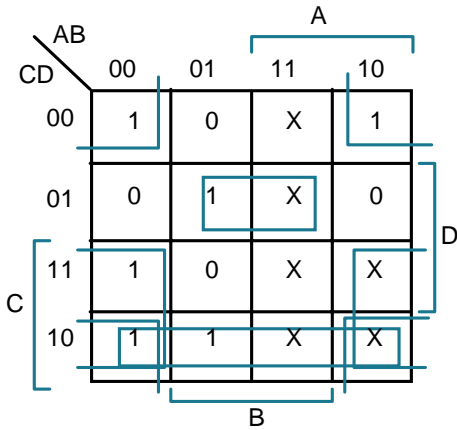
K-map for  $C_0$



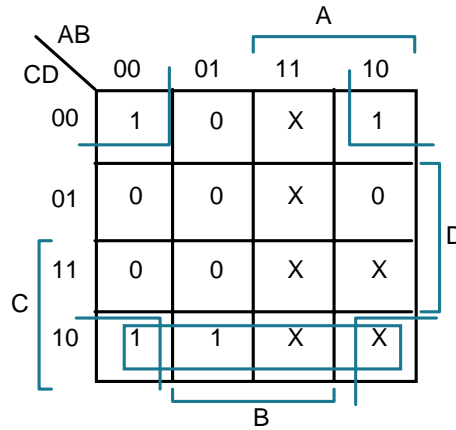
K-map for  $C_1$



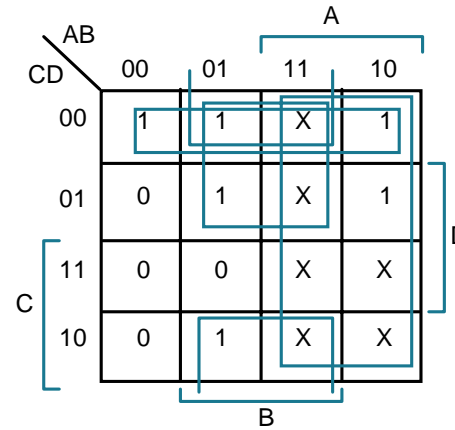
K-map for  $C_2$



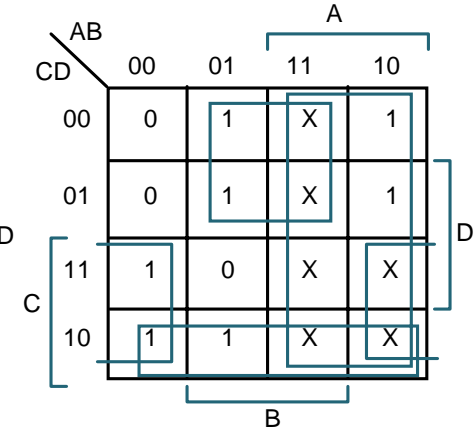
K-map for  $C_3$



K-map for  $C_4$



K-map for  $C_5$



K-map for  $C_6$

$$C_0 = A + B D + C + B' D'$$

$$C_1 = C' D' + C D + B'$$

$$C_2 = B + C' + D$$

$$C_3 = B' D' + C D' + B C' D + B' C$$

$$C_4 = B' D' + C D'$$

$$C_5 = A + C' D' + B D' + B C'$$

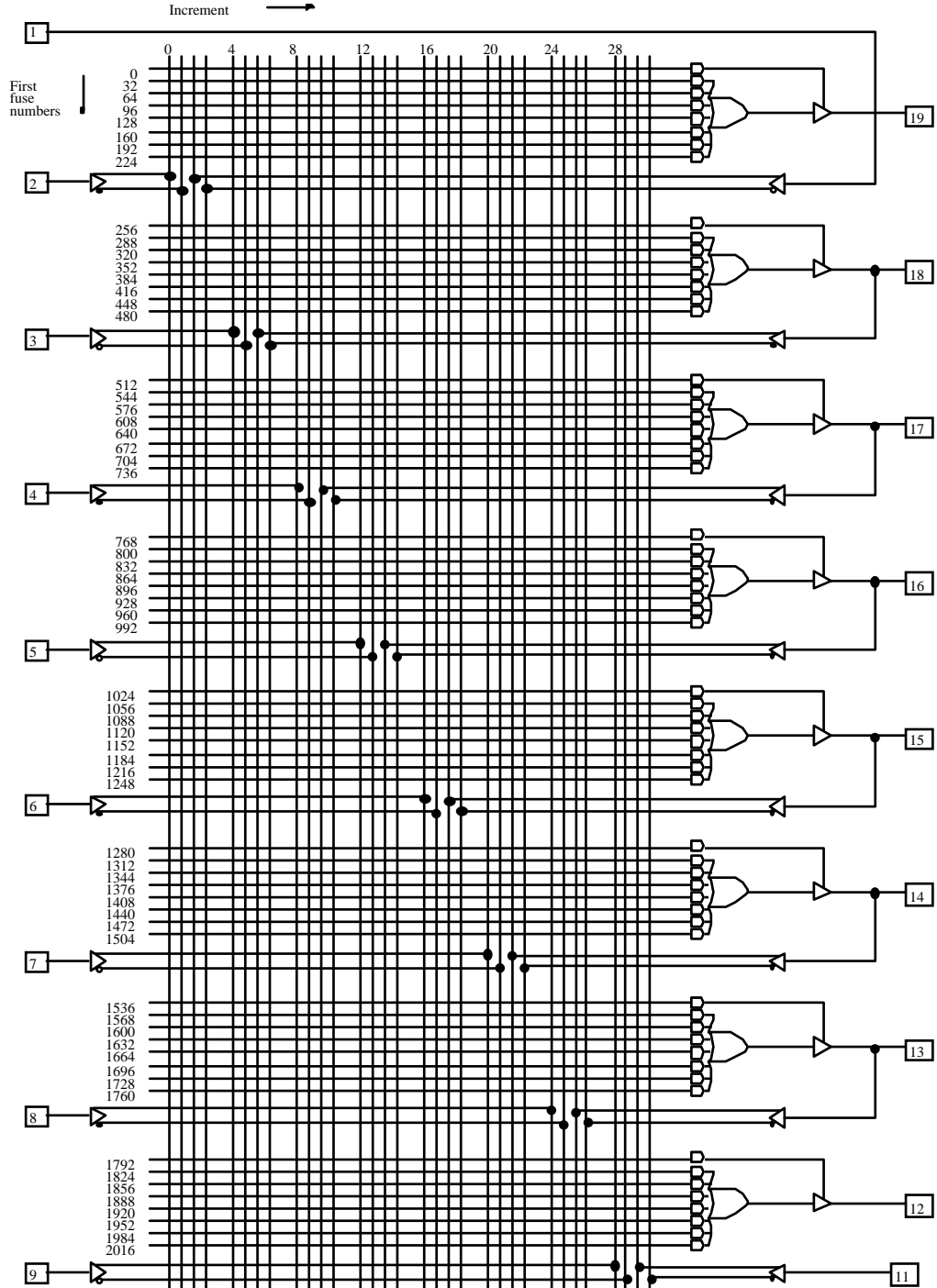
$$C_6 = A + C D' + B C' + B' C$$

**14 Unique Product Terms**

# 7-Segment Display Decoder (cont'd)

**16H8PAL**  
*(10 external inputs,  
