Electrical Systems I

Basic Elements

- Active Elements: OP Amp etc.

 (Has transistors/amplifiers that require active source of power to work)
- Passive Elements: Inductor, Resistor, Capacitor etc. (Simply respond to an applied voltage or current.)

- Current : the rate of flow of charge
- Charge: (electric charge) the integral of current with respect to time [C]

$$i = \frac{dq}{dt}$$
 [ampere]= $\frac{[\text{coulomb}]}{[\text{sec}]}$

• Voltage: electromotive force needed to produce a flow of current in a wire [V] Change in energy as the charge is passed through a component.

$$[V] = [J/C]$$

• Power: product of voltage and current

$$[W] = [J/Sec]$$

Basic Elements - Resistance

•Resistance: the change in voltage required to make a unit change in current.

Analogous to →

$$R = \frac{Change \ in \ voltage}{Change \ in \ current} = \frac{[V]}{[A]} = [Ohm(\Omega)]$$

• Resistor

$$V_R = R \cdot i_R$$
 $R = \frac{V_R}{i_R}$

Basic Elements - Capacitance

•Capacitance: the change in the quantity of electric charge required to make a unit change in voltage.

Analogous to →

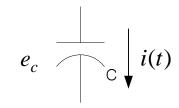
$$C = \frac{[Coulomb]}{[V]} = [Farad(F)]$$

• Capacitor: two conductor separated by non-conducting medium.

$$i = dq / dt$$
, $e_c = q / C \rightarrow i =$

$$\therefore e_c(t) =$$

$$I(s) = V(s) =$$



Basic Elements - Inductance

• Inductance: An electromotive force induced in a circuit, if the circuit lies in a time -varying magnetic field.

Analogous to →

$$L = \frac{[V]}{[A/\sec]} = [Henry(H)]$$

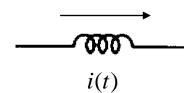
• Inductor:
$$e_L$$

$$r_L =$$

$$V(s) =$$

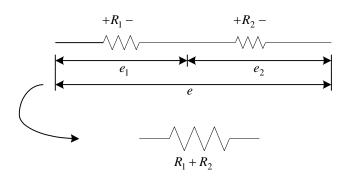
$$\therefore i_L(t) =$$

$$I(s) =$$



Series & Parallel Resistance

• Series Resistance

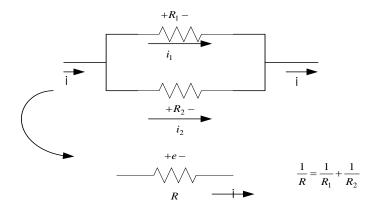


$$e_1 = iR_1, \quad e_2 = iR_2$$

$$e =$$

Series/Parallel Capacitance?

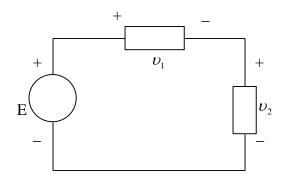
• Parallel Resistance



$$-e_1 + e_2 = 0 \implies$$

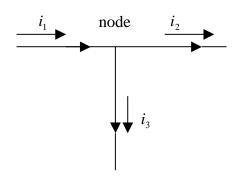
Kirchhoff's laws

1. The algebraic sum of the potential difference around a closed path equals zero.



$$-\upsilon_1-\upsilon_2+E=0$$

2. The algebraic sum of the currents entering (or leaving) a nod is equal to zero.

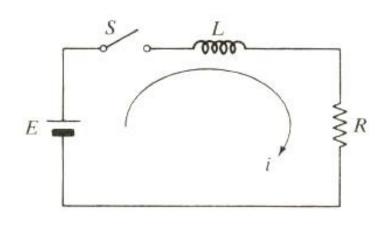


$$i_1 - i_2 - i_3 = 0$$

Current in = Current out

$$\rightarrow i_1 = i_2 + i_3$$

Mathematical Modeling of Electrical Systems



The switch S is closed at t=0

or

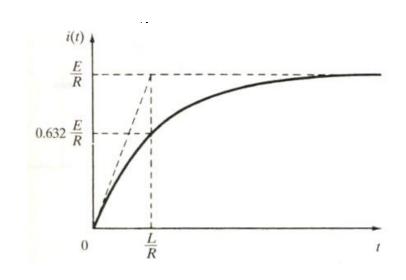
At the instant that switch S is closed, the current i(0) = 0

