Class Handout: Chapter 5 Input-Output Stability

2006 Fall

I. \mathcal{L} Stability

y = Hu

where $y: [0,\infty) \to \mathbb{R}^q$ and $u: [0,\infty) \to \mathbb{R}^m$ are functions, and H is a mapping (or operator) between two signals.

To measure the size of signals, let us introduce the norm function $\|\cdot\|$, with which it holds that

• ||u|| = 0 if and only if $u(t) \equiv 0$,

•
$$||au|| = |a|||u||,$$

• $||u_1 + u_2|| \le ||u_1|| + ||u_2||.$ For example,

$$\|u\|_{\mathcal{L}_{\infty}} := \sup_{t \ge 0} \|u(t)\|$$
$$\|u\|_{\mathcal{L}_{p}} := \left(\int_{0}^{\infty} \|u(t)\|^{p} dt\right)^{1/p}$$

(Here, the norm inside the integral is not necessarily equal to the *p*-norm, but usually it is.)

The set of all piecewise continuous functions whose \mathcal{L}_p $(p \in [1,\infty])$ norm is finite, is In fact, 'piecewise denoted by \mathcal{L}_p^m . Sub-, or super-script may be omitted if clear from the context.

continuous' can be replaced by 'measurable'.

"Extended space":

$$\mathcal{L}_e^m = \{ u : u_\tau \in \mathcal{L}^m, \forall \tau \in [0, \infty) \}$$

where u_{τ} is a truncation of u defined by

$$u_{\tau}(t) = \begin{cases} u(t), & 0 \le t \le \tau, \\ 0, & t > \tau \end{cases}$$

The extended space is a linear space that contains the unextended space as a subset. It allows us to deal with unbounded growing signals (e.g., u(t) = t belongs to $\mathcal{L}_{\infty,e}$).

"Causal" mapping H:

$$(Hu)_{\tau} = (Hu_{\tau})_{\tau}, \qquad \forall \tau \ge 0.$$

Causality is an intrinsic property of dynamical systems represented by state models.

Definition 5.1

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A mapping $H : \mathcal{L}_e^m \to \mathcal{L}_e^q$ is \mathcal{L} stable if \exists a class- \mathcal{K} function α , and $\beta > 0$ s.t.

$$\|(Hu)_{\tau}\|_{\mathcal{L}} \le \alpha(\|u_{\tau}\|_{\mathcal{L}}) + \beta$$

for all $u \in \mathcal{L}_e^m$ and $\tau \in [0, \infty)$.

It is finite-gain \mathcal{L} stable if \exists nonnegative constants γ and β s.t.

$$\|(Hu)_{\tau}\|_{\mathcal{L}} \le \gamma \|u_{\tau}\|_{\mathcal{L}} + \beta \tag{1}$$

for all $u \in \mathcal{L}_e^m$ and $\tau \in [0, \infty)$.

- β is a bias term, which possibly considers the initial condition.

• If (1) holds, then the system has an \mathcal{L} gain less than or equal to γ . The smallest γ s.t. (1) holds is called the gain of the system.

• For causal, \mathcal{L} stable systems, we have

$$u \in \mathcal{L}^m \Rightarrow Hu \in \mathcal{L}^q$$

and

$$||Hu||_{\mathcal{L}} \le \alpha(||u||_{\mathcal{L}}) + \beta, \qquad \forall u \in \mathcal{L}^m.$$

• \mathcal{L}_{∞} stability is the familiar notion of BIBO stability.

Example 5.1 Let

$$h(u) = a + b \tanh cu = a + b \frac{e^{cu} - e^{-cu}}{e^{cu} + e^{-cu}}$$

Then

$$|h(u)| \le a + bc|u|$$

Hence, h is finite-gain \mathcal{L}_{∞} stable with $\gamma = bc$ and $\beta = a$.

If a = 0, h is \mathcal{L}_p stable with zero bias, gain $\gamma = bc$, for each $p \in [1, \infty]$ since

$$\int_0^\infty |h(u(t))|^p dt \le (bc)^p \int_0^\infty |u(t)|^p dt.$$

If $h(u) = u^2$, it is \mathcal{L}_{∞} stable with zero bias and $\alpha(r) = r^2$, but is not finite-gain \mathcal{L}_{∞} stable.

Example 5.2 Consider a causal convolution

$$y(t) = \int_0^t h(t - \sigma) u(\sigma) d\sigma$$

where h(t) = 0 for t < 0. Suppose that $h \in \mathcal{L}_{1,e}$; that is, for every $\tau \in [0, \infty)$,

$$\|h_{\tau}\|_{\mathcal{L}_{1}} = \int_{0}^{\infty} |h_{\tau}(\sigma)| d\sigma = \int_{0}^{\tau} |h(\sigma)| d\sigma < \infty.$$

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If $u \in \mathcal{L}_{\infty,e}$ and $\tau \geq t$, then

$$\begin{aligned} |y(t)| &\leq \int_0^t |h(t-\sigma)| \ |u(\sigma)| d\sigma \\ &\leq \int_0^t |h(t-\sigma)| d\sigma \sup_{0 \leq \sigma \leq \tau} |u(\sigma)| = \int_0^t |h(s)| ds \sup_{0 \leq \sigma \leq \tau} |u(\sigma)| \end{aligned}$$

Consequently,

$$\|y_{\tau}\|_{\mathcal{L}_{\infty}} \leq \|h_{\tau}\|_{\mathcal{L}_{1}}\|u_{\tau}\|_{\mathcal{L}_{\infty}}, \qquad \forall \tau \in [0,\infty).$$

This does not mean \mathcal{L} stability of the system, because $||h_{\tau}||_{\mathcal{L}_1}$ is not uniformly bounded with respect to τ . (For example, h(t) may diverge.)