 6 feedback inputs  
 8 outputs)*

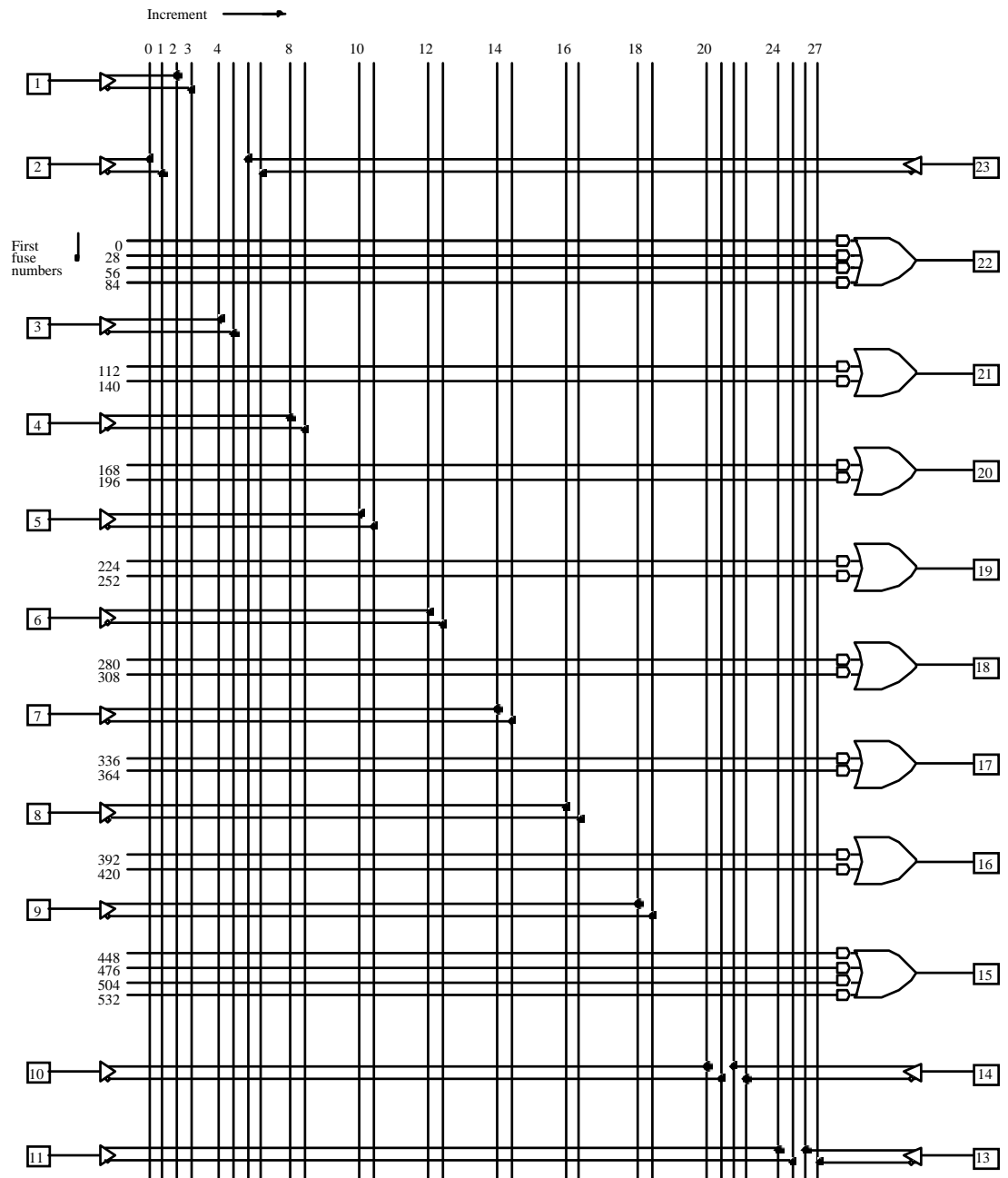
**Can Implement  
 the function**



# 7-Segment Display Decoder (cont'd)

**14H8PAL**  
*(14 inputs, 8 outputs, but only two 4-input product terms can be computed.)*

**Cannot Implement the function**



Note: Fuse number = first fuse number + increment

# 7-Segment Display Decoder (cont'd)

## CAD (espresso) input

```
.i 4
.o 7
.ilb a b c d
.ob c0 c1 c2 c3 c4 c5 c6
.p 16
0000 1111110
0001 0110000
0010 1101101
0011 1111001
0100 0110011
0101 1011011
0110 1011111
0111 1110000
1000 1111111
1001 1110011