Laplace Transformation:

$$i(0) = 0$$
 $\rightarrow (Ls + R)I(s) = \frac{E}{s}$

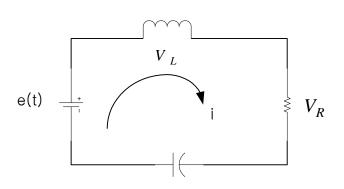
$$I(s) = =$$

$$\therefore i(t) =$$



Examples of Circuit Analysis

ex) R-L-C Circuit



$$V_o(s) =$$

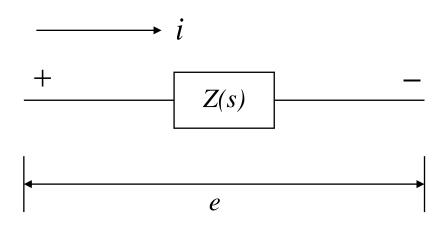
$$V_{o}(s) =$$

$$\begin{aligned} -V_L - V_R - V_C + e(t) &= 0 \\ V_L &= L \frac{di}{dt}, \quad V_R = iR, \quad V_c = \frac{1}{C} \int i \, dt + V_C(t) \\ \frac{dV_c}{dt} &= \frac{1}{C} i \\ L \frac{di}{dt} + Ri &= e(t) - V_C(t) \end{aligned}$$

Laplace Transform,
$$\frac{(Ls+R+\frac{1}{Cs})I(s)=E(s)}{\frac{I(s)}{E(s)}} = \frac{1}{(Ls+R+\frac{1}{Cs})}$$

Step Response?

Complex Impedance



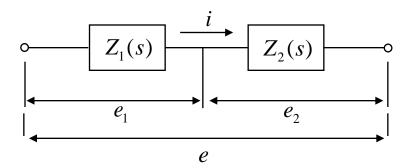
$$I(s) = \frac{E(s)}{Z(s)}$$

$$E(s) = Z(s)I(s)$$

Z(s): complex impedance

Complex Impedance

The complex impedance Z(s) of a two-terminal circuit is:



$$Z(s) = E(s) =$$

$$E_1(s) = Z_1(s)I(s),$$
 $E_2(s) = Z_2(s)I(s)$

Direct derivation of transfer function, without writing differential equations first.

$$E(s) = E_1(s) + E_2(s)$$

$$= Z_1(s)I(s) + Z_2(s)I(s)$$

$$= (Z_1(s) + Z_2(s))I(s)$$

Complex Impedance

$$\begin{array}{c}
 & \stackrel{I}{\longrightarrow} & \stackrel{R}{\longrightarrow} & \stackrel{R}{$$

$$e = Ri$$
, $E(s) = RI(s)$

$$Z(s) =$$

$$\frac{de}{dt} = \frac{1}{C}i$$

$$sE(s) = \frac{1}{C}I(s) \rightarrow E(s) = \frac{1}{Cs}I(s)$$

$$\therefore Z(s) = \cdot$$

Inductance :
$$\longrightarrow i + L -$$

$$e = L\frac{di}{dt}, \qquad E(s) = Ls\,I(s)$$

$$\therefore Z(s) =$$

Component	Voltage-current	Current-voltage	Voltage-charge	Impedance $Z(s) = V(s)/I(s)$	Admittance $Y(s) = I(s)/V(s)$
— (— Capacitor	$v(t) = \frac{1}{C} \int_0^1 i(\tau) d\tau$	$i(t) = C \frac{dv(t)}{dt}$	$v(t) = \frac{1}{C}q(t)$	$\frac{1}{Cs}$	Cs
-\\\\\- Resistor	v(t) = Ri(t)	$i(t) = \frac{1}{R}v(t)$	$v(t) = R \frac{dq(t)}{dt}$	R	$\frac{1}{R} = G$
——————————————————————————————————————	$v(t) = L \frac{di(t)}{dt}$	$i(t) = \frac{1}{L} \int_0^1 v(\tau) \ d\tau$	$v(t) = L \frac{d^2 q(t)}{dt^2}$	Ls	$\frac{1}{Ls}$

