Spherical Waves and Radiation

\succ For large r,

Localized current sources radiate fields in the form of <u>Spherical Waves</u>

$$\vec{E} = \vec{a}_E ZI \frac{e^{-jkr}}{r} f(\theta, \phi)$$
$$\vec{H} = \frac{1}{\eta} \vec{a}_r \times \vec{E} \qquad \left| \vec{a}_E \right| = \left| \vec{a}_r \right| = 1$$

 $\widehat{P} = \frac{1}{2} \operatorname{Re} \left\{ \overrightarrow{E} \times \overrightarrow{H}^* \right\} = \overrightarrow{a}_r \frac{1}{2\eta} \frac{|ZI|^2}{r^2} |f(\theta, \phi)|^2$ I = termin $\overrightarrow{P} = \frac{1}{2} \operatorname{Re} \left\{ \overrightarrow{E} \times \overrightarrow{H}^* \right\} = \overrightarrow{a}_r \frac{1}{2\eta} \frac{|ZI|^2}{r^2} |f(\theta, \phi)|^2$ $\eta = 120\pi$

> Antenna pattern = $|f(\theta, \phi)|^2$



$$\begin{split} I &= terminal \ current \\ Z &= constant \ with \ units \ of \ ohms \\ \eta &= 120\pi \end{split}$$

Total Radiated Power

$$P = \int_{sphere} \vec{P} \cdot \vec{a}_r dA$$
$$dA = r^2 \sin \theta d\theta d\phi$$
$$P_T = \frac{1}{2\eta} |ZI|^2 \int_0^{2\pi} \int_0^{\pi} |f(\theta, \phi)|^2 \sin \theta d\theta d\phi$$
$$P_T \text{ is independent of } r$$

Normalization for
$$f(\theta, \phi)$$
 is:

$$\int_{0}^{2\pi} \int_{0}^{\pi} |f(\theta, \phi)|^{2} \sin \theta d\theta d\phi = 4\pi$$
Then: $\vec{P} = \vec{a}_{r} P_{T} \frac{|f(\theta, \phi)|^{2}}{4\pi r^{2}}$



Antenna Gain and Radiation Resistance

> Directive gain = $g(\theta, \phi) = |f(\theta, \phi)|^2$

- > Antenna Gain = G = Max. value of $g(\theta, \phi)$
- > If <u>isotropic antennas could exist, then</u> $|f(\theta, \phi)|^2 = 1, G = 1$
- Radiation Resistance R_r: effective resistance seen at antenna terminals

$$\frac{1}{2}|I|^2 R_r = P_T = \frac{4\pi}{2\eta}|ZI|^2$$
$$R_r = \frac{4\pi}{\eta}|Z|^2$$
$$|Z| = \sqrt{\frac{R_r\eta}{4\pi}}$$

Examples of Antennas

Half wave dipole

Dipole in corner reflector



Half Wave Dipole Antenna



Receiving Antennas and Reciprocity

$$V_{1} + V_{2}$$

$$V_{1} + V_{2}$$
For a linear two-port
$$V_{1} + V_{2}$$
Reciprocity

 $V_1 = Z_{11}I_1 + Z_{12}I_2$ $V_2 = Z_{21}I_1 + Z_{22}I_2$

 $Z_{12} = Z_{21}$



Circuit relation for Radiation into Free Space



 $P_T = (1/2)Re(V_1I_1^*) = (1/2)Re(Z_{11}|I_1|2) = (1/2)R_{r1}|I_1|2$

where R_{r1} = radiation resistance of antenna 1

Therefore: $Z_{11} = R_{r1} + jX_1$ Similarly: $Z_{22} = R_{r2} + jX_2$

where R_{r2} = radiation resistance of antenna 2

Wireless channel modeling

Received Power and Path Loss Ratio



$$I_{2} = -I_{1} \frac{Z_{12}}{\left(Z_{22} - Z_{12} + Z_{22}^{*}\right) + Z_{12}} = -I_{1} \frac{Z_{12}}{2R_{r2}}$$

$$P_{R} = \frac{1}{2} R_{r2} |I_{2}|^{2} = \frac{1}{8} \frac{|Z_{12}|^{2}}{R_{r2}} |I_{1}|^{2} = \frac{1}{8} \frac{|V_{oc}|^{2}}{R_{r2}}$$

$$P_{T} = \frac{1}{2} |I_{1}|^{2} \operatorname{Re} \left\{ \left(Z_{11} - Z_{12}\right) + \frac{\left(2R_{r2} - Z_{12}\right)Z_{12}}{\left(2R_{r2} - Z_{12}\right) + Z_{12}} \right\} = \frac{1}{2} |I_{1}|^{2} \frac{2R_{r1}R_{r2} - \operatorname{Re} \left\{Z_{12}^{2}\right\}}{2R_{r2}}$$

$$PG = \frac{P_{R}}{P_{T}} = \frac{|Z_{12}|^{2}}{4R_{r2}R_{r1} - 2\operatorname{Re} \left\{Z_{12}^{2}\right\}}$$

Effective Area of Receiving Antenna



Effective Area for Hertzian Dipole



For matched termination:

$$P_{R} = \vec{P} \cdot \vec{a}_{r} A_{e} = A_{e} \left(\frac{1}{2\eta} \left| E \right|^{2} \right) \qquad P_{R} = \frac{\left| V_{OC} \right|^{2}}{8R_{r}} = \frac{\left(l E \sin \theta \right)^{2}}{8\eta \left(l / \lambda \right)^{2} \left(4\pi / 6 \right)}$$
$$A_{e} = \frac{3}{8\pi} (\lambda \sin \theta)^{2} = g(\theta) \frac{\lambda^{2}}{4\pi}$$

Wireless channel modeling