1010 -----
1011 -----
1100 -----
1101 -----
1110 -----
1111 -----
.e
```

## CAD (espresso) output

```
.i 4
.o 7
.ilb a b c d
.ob c0 c1 c2 c3 c4 c5 c6
.p 9
-10- 0000001
-01- 0001001
-0-1 0110000
-101 1011010
--00 0110010
--11 1110000
-0-0 1101100
1--- 1000011
-110 1011111
.e
```

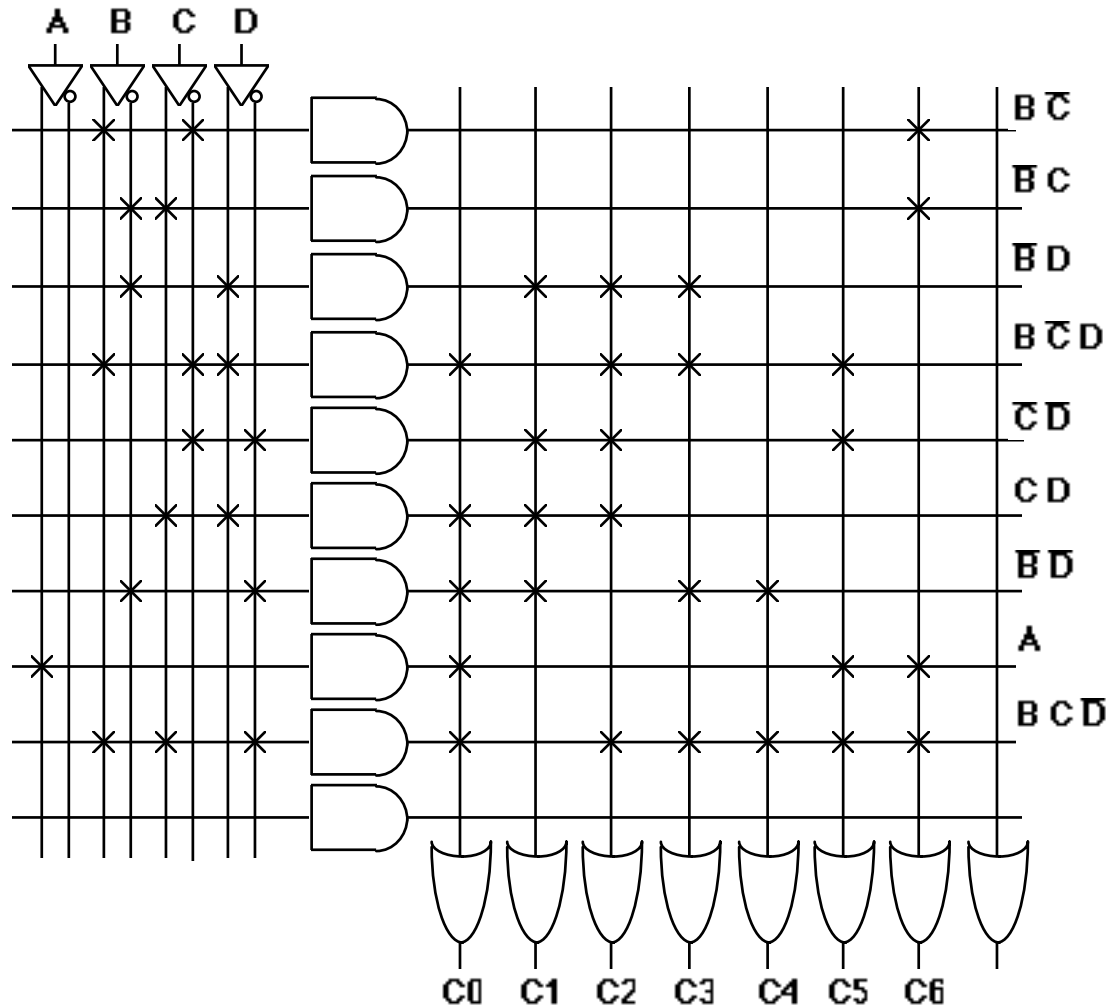
$$\begin{aligned}
 C0 &= B C' D + C D + B' D' + B C D' + A \\
 C1 &= B' D + C' D' + C D + B' D' \\
 C2 &= B' D + B C' D + C' D' + C D + B C D' \\
 C3 &= B C' D + B' D + B' D' + B C D' \\
 C4 &= B' D' + B C D' \\
 C5 &= B C' D + C' D' + A + B C D' \\
 C6 &= B' C + B C' + B C D' + A
 \end{aligned}$$