Assume that $h \in \mathcal{L}_1$; that is,

$$\|h\|_{\mathcal{L}_1} = \int_0^\infty |h(\sigma)| d\sigma < \infty.$$

Then, it holds that

$$\|y_{\tau}\|_{\mathcal{L}_{\infty}} \le \|h\|_{\mathcal{L}_{1}} \|u_{\tau}\|_{\mathcal{L}_{\infty}}, \qquad \forall \tau \in [0,\infty),$$

which means that the system is finite-gain \mathcal{L}_{∞} stable.

In fact, this assumption implies finite-gain \mathcal{L}_p stability for any $p \in [1, \infty]$. See the textbook pp. 199 and 200.

Definition 5.2 (Small-signal \mathcal{L} stability)

A mapping $H : \mathcal{L}_e^m \to \mathcal{L}_e^q$ is small-signal (finite-gain) \mathcal{L} stable if $\exists r > 0$ s.t. it is (finite-gain) \mathcal{L} stable for all $u \in \mathcal{L}_e^m$ with $\sup_{0 \le t \le \tau} \|u(t)\| \le r$.

Example 5.3 See the textbook.

II. \mathcal{L} Stability of State Models

Consider

$$\dot{x} = f(t, x, u),$$
 $x(0) = x_0,$
 $y = h(t, x, u)$

where $x \in \mathbb{R}^n$, $u \in \mathbb{R}^m$, $y \in \mathbb{R}^p$, $f : [0, \infty) \times D \times D_u \to \mathbb{R}^n$ is piecewise continuous in tand locally Lipschitz in (x, u), $h : [0, \infty) \times D \times D_u \to \mathbb{R}^q$ is piecewise continuous in t and continuous in (x, u), D is a domain containing x = 0, D_u is a domain containing u = 0.

* For each fixed $x_0 \in D$, the above system defines an operator H that assigns to each input signal u(t) the corresponding output signal y(t).

Assume that x = 0 of

$$\dot{x} = f(t, x, 0)$$

is an equilibrium.

Theorem 5.1

Let r > 0 and $r_u > 0$ s.t. $\{ \|x\| \le r \} \subset D$ and $\{ \|u\| \le r_u \} \subset D_u$. • IF x = 0 of $\dot{x} = f(t, x, 0)$ is LES, and $\exists V(t, x)$ s.t.

$$c_1 \|x\|^2 \le V(t, x) \le c_2 \|x\|^2 \tag{2}$$

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$$\frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t, x, 0) \le -c_3 \|x\|^2 \tag{3}$$

$$\left\|\frac{\partial V}{\partial x}(t,x)\right\| \le c_4 \|x\| \tag{4}$$

for all $(t, x) \in [0, \infty) \times D$, and, IF

$$\|f(t, x, u) - f(t, x, 0)\| \le L \|u\|$$
$$\|h(t, x, u)\| \le \eta_1 \|x\| + \eta_2 \|u\|$$

for all $(t, x, u) \in [0, \infty) \times D \times D_u$,

THEN, for each $||x_0|| \leq r\sqrt{c_1/c_2}$, the system is small-signal finite-gain \mathcal{L}_p stable $(p \in [1, \infty])$. In particular, for each $u \in \mathcal{L}_{p,e}$ with $\sup_{0 \leq t \leq \tau} ||u(t)|| \leq \min\{r_u, c_1c_3r/(c_2c_4L)\}$, the output y(t) satisfies

$$\|y_{\tau}\|_{\mathcal{L}_p} \le \gamma \|u_{\tau}\|_{\mathcal{L}_p} + \beta \tag{5}$$

for all $\tau \in [0, \infty)$, with

$$\gamma = \eta_2 + \frac{\eta_1 c_2 c_4 L}{c_1 c_3}, \qquad \beta = \eta_1 \|x_0\| \sqrt{\frac{c_2}{c_1}} \rho$$

where

$$\rho = \begin{cases} 1, & \text{if } p = \infty, \\ \left(\frac{2c_2}{c_3 p}\right)^{1/p}, & \text{if } p \in [1, \infty) \end{cases}$$

• IF the origin is GES and $D = R^n$, $D_u = R^m$, THEN the above holds for any x_0 and u(t).

Proof. Derivative of V along $\dot{x} = f(t, x, u)$ is

$$\dot{V} = \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t, x, 0) + \frac{\partial V}{\partial x} [f(t, x, u) - f(t, x, 0)]$$

$$\leq -c_3 \|x\|^2 + c_4 L \|x\| \|u\|$$

(At time point, the proof for global case is actually done already. Can you see?)

(For numerical analysis, we continue...) Take $W(t) = \sqrt{V(t, x(t))}$.