Component	Force-velocity	Force-displacement	Impedence $Z_M(s) = F(s)/X(s)$
Spring $x(t)$ $f(t)$ K	$f(t) = K \int_0^t v(\tau) d\tau$	f(t) = Kx(t)	K
Viscous damper $x(t)$ $f(t)$	$f(t) = f_{\nu}\nu(t)$	$f(t) = f_{\nu} \frac{dx(t)}{dt}$	$f_{ u}s$
Mass $x(t)$ $M \rightarrow f(t)$	$f(t) = M \frac{dv(t)}{dt}$	$f(t) = M \frac{d^2 x(t)}{dt^2}$	Ms ²

Examples of Complex Impedance

Series Impedances

ex1)
$$\stackrel{i}{\longrightarrow} {}_{+}^{R_{-}} {}_{+}^{L_{-}} {}_{+}^{C} {}_{-} \\
\stackrel{e_{R}}{\longrightarrow} {}_{e_{L}} {}_{e_{C}}$$

$$e = e_R + e_L + e_C$$

$$e_R = iR$$
, $e_L = L\frac{di}{dt}$, $\frac{de_C}{dt} = \frac{1}{C}i$

$$E(s) = E_R(s) + E_L(s) + E_C(s)$$

=

=

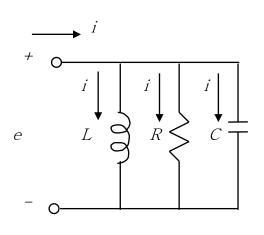
Z(s) E(s)

$$\therefore Z(s) = R + Ls + \frac{1}{Cs} = Z_R(s) + Z_L(s) + Z_C(s)$$

Examples of Complex Impedance

Parallel Impedances

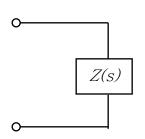




$$i = i_L + i_R + i_C$$
 $E(s) = Z(s)I(s)$

$$I(s) =$$

$$= \left(\right) E(s)$$

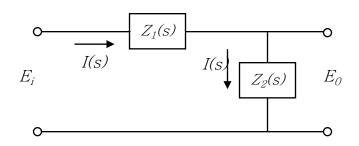


$$\therefore Z(s) = \frac{1}{\frac{1}{Z_R(s)} + \frac{1}{Z_L(s)} + \frac{1}{Z_C(s)}} = \frac{1}{\frac{1}{Ls} + \frac{1}{R} + Cs}$$

E(s)

Examples of Complex Impedance

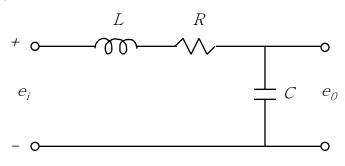
Deriving transfer functions of Electrical circuits by the use of complex impedances.



$$E_i(s) = Z_1(s)I(s) + Z_2(s)I(s), \qquad E_o(s) = Z_2(s)I(s)$$

$$\frac{E_o(s)}{E_i(s)} =$$

ex)