**63 Literals, 20 Gates**  
**9 Unique Product Terms!**

*The results are more complex than those of manual method. However, #unique product terms has been reduced 15->9*

# 7-Segment Display Decoder (cont'd)

## - PLA Implementation





# 7-Segment Display Decoder (cont'd)

## - Multilevel Implementation

$$X = C' + D'$$

$$Y = B' C'$$

$$C0 = C3 + A' B X' + A D Y$$

$$C1 = Y + A' C5' + C' D' C6$$

$$C2 = C5 + A' B' D + A' C D$$

$$C3 = C4 + B D C5 + A' B' X'$$

$$C4 = D' Y + A' C D'$$

$$C5 = C' C4 + A Y + A' B X$$

$$C6 = A C4 + C C5 + C4' C5 + A' B' C$$

**52 literals**

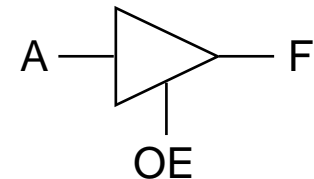
**33 gates**

**Ineffective use of don't cares**

# Non-Gate Logic

## - Tri-State Outputs

- The Third State
  - Logic States: "0", "1"
  - Don't Care/Don't Know State: "X" (must be some value in real circuit!)
  - Third State: "Z": high impedance: infinite resistance, no connection
- Tri-state gates:
  - output values are "0", "1", and "Z"
  - additional input: output enable (OE)
    - When OE is high, this gate is a non-inverting "buffer"
    - When OE is low, it is as though the gate was disconnected from the output!
  - This allows more than one gate to be connected to the same output wire, as long as only one has its output enabled at the same time

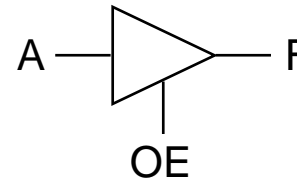


# Tri-State Outputs (cont'd)

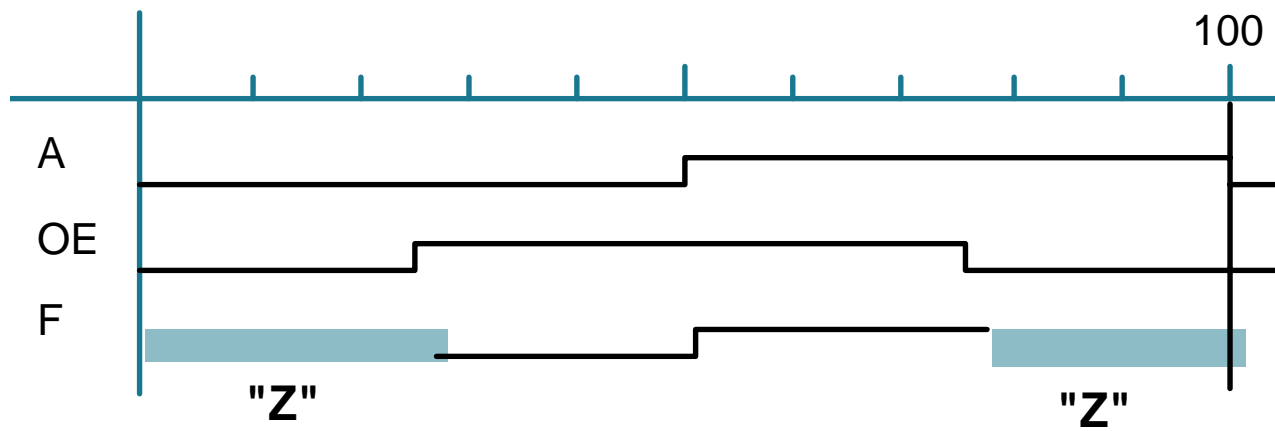
- Tri-state gates (cont'd)