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If $V(t, x(t)) \neq 0$,

$$\dot{W} = \frac{\dot{V}}{2\sqrt{V}} \le \frac{1}{2\sqrt{V}} \left(-c_3 \|x\|^2 + c_4 L \|x\| \|u\| \right)$$
$$\le \frac{1}{2\sqrt{V}} \left(-\frac{c_3}{c_2} V(t, x) + c_4 L \frac{\sqrt{V}}{\sqrt{c_1}} \|u\| \right)$$
$$= -\frac{c_3}{2c_2} W + \frac{c_4 L}{2\sqrt{c_1}} \|u\|$$

If V(t, x(t)) = 0 (i.e., x(t) = 0), it can be shown that

$$D^+W(t) \le \frac{c_4L}{2\sqrt{c_1}} \|u(t)\|.$$

Hence, for all V(t, x(t)), we have

$$D^+W(t) \le -\frac{1}{2}\frac{c_3}{c_2}W + \frac{c_4L}{2\sqrt{c_1}}\|u(t)\|.$$

By the comparison lemma,

$$W(t) \le e^{-t\frac{c_3}{2c_2}}W(0) + \frac{c_4L}{2\sqrt{c_1}} \int_0^t e^{-(t-\tau)\frac{c_3}{2c_2}} \|u(\tau)\| d\tau.$$

We then obtain (since $\sqrt{c_1} \|x(t)\| \le W(t) \le \sqrt{c_2} \|x(t)\|$)

$$\begin{aligned} \|x(t)\| &\leq \sqrt{\frac{c_2}{c_1}} \|x_0\| e^{-t\frac{c_3}{2c_2}} + \frac{c_4L}{2c_1} \int_0^t e^{-(t-\tau)\frac{c_3}{2c_2}} \|u(\tau)\| d\tau \\ &= \sqrt{\frac{c_2}{c_1}} \|x_0\| e^{-t\frac{c_3}{2c_2}} + \frac{c_4L}{2c_1}\frac{2c_2}{c_3} \left(1 - e^{-t\frac{c_3}{2c_2}}\right) \left(\sup_{0 \leq \sigma \leq t} \|u(\sigma)\|\right) \end{aligned}$$

(We now check if $x(t) \in B_r$ for all $t \ge 0$ so that the whole analysis is valid.) Then, since

$$||x_0|| \le r\sqrt{\frac{c_1}{c_2}}, \qquad \sup_{0\le \sigma\le t} ||u(\sigma)|| \le \frac{c_1c_3r}{c_2c_4L},$$

we have

$$||x(t)|| \le re^{-t\frac{c_3}{2c_2}} + \left(1 - e^{-t\frac{c_3}{2c_2}}\right)r = r.$$

(We now obtain (5).) From the assumption, we have

$$||y(t)|| \le k_1 e^{-at} + k_2 \int_0^t e^{-a(t-\tau)} ||u(\tau)|| d\tau + k_3 ||u(t)||$$

where

$$k_1 = \sqrt{\frac{c_2}{c_1}} \|x_0\| \eta_1, \qquad k_2 = \frac{c_4 L \eta_1}{2c_1}, \qquad k_3 = \eta_2, \qquad a = \frac{c_3}{2c_2}.$$

 Set

$$y_1(t) = k_1 e^{-at}, \qquad y_2(t) = k_2 \int_0^t e^{-a(t-\tau)} ||u(\tau)|| d\tau, \qquad y_3(t) = k_3 ||u(t)||.$$

Then, for any $p \in [1, \infty]$, we have

 $\|y_{2,\tau}\|_{\mathcal{L}_p} \le \frac{k_2}{a} \|u_{\tau}\|_{\mathcal{L}_p}$

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Exercise 5.6 and Exercise 3.24 are employed here.

since $||h||_{\mathcal{L}_1} = 1/a$ and from Example 5.2, and

$$\|y_{3,\tau}\|_{\mathcal{L}_p} \le k_3 \|u_\tau\|_{\mathcal{L}_p}$$

For the term y_1 ,

$$\|y_{1,\tau}\|_{\mathcal{L}_p} \le k_1 \rho$$

where

$$\rho = \begin{cases} 1, & \text{if } p = \infty, \\ \left(\frac{1}{ap}\right)^{1/p}, & \text{if } p \in [1, \infty) \end{cases}$$

Thus, by the triangular inequality,

$$\gamma = k_3 + \frac{k_2}{a}, \qquad \beta = k_1 \rho.$$

On the other hand, the global case follows easily.

Corollary 5.1

IF the origin of $\dot{x} = f(t, x, 0)$ is GES(LES) and

$$\|f(t, x, u) - f(t, x, 0)\| \le L \|u\|$$
$$\|h(t, x, u)\| \le \eta_1 \|x\| + \eta_2 \|u\|$$

THEN the original system is (small-signal) finite-gain \mathcal{L}_p stable for each $p \in [1, \infty]$.

Corollary 5.2

A LTI system is finite-gain \mathcal{L}_p stable if A is Hurwitz.

(In particular,

$$\gamma = \|D\|_2 + \frac{2\lambda_{\max}^2(P)\|B\|_2\|C\|_2}{\lambda_{\min}(P)}$$
$$\beta = \|C\|_2\|x_0\|\sqrt{\frac{\lambda_{\max}(P)}{\lambda_{\min}(P)}}\rho$$
$$\rho = \begin{cases} 1, \qquad p = \infty, \\ \left(\frac{2\lambda_{\max}(P)}{p}\right)^{\frac{1}{p}}, \quad p \in [1,\infty). \end{cases}$$

where $A^T P + P A = -I$.)

Example 5.4

 $\dot{x} = -x - x^3 + u,$ $x(0) = x_0,$ $y = \tanh x + u$

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With $V(x) = x^2/2$, we can show that the system is finite-gain \mathcal{L}_p stable.