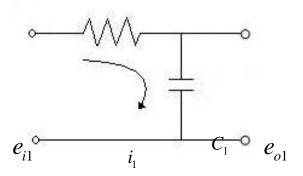
$$\frac{E_o(s)}{E_i(s)} =$$

$$Z_1(s) = \qquad \qquad Z_2(s) =$$

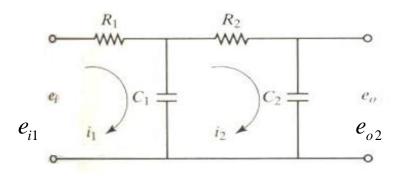
$$\frac{E_o(s)}{E_i(s)} =$$

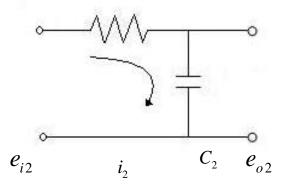
Transfer Functions of Cascaded Elements

Consider two RC circuits



$$\frac{E_{o1}(s)}{E_{i1}(s)} = \frac{1}{R_1 C_1 s + 1} = G_1(s)$$

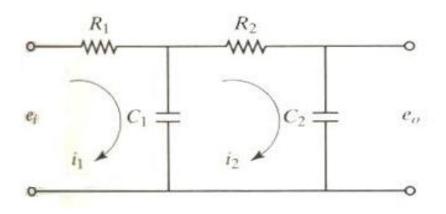




$$\frac{E_{o2}(s)}{E_{i2}(s)} = \frac{1}{R_2 C_2 s + 1} = G_2(s)$$

$$\frac{E_{o2}(s)}{E_{i1}(s)} = ?$$

Transfer Functions of Cascaded Elements Loading Effect

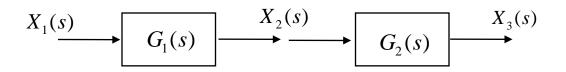


$$T.F : \frac{E_o(s)}{E_i(s)} = \frac{1}{(R_1C_1s+1)(R_2C_2s+1) + R_1C_2s} \neq \frac{1}{(R_1C_1s+1)(R_2C_2s+1)}$$

$$\therefore \frac{E_o(s)}{E_i(s)} \neq \frac{E_{o1}(s)}{E_{i1}(s)} \cdot \frac{E_{o2}(s)}{E_{i2}(s)} \longrightarrow \text{Loading effect}$$

Transfer Functions of Cascade Elements

Input Impedance, Output Impedance



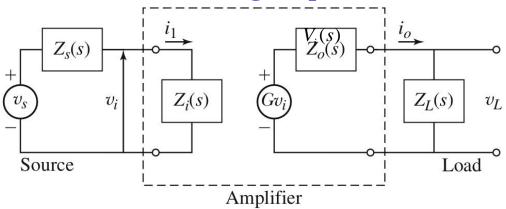
$$G(s) = \frac{X_3(s)}{X_1(s)} = \frac{X_2(s)}{X_1(s)} \cdot \frac{X_3(s)}{X_2(s)} = G_1(s)G_2(s)$$

If the "input Impedance" of the second element is infinite, the output of the first element is not affected by connecting it to the second element.

Then,
$$G(s) = G_1(s)G_2(s)$$

Transfer Functions of Cascade Elements

Isolating Amplifier



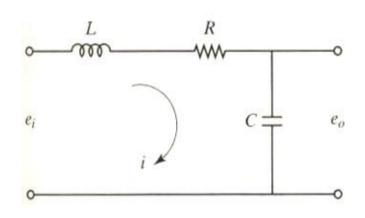
This amplifier circuit has to

- 1. Not affect the behavior of the source circuit.
- 2. Not be affected by the loading circuit.

This isolating amplifier circuit has to have

- 1. a very high input impedance,
- 2. very low output impedance

State-Space Mathematical Modeling of Electrical Systems



By Kirchhoff's voltage law

$$L\frac{di}{dt} + Ri + v_c = e_i$$
, $\frac{dv_c}{dt} = \frac{1}{C}i$, $e_o = v_c$

Assume, initial condition is 0,

$$LsI(s) + RI(s) + V_c(s) = E_i(s), \qquad V_c(s) = \frac{1}{Cs}I(s)$$

$$T.F : \frac{E_o(s)}{E_i(s)} = \frac{1}{LCs^2 + RCs + 1}$$

State-Space Mathematical Modeling of Electrical Systems

Differential equation :
$$\ddot{e}_o + \frac{R}{L}\dot{e}_o + \frac{1}{LC}e_o = \frac{1}{LC}e_i$$

State variable :
$$x_1 = e_o$$
, $x_2 = \dot{e}_o$

Input and output:
$$u = e_i$$
, $y = e_o = x_1$

State-space equation:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{1}{LC} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{LC} \end{bmatrix} u, \quad y = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$