- Truth table

A	OE	F
X	0	Z
0	1	0
1	1	1

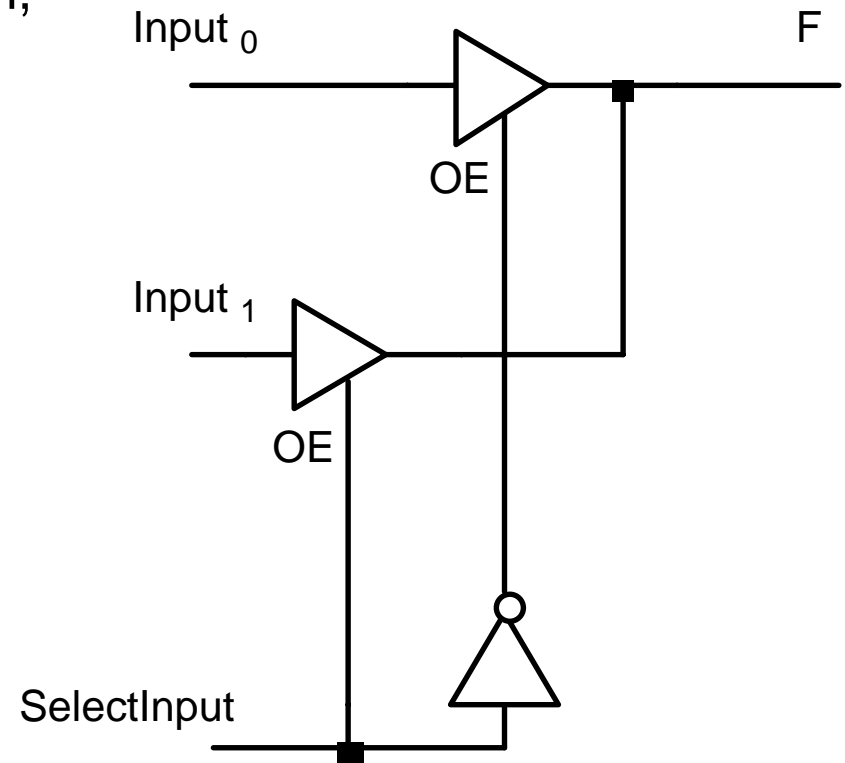


- Non-inverting buffer's timing waveform



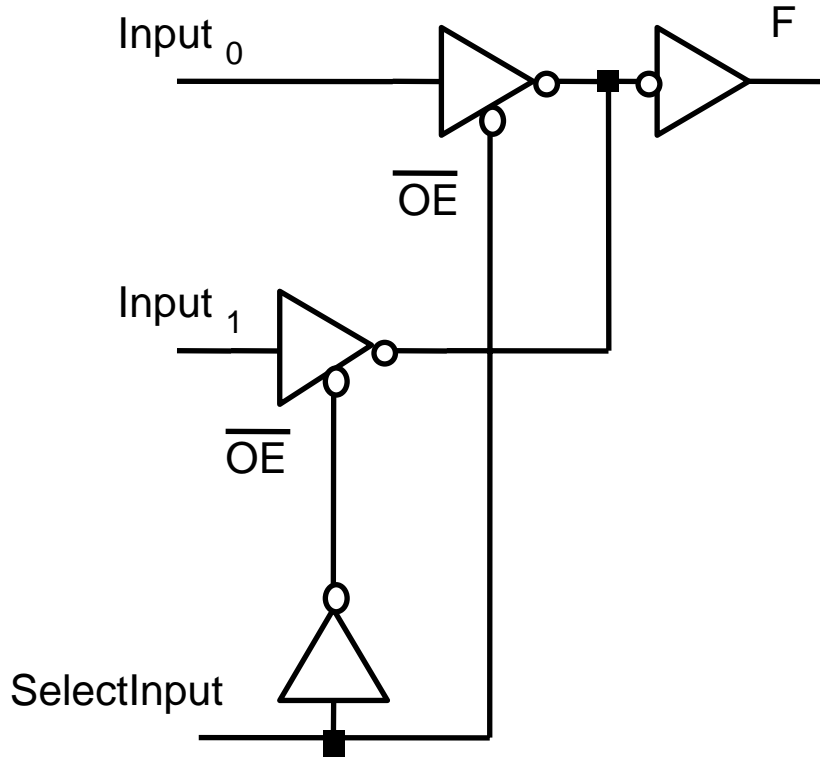
# Tri-State Outputs (cont'd)