Let us consider LUAS case restricting to \mathcal{L}_{∞} stability.

Theorem 5.2 (local version)

Let r > 0 and $r_u > 0$ s.t. $\{||x|| \le r\} \subset D$ and $\{||u|| \le r_u\} \subset D_u$. IF x = 0 of $\dot{x} = f(t, x, 0)$ is LUAS, and $\exists V(t, x)$ s.t.

$$\begin{aligned} \alpha_1(\|x\|) &\leq V(t,x) \leq \alpha_2(\|x\|) \\ \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t,x,0) \leq -\alpha_3(\|x\|) \\ \left\| \frac{\partial V}{\partial x}(t,x) \right\| \leq \alpha_4(\|x\|) \end{aligned}$$

for all $(t, x) \in [0, \infty) \times D$, and,

 \mathbf{IF}

$$\|f(t, x, u) - f(t, x, 0)\| \le \alpha_5(\|u\|)$$
$$\|h(t, x, u)\| \le \alpha_6(\|x\|) + \alpha_7(\|u\|) + \eta$$

for all $(t, x, u) \in [0, \infty) \times D \times D_u$,

THEN, for each $||x_0|| \leq \alpha_2^{-1}(\alpha_1(r))$, the system is small-signal \mathcal{L}_{∞} stable.

Proof.

Derivative of V along $\dot{x}=f(t,x,u)$ is

$$\dot{V} = \frac{\partial V}{\partial t} + \frac{\partial V}{\partial x} f(t, x, 0) + \frac{\partial V}{\partial x} [f(t, x, u) - f(t, x, 0)]$$

$$\leq -\alpha_3(\|x\|) + \alpha_4(\|x\|)\alpha_5(\|u\|)$$

$$\leq -(1 - \theta)\alpha_3(\|x\|) - \theta\alpha_3(\|x\|) + \alpha_4(r)\alpha_5\left(\sup_{0 \le t \le \tau} \|u(t)\|\right)$$

where $0 < \theta < 1$. Set

$$\mu = \alpha_3^{-1} \left(\frac{\alpha_4(r)\alpha_5 \left(\sup_{0 \le t \le \tau} \|u(t)\| \right)}{\theta} \right).$$

We consider only such u(t) that $\sup_{0 \le t \le \tau} \|u(t)\|$ is small enough for $\mu < \alpha_2^{-1}(\alpha_1(r))$. Then,

$$\dot{V} \le -(1-\theta)\alpha_3(\|x\|), \qquad \forall \|x\| \ge \mu.$$

From Theorem 4.18,

$$\|x(t)\| \le \beta(\|x_0\|, t) + \gamma\left(\sup_{0 \le t \le \tau} \|u(t)\|\right)$$

for all $0 \leq t \leq \tau$. Hence,

$$\begin{aligned} \|y(t)\| &\leq \alpha_6 \left(\beta(\|x_0\|, t) + \gamma \left(\sup_{0 \leq t \leq \tau} \|u(t)\|\right)\right) + \alpha_7(\|u(t)\|) + \eta \\ &\leq \alpha_6(2\beta(\|x_0\|, t)) + \alpha_6 \left(2\gamma \left(\sup_{0 \leq t \leq \tau} \|u(t)\|\right)\right) + \alpha_7(\|u(t)\|) + \eta \end{aligned}$$

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Therefore, we have

$$\begin{split} \|y_{\tau}\|_{\mathcal{L}_{\infty}} &\leq \gamma_0(\|u_{\tau}\|_{\mathcal{L}_{\infty}}) + \beta_0\\ \gamma_0 &= \alpha_6 \circ 2\gamma + \alpha_7, \qquad \beta_0 = \alpha_6(2\beta(\|x_0\|, 0)) + \eta \end{split}$$

Theorem 5.3 (global version)

IF $D = \mathbb{R}^n$, $D_u = \mathbb{R}^m$, the system $\dot{x} = f(t, x, u)$ is ISS, and

$$||h(t, x, u)|| \le \alpha_1(||x||) + \alpha_2(||u||) + \eta,$$

THEN, the system is \mathcal{L}_{∞} stable.

Proof. Trivial. Isn't it?

* Think about why the global asymptotic stable case needs so strong property (ISS)? Consider

$$\dot{x} = -\frac{x}{1+x^2} + u,$$

which is GAS (but not GES), and is not \mathcal{L}_{∞} stable (also not ISS).

III. \mathcal{L}_2 Gain

" \mathcal{L}_2 stability plays a special role in systems analysis. It is natural to work with squareintegrable signals, which can be viewed as finite-energy signals. In many control problems, the system is represented as an input-output map, from a disturbance input u to a controlled output y, which is required to be small."

Here we study how to calculate the \mathcal{L}_2 gain for TI systems.

Theorem 5.4 Consider

$$\dot{x} = Ax + Bu$$
$$y = Cx + Du$$

where A is Hurwitz. Let $G(s) = C(sI - A)^{-1}B + D$. Then, the \mathcal{L}_2 gain is

$$\sup_{\omega \in \mathbb{R}} \|G(j\omega)\|_2 = \sup_{\omega \in \mathbb{R}} \sqrt{\lambda_{\max}[G^T(-j\omega)G(j\omega)]}.$$

Proof. Due to linearity, we set x(0) = 0. From Fourier transform theory, for a causal signal $y \in \mathcal{L}_2$,

$$Y(j\omega) = \int_0^\infty y(t)e^{-j\omega t}dt, \qquad Y(j\omega) = G(j\omega)U(j\omega).$$

By Parseval's theorem,

$$\begin{split} \|y\|_{\mathcal{L}_{2}}^{2} &= \int_{0}^{\infty} y^{T}(t)y(t)dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} Y^{*}(j\omega)Y(j\omega)d\omega \\ &= \frac{1}{2\pi} \int_{-\infty}^{\infty} U^{*}(j\omega)G^{T}(-j\omega)G(j\omega)U(j\omega)d\omega \\ &\leq \left(\sup_{\omega \in \mathbb{R}} \|G(j\omega)\|_{2}\right)^{2} \frac{1}{2\pi} \int_{-\infty}^{\infty} U^{*}(j\omega)U(j\omega)d\omega \\ &= \left(\sup_{\omega \in \mathbb{R}} \|G(j\omega)\|_{2}\right)^{2} \|u\|_{\mathcal{L}_{2}}^{2} \end{split}$$

which shows that the \mathcal{L}_2 gain is less than or equal to $\sup_{\omega \in \mathbb{R}} \|G(j\omega)\|_2$. See Appendix C.10 to show that the \mathcal{L}_2 gain is, in fact, equal to $\sup_{\omega \in \mathbb{R}} \|G(j\omega)\|_2$.

Consider

$$\dot{x} = f(x) + G(x)u,$$
 $x(0) = x_0,$
 $y = h(x)$

where f is locally Lipschitz, G and h are continuous, and f(0) = 0 and h(0) = 0.

Theorem 5.5

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IF $\exists C^1$ positive semidefinite function V(x) and $\gamma > 0$ s.t.

$$\mathcal{H}(V, f, G, h, \gamma) := \frac{\partial V}{\partial x} f(x) + \frac{1}{2\gamma^2} \frac{\partial V}{\partial x} G(x) G^T(x) \left(\frac{\partial V}{\partial x}\right)^T + \frac{1}{2} h^T(x) h(x) \le 0,$$

THEN, for each $x_0 \in \mathbb{R}^n$, the system is finite-gain \mathcal{L}_2 stable and the gain $\leq \gamma$.

* Hamilton-Jacobi inequality

 \ast Compare with Theorem 5.1—no exponential stability, but needs a solution to HJ eq.

Proof.

$$\begin{split} \dot{V} &= L_f V(x) + L_G V(x) u = -\frac{1}{2} \gamma^2 \| u - \frac{1}{\gamma^2} (L_G V)^T(x) \|_2^2 + L_f V(x) \\ &+ \frac{1}{2\gamma^2} L_G V(x) (L_G V)^T(x) + \frac{1}{2} \gamma^2 \| u \|_2^2 \\ &\leq \frac{1}{2} \gamma^2 \| u \|_2^2 - \frac{1}{2} \| y \|_2^2 - \frac{1}{2} \gamma^2 \| u - \frac{1}{\gamma^2} (L_G V)^T(x) \|_2^2 \\ &\leq \frac{1}{2} \gamma^2 \| u \|_2^2 - \frac{1}{2} \| y \|_2^2 \end{split}$$

Thus,

$$V(x(\tau)) - V(x(0)) \le \frac{1}{2}\gamma^2 \int_0^\tau \|u(t)\|_2^2 dt - \frac{1}{2}\int_0^\tau \|y(t)\|_2^2 dt.$$

Since $V(x) \ge 0$, we have

$$\int_0^\tau \|y(t)\|_2^2 \le \gamma^2 \int_0^\tau \|u(t)\|_2^2 dt + 2V(x_0).$$

Then,

$$||y_{\tau}||_{\mathcal{L}_{2}} \leq \gamma ||u_{\tau}||_{\mathcal{L}_{2}} + \sqrt{2V(x_{0})}.$$

Example 5.8

$$\dot{x}_1 = x_2$$

 $\dot{x}_2 = -ax_1^3 - kx_2 + u, \qquad a > 0, k > 0,$
 $y = x_2$

Let $V(x) = \alpha(ax_1^4/4 + x_2^2/2)$. Then, $L_f V = -\alpha kx_2^2$, $L_G V = \alpha x_2$, $h(x) = x_2$, so that

$$\mathcal{H} = \left(-\alpha k + \frac{\alpha^2}{2\gamma^2} + \frac{1}{2}\right) x_2^2.$$

That is, if

$$\gamma^2 \ge \frac{\alpha^2}{2\alpha k - 1}$$

then the system is \mathcal{L}_2 stable. The right-hand side has a minimum $1/k^2$ for $\alpha = 1/k$, so with $\gamma = 1/k$, we conclude that the system is finite-gain \mathcal{L}_2 stable with the gain less than or equal to 1/k.

Example 5.9 Consider the system with the property

$$L_f W(x) \le -kh^T(x)h(x), \qquad k > 0$$
$$L_G W(x) = h^T(x)$$

where W(x) is a C^1 positive semidefinite function.

Let $V(x) = \alpha W(x)$. Then,

$$\mathcal{H} = \left(-\alpha k + \frac{\alpha^2}{2\gamma^2} + \frac{1}{2}\right)h^T(x)h(x).$$

By Example 5.8, we conclude that the system is finite-gain \mathcal{L}_2 stable with the gain $\leq 1/k$.