- Using tri-state gates to implement an economical multiplexer:
  - When SelectInput is asserted high, Input1 is connected to F
  - When SelectInput is driven low, Input0 is connected to F
  - This is essentially a 2:1 Mux



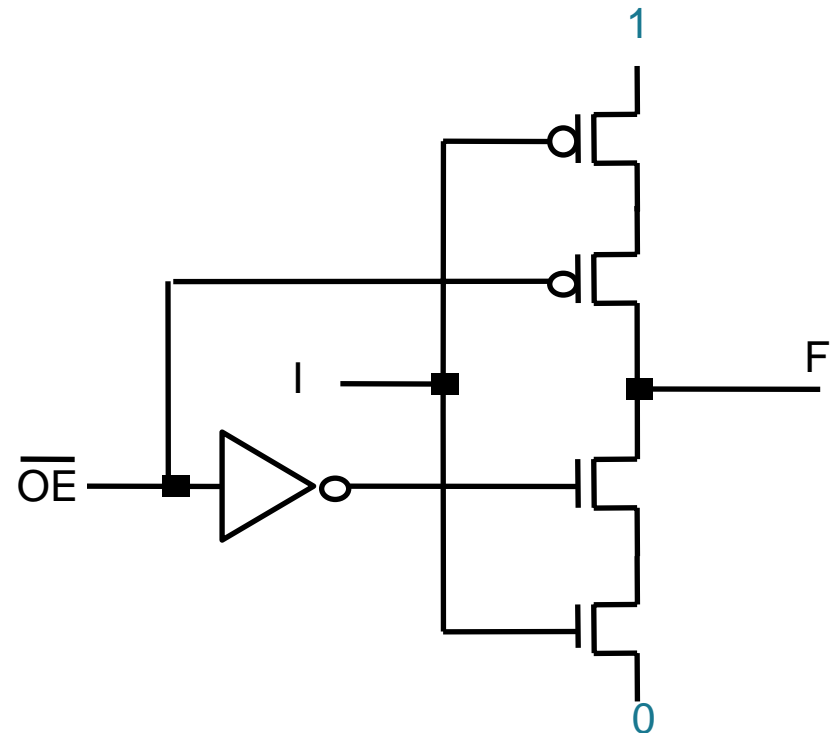
# Tri-State Outputs (cont'd)

- Alternative Tri-state Fragment



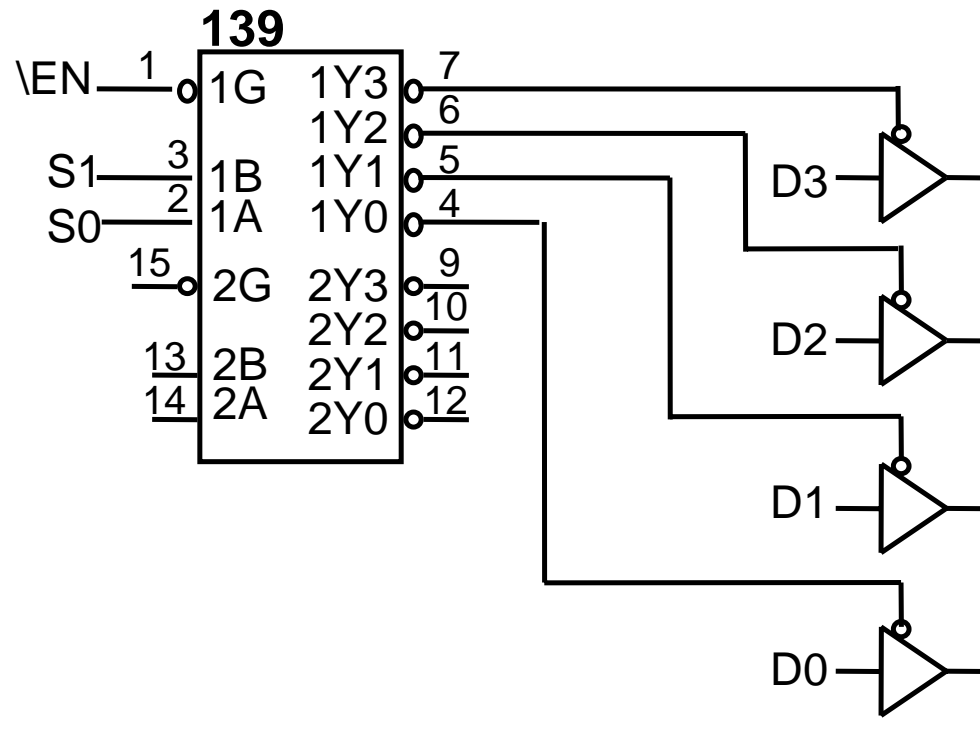
*Active low tri-state enables plus inverting tri-state buffers*

- Switch Level Implementation of tri-state gate



# Tri-State Outputs (cont'd)

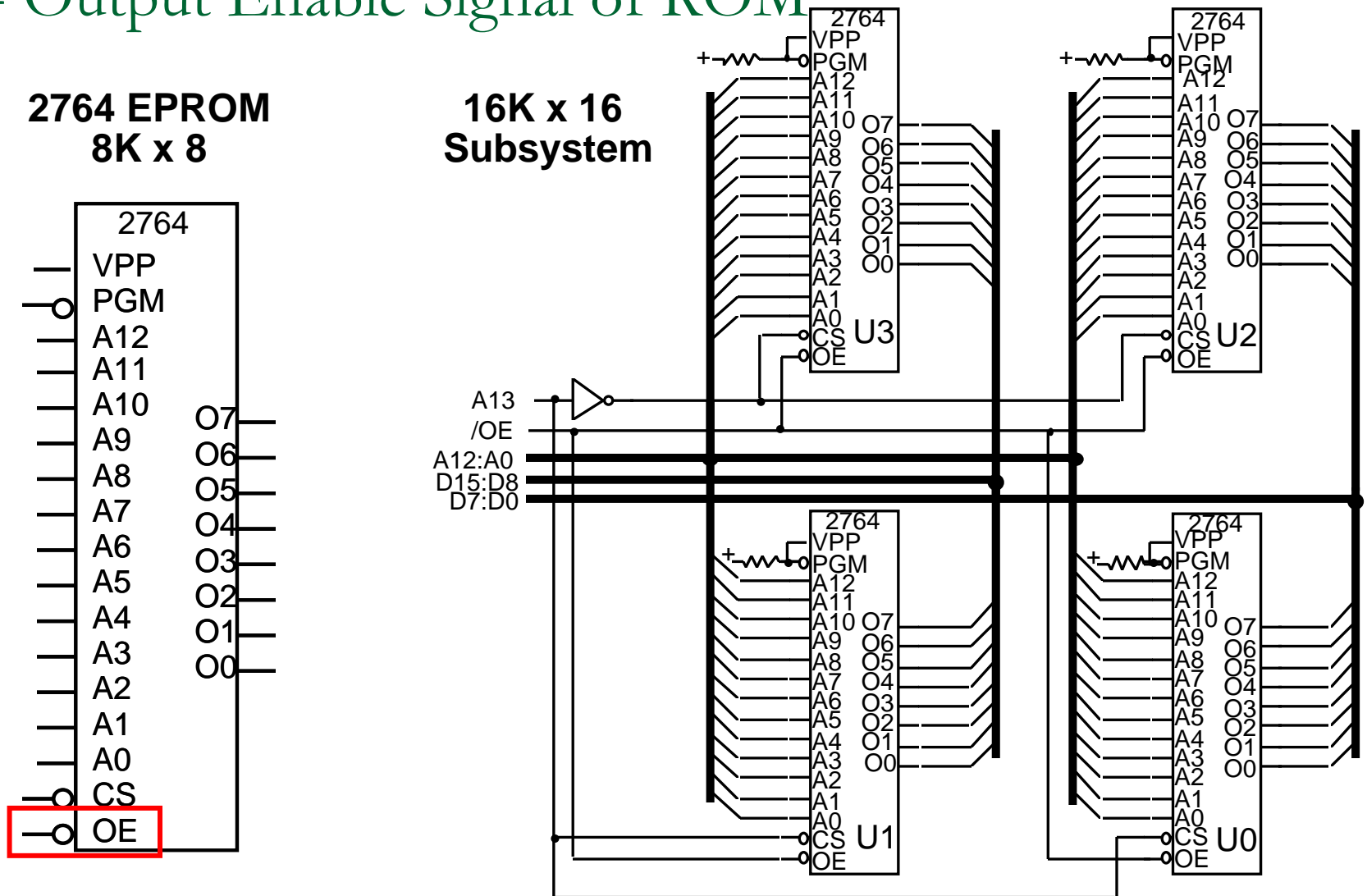
- 4:1 Multiplexer, Revisited



*Decoder + 4 tri-state Gates*

# Tri-State Outputs (cont'd)

## - Output Enable Signal of ROM



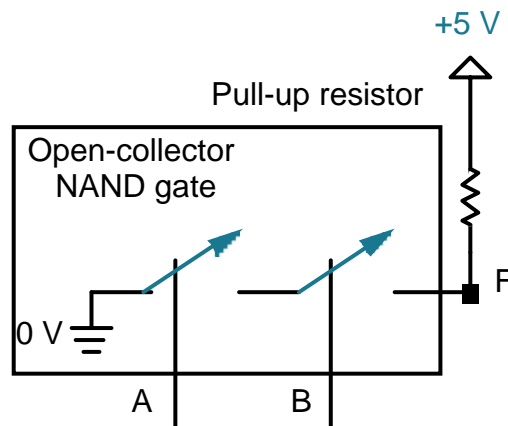
# Non-Gate Logic (cont'd)