Example 5.10 Consider the system with the property

$$L_f W(x) \le 0$$

 $L_G W(x) = h^T(x)$

where W(x) is a C^1 positive semidefinite function.

Let an output feedback control

$$u = -ky + v, \qquad k > 0.$$

Then, the closed-loop system becomes

$$\dot{x} = f(x) - kG(x)G^{T}(x)\left(\frac{\partial W}{\partial x}\right)^{T} + G(x)v$$
$$y = h(x)$$

This system is, in fact, the case of Example 5.9. (Verify!)

So, this system is finite-gain \mathcal{L}_2 stable with the gain $\leq 1/k$ from v to y. Note that the gain can be arbitrarily assigned with the feedback.

Example 5.11 Consider

$$\dot{x} = Ax + Bu$$
$$y = Cx$$

Suppose $\exists P \ge 0$ s.t.

$$PA + A^T P + \frac{1}{\gamma^2} PBB^T P + C^T C = 0$$

(the Riccati equation), with some $\gamma > 0$.

Then, $V(x) = \frac{1}{2}x^T P x$ satisfies the HJ equation for this system, i.e., $\mathcal{H} = 0$. Thus, the system is finite-gain \mathcal{L}_2 stable with the gain $\leq \gamma$. (In fact, the Riccati equation has a solution $P \geq 0$ if and only if the system's \mathcal{L}_2 gain is less than or equal to γ .)

• From the proof of Theorem 5.5, we note that, if the assumptions hold only on a finite domain D, then we obtain the same conclusion as long as the solution x(t) stays in D. (Corollary 5.4.) This possibly restricts the class of input signal, which also depends on the initial condition x_0 .

• If $\dot{x} = f(x)$ is AS and if $||x_0||$ and $\sup_{0 \le t \le \tau} ||u(t)||$ are sufficiently small (relative to D), then the solution x(t) remains in the neighborhood of the origin. This leads to the following Lemma.

Consider

$$\dot{x} = f(x) + G(x)u, \qquad x(0) = x_0,$$
$$y = h(x)$$

where f is C^1 , G and h are continuous on a set $D \subset \mathbb{R}^n$, and f(0) = 0 and h(0) = 0.

Lemma 5.1

IF the origin of $\dot{x} = f(x)$ is AS, and $\exists C^1$ positive semidefinite function V(x) and $\gamma > 0$ s.t.

$$\mathcal{H}(V, f, G, h, \gamma) := \frac{\partial V}{\partial x} f(x) + \frac{1}{2\gamma^2} \frac{\partial V}{\partial x} G(x) G^T(x) \left(\frac{\partial V}{\partial x}\right)^T + \frac{1}{2} h^T(x) h(x) \le 0$$

on a domain D,

THEN $\exists k_1 > 0$ s.t. for each $||x_0|| \leq k_1$, the system is small-signal finite-gain \mathcal{L}_2 stable with the gain $\leq \gamma$.

Proof. We can apply Corollary 5.4, if we show that x(t) stays in a neighborhood of the origin.

By the converse theorem, $\exists a C^1$ function W(x) and $r_0 > 0$ s.t.

$$\alpha_1(\|x\|) \le W(x) \le \alpha_2(\|x\|)$$
$$\frac{\partial W}{\partial x} f(x) \le -\alpha_3(\|x\|)$$

for all $x \in B_{r_0}$, and without loss of generality, we assume that $B_{r_0} \subset D$.

Let k and L be an upper bound of $\|\partial W/\partial x\|$ and $\|G(x)\|$, respectively. Then,

$$\begin{split} \dot{W}(x,u) &\leq \frac{\partial W}{\partial x} f(x) + \frac{\partial W}{\partial x} G(x) u \leq -\alpha_3(\|x\|) + kL \|u\| \\ &\leq -(1-\theta)\alpha_3(\|x\|) - \theta\alpha_3(\|x\|) + kL \sup_{0 \leq t \leq \tau} \|u(t)\| \\ &\leq -(1-\theta)\alpha_3(\|x\|), \qquad \forall \|x\| \geq \alpha_3^{-1} \left(kL \sup_{0 \leq t \leq \tau} \|u(t)\|/\theta\right) \end{split}$$

where $0 < \theta < 1$.

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This implies that (you may recall Theorem 4.18), $\exists k_1, k_2 > 0$ s.t., for every $||x_0|| < k_1$ and for inputs such that $\sup_{0 \le t \le \tau} ||u(t)|| \le k_2$, the solution $x(t) \in B_{r_0}$. Hence, the conclusion follows from Corollary 5.4.

Lemma 5.2

IF $\exists C^1$ positive *semidefinite* function V(x) and $\gamma > 0$ s.t.

$$\mathcal{H}(V, f, G, h, \gamma) := \frac{\partial V}{\partial x} f(x) + \frac{1}{2\gamma^2} \frac{\partial V}{\partial x} G(x) G^T(x) \left(\frac{\partial V}{\partial x}\right)^T + \frac{1}{2} h^T(x) h(x) \le 0$$

on D, and 'no solution of $\dot{x} = f(x)$ can stay identically in $S = \{x \in D : h(x) = 0\}$ other than the trivial solution $x(t) \equiv 0$ ',

THEN the origin of $\dot{x} = f(x)$ is AS, and $\exists k_1 > 0$ s.t. for each $||x_0|| \leq k_1$, the system is small-signal finite-gain \mathcal{L}_2 stable with the gain $\leq \gamma$.

- In this lemma the Lyapunov function V(x) came from the HJ inequality.
- How to interpret the phrase: 'no solution of $\dot{x} = f(x)$ can stay identically in $S = \{x \in D :$

h(x) = 0} other than the trivial solution $x(t) \equiv 0$? (Not clear in the textbook. Discuss it in the proof.) I would say 'there is no *locally invariant set* in S except the origin'. By a locally invariant set L in S, I mean that, if $x(0) = x_0 \in L$, there exist nonnegative constants δ_1 and δ_2 , not both zero, s.t. $x(t) \in L$ for $t \in [-\delta_1, \delta_2]$.

Locally invariant set \subset Invariant set

Proof. The proof is done by Lemma 5.1 if we show that the origin of $\dot{x} = f(x)$ is AS. From the HJ inequality,

$$\frac{\partial V}{\partial x}f(x) \leq -\frac{1}{2\gamma^2}\frac{\partial V}{\partial x}G(x)G^T(x)\left(\frac{\partial V}{\partial x}\right)^T - \frac{1}{2}h^T(x)h(x) \leq -\frac{1}{2}h^T(x)h(x),$$

which seemingly means that the origin is stable, and by employing the LaSalle's theorem, we have the conclusion. (In particular, the set $\{x \in D : \frac{\partial V}{\partial x}f(x) = 0\} \subset S$. In addition, since there is no locally invariant set in S except the origin, the only invariant set in S is the origin. So, the LaSalle's theorem can be applied.) This is true if V(x) is positive definite.

However, since V(x) is assumed just positive semidefinite, we now prove that, from the assumption, the function V(x) is in fact positive definite.

Let r > 0 be s.t. $B_r \subset D$ (strict inclusion). Then, for each $x \in B_r$, $\exists \delta > 0$ s.t. the solution from x, $\phi(t; x)$, stays in D for $t \in [0, \delta]$. Integrating the above inequality over $[0, \delta]$, we obtain

$$V(\phi(\delta;x)) - V(x) \le -\frac{1}{2} \int_0^\delta \|h(\phi(t;x))\|_2^2 dt.$$

Using $V(\phi(\delta; x)) \ge 0$, we have

$$V(x) \ge \frac{1}{2} \int_0^\delta \|h(\phi(t;x))\|_2^2 dt.$$

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Now suppose that $\exists x \neq 0$ s.t. V(x) = 0 (not positive definite). Then, the above implies that $h(\phi(t; x)) = 0$ for $t \in [0, \delta]$. This means that $x \neq 0$ is an element of locally invariant set. However, since the only element of the locally invariant set is the origin, x = 0 (contradiction).

Example 5.12 DIY.

IV. FEEDBACK SYSTEMS: THE SMALL-GAIN THEOREM



From the textbook,

• The formalism of input-output stability is particularly useful in studying stability of interconnected systems, since the gain of a system allows us to track how the norm of a signal increases or decreases as it passes through the system.

• See Figure 5.1. $H_1: \mathcal{L}_e^m \to \mathcal{L}_e^q$ and $H_2: \mathcal{L}_e^q \to \mathcal{L}_e^m$. Suppose both system are finite-gain \mathcal{L} stable; that is,

$$\begin{aligned} \|y_{1,\tau}\|_{\mathcal{L}} &\leq \gamma_1 \|e_{1,\tau}\|_{\mathcal{L}} + \beta_1, \qquad \forall e_1 \in \mathcal{L}_e^m, \forall \tau \in [0,\infty) \\ \|y_{2,\tau}\|_{\mathcal{L}} &\leq \gamma_2 \|e_{2,\tau}\|_{\mathcal{L}} + \beta_2, \qquad \forall e_2 \in \mathcal{L}_e^q, \forall \tau \in [0,\infty) \end{aligned}$$

(finite-gain for simplicity).

• Suppose that the feedback system is *well defined* in the sense that for every pair of inputs $u_1 \in \mathcal{L}_e^m$ and $u_2 \in \mathcal{L}_e^q$, \exists unique outputs $e_1, y_2 \in \mathcal{L}_e^m$ and $e_2, y_1 \in \mathcal{L}_e^q$.

• Define $u = [u_1; u_2]$, $y = [y_1; y_2]$, $e = [e_1; e_2]$. We ask if the feedback system, viewed as a mapping from u to y (or to e since they are equivalent in the sense of Exercise 5.21, your homework), is finite-gain \mathcal{L} stable.

• If the external input and output of interest is just u_1 and y_1 , why should we consider \mathcal{L} stability from both (u_1, u_2) to both (y_1, y_2) ? The reason is that, by considering all the pairs of inputs and outputs, any possibly hidden internal mode comes out. (See Exercise 5.20.)

Theorem 5.6 (Small-gain Theorem)

IF $\gamma_1 \gamma_2 < 1$,

THEN the feedback system is finite-gain ${\cal L}$ stable.