## - Open-Collector Logic

### ■ Open Collector

- another way to connect multiple gates to the same output wire
- The gate only has the ability to pull its output low; it cannot actively drive the wire high
- this is done by pulling the wire up to a logic 1 voltage through a resistor

### ■ Switch representation of an OC NAND



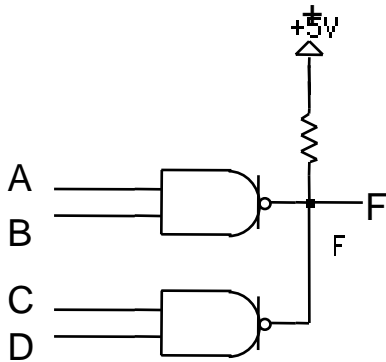
- When A=1 and B=1, equivalent circuit of OC-NAND:

$$(R_A + R_B) \cdot 0.5V$$



# Open-Collector Logic (cont'd)

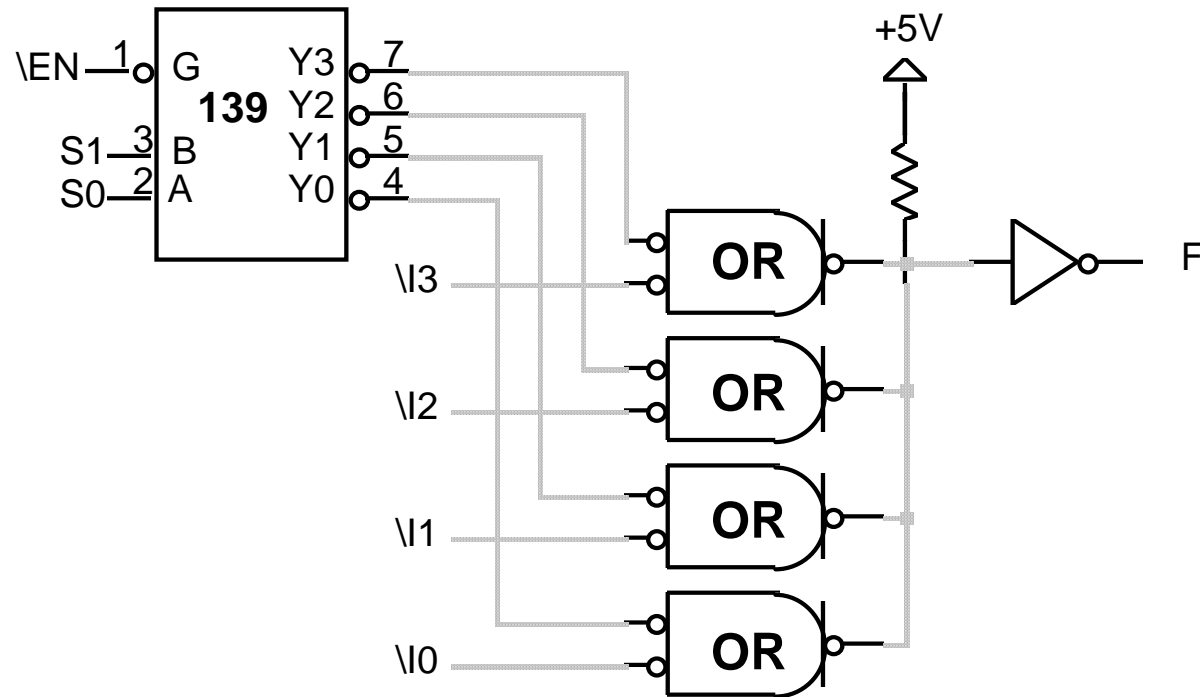
- Example: OC-NANDs in wired-AND configuration



- If A and B are "1", output is actively pulled low
- If C and D are "1", output is actively pulled low
- If one gate is low, the other high, then low wins
- If both gates are "1", the output floats, pulled high by resistor
- Hence, the two NAND functions are AND'd together!

# Open-Collector Logic (cont'd)

- 4:1 Multiplexer



*Decoder + 4 Open Collector Gates*

# Summary

- Theme:
  - How to construct digital systems with more complex logic building blocks than discrete gates
- Overview of the different classes of basic components:
  - Fixed logic
  - Look-up table based logic (ROM)
  - Steering logic (Mux, Demux)
  - PLA, PAL
  - FPGA
- Non-gate logic
  - Enables efficient multiplexing of large numbers of signals
  - Tristate
  - Open-collector