Proof.

$$e_{1\tau} = u_{1\tau} - (H_2 e_2)_{\tau}, \qquad e_{2\tau} = u_{2\tau} + (H_1 e_1)_{\tau}$$

Then,

$$\begin{aligned} \|e_{1\tau}\|_{\mathcal{L}} &\leq \|u_{1}\tau\|_{\mathcal{L}} + \|(H_{2}e_{2})_{\tau}\|_{\mathcal{L}} \leq \|u_{1}\tau\|_{\mathcal{L}} + \gamma_{2}\|e_{2\tau}\|_{\mathcal{L}} + \beta_{2} \\ &\leq \|u_{1}\tau\|_{\mathcal{L}} + \gamma_{2}(\|u_{2\tau}\|_{\mathcal{L}} + \gamma_{1}\|e_{1\tau}\|_{\mathcal{L}} + \beta_{1}) + \beta_{2} \\ &= \gamma_{1}\gamma_{2}\|e_{1\tau}\|_{\mathcal{L}} + (\|u_{1\tau}\|_{\mathcal{L}} + \gamma_{2}\|u_{2\tau}\|_{\mathcal{L}} + \beta_{2} + \gamma_{2}\beta_{1}) \end{aligned}$$

Since $\gamma_1 \gamma_2 < 1$,

$$\|e_{1\tau}\|_{\mathcal{L}} \le \frac{1}{1 - \gamma_1 \gamma_2} (\|u_{1\tau}\|_{\mathcal{L}} + \gamma_2 \|u_{2\tau}\|_{\mathcal{L}} + \beta_2 + \gamma_2 \beta_1)$$

for all $\tau \in [0, \infty)$. Similarly,

$$\|e_{2\tau}\|_{\mathcal{L}} \leq \frac{1}{1 - \gamma_1 \gamma_2} (\|u_{2\tau}\|_{\mathcal{L}} + \gamma_1 \|u_{1\tau}\|_{\mathcal{L}} + \beta_1 + \gamma_1 \beta_2)$$

for all $\tau \in [0,\infty)$. The proof completes since $\|e\|_{\mathcal{L}} \le \|e_1\|_{\mathcal{L}} + \|e_2\|_{\mathcal{L}}$.

Example 5.17
H₁
$$(A: Humitz)$$

 $G_{1}(f) = C(SZ-A)^{-}B$
 $J_{2} = Y(A, e_{1}) = e_{1}$
 $H_{2} = V(A, e_{2}) = v_{1}$
 $H_{2} = Y_{2} = Y(A, e_{2}) = v_{1}$
 $H_{2} = Y_{2} = (A + e_{1}) = v_{2}$
 $H_{2} = Y_{2} = (A + e_{2}) = v_{2}$
 $H_{1} : finite - jain d_{2} gain : Y_{1} = ap ||G_{2}(ju)||_{2}$
 $H_{2} : v_{2} = Y_{2}$
 $H_{2} : v_{3} = y_{3}$
 $H_{2} : v_{3} = y_{3}$
 $H_{3} : finite - jain d_{2} gain : Y_{3} = y_{3}$

$$\begin{array}{rcl} & & & & \\ & & & \\ & & & \\ & & & \\ & & \\ & & \\ & & \\ & = \\ \end{array} \begin{array}{rcl} & & & \\ & & \\ & & \\ \end{array} \begin{array}{rcl} & & \\ \end{array} \end{array} \begin{array}{rcl}$$

Contailer u= r(x) only considers of (not -2).

$$\frac{Ass}{x}: \quad \forall (x) \text{ is designed } p.t.$$

$$\dot{x} = f(x, \quad \forall (x) \neq d)$$

$$\|x\|_{\mathcal{L}} \leq \sum \|d\|_{\mathcal{L}} \neq \beta \quad \lambda < \sigma$$
what about the aduator dynamics?

× Robustness to unmodeled actuator dynamics

$$\dot{x} = f(x, Cz + d_l)$$

 $\epsilon \dot{z} = Az + B(-F(x) + dz)$

$$\dot{x} = f(x, C(\gamma - A^{-1}B(\gamma(x) + dz)) + d_1)$$

$$= f(x, C\gamma + \gamma(x) + d_1 + dz)$$

$$\epsilon \dot{\gamma} = A(\gamma - A^{-1}B(\gamma + dz)) + B(\gamma + dz) + \epsilon A^{-1}B(\dot{\gamma} + dz)$$

$$= A\gamma + \epsilon A^{-1}B(\dot{\gamma} + dz)$$

$$(\dot{\gamma} = \frac{\partial \gamma}{\partial x} f(\gamma, \gamma + d + c\gamma))$$

$$H_{1}$$

$$d \qquad e_{1} \qquad \boxed{x = f(x, r(x) + e_{1})} \qquad y_{1} = \overrightarrow{y}$$

$$y_{1} = \frac{\partial x}{\partial x} f(x, r(x) + e_{1})$$

$$H_{2}$$

$$\frac{\eta}{z} = \frac{1}{z} A \eta + A^{-1} B e_{2} \qquad e_{2}$$

$$y_{2} = -C \eta$$

$$\underline{Asr}: \left\| \frac{\partial \mathcal{S}}{\partial x} \otimes f(x, r(x + e_{0})) \right\| \leq \alpha \|x\| + \alpha \|e_{1}\|$$

Hi:
$$\|y_i\|_{\mathcal{L}} \leq (c_1\lambda + c_2) \|e_i\|_{\mathcal{L}} + c_i f$$

= λ_i

H2: Hurwitz LTI (Constary 5.2)

$$|| y_2 ||_{\mathcal{L}} = \frac{2\lambda \max^2 (p) || B ||_2 || C ||_2}{\lambda \min (p)} \quad || e_2 ||_{\mathcal{L}} + \binom{2}{2}$$

$$= \varepsilon \lambda_f \quad || e_2 ||_{\mathcal{L}} + \binom{2}{2}$$

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