

Mathematical Modeling of Dynamic Systems in State Space II

Transfer Function and State Equation

1. $\frac{z(s)}{u(s)} = \frac{3}{s+2}$ $z(s)(s+2) = 3u(s)$ $x_1 = z(t), \dot{x}_1 = \dot{z} = -2z(t)$
 $\dot{z}(t) + 2z(t) = 3u(t)$ $\begin{bmatrix} \dot{x}_1 = -2x_1 + 3u(t) \\ y = x_1 \end{bmatrix}$

2. $\frac{z(s)}{u(s)} = \frac{5s+3}{s+2} = \frac{5(s+2)-7}{s+2} = 5 - \frac{7}{s+2}$

$$z(s) = 5u(s) - \frac{7}{s+2}u(s)$$

$$\text{let } z_1(s) = \frac{7}{s+2}u(s), \dot{z}_1 + 2z_1 = 7u$$

$$\text{let } x_1 = z_1, \begin{bmatrix} \dot{x}_1 = -2x_1 + 7u \\ y = -z_1 + 5u \\ = -x_1 + 5u \end{bmatrix}$$

Another way,

$$\frac{z(s)}{u(s)} = \frac{5s+3}{s+2}$$

$$\frac{z(s)}{5s+3} = \frac{u(s)}{s+2} = Q(s)$$

$$\dot{q} + 2q = u(t)$$

$$z(t) = 5\dot{q} + 3q$$

$$= 5(-2q + u) + 3q$$

$$= 5u - 7q$$

$$\text{let } x_1 = q, \begin{bmatrix} \dot{x}_1 = -2x_1 + u \\ y = -7x_1 + 5u \end{bmatrix}$$

2nd order systems

State Equation and Transfer Function

ex1) Consider a system defined by $\ddot{y} + 6\dot{y} + 11y = 6u$

Choose state variables, $x_1 = y$ \rightarrow Phase variables (each subsequent state variable is defined to be the derivative of the previous state variable.)
 $x_2 = \dot{y}$
 $x_3 = \ddot{y}$

Then we obtain, $\dot{x}_1 = x_2$,
 \vdots
 $\dot{x}_3 = \ddot{y} = -6x_1 - 11x_2 - 6x_3 + 6u$

$$\therefore \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 6 \end{bmatrix} u, \quad y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

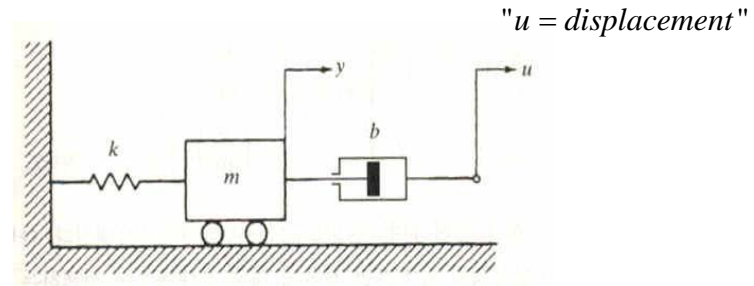
State Equation and Transfer Function

ex2) Consider a mechanical system,

$$m\ddot{y} = -ky - b(\dot{y} - \dot{u}), \quad \ddot{y} = -\frac{k}{m}y - \frac{b}{m}\dot{y} + \frac{b}{m}\dot{u}$$

Choose state variables, $x_1 = y, x_2 = \dot{y}$

$$\dot{x}_1 = x_2, \quad \dot{x}_2 = -\frac{k}{m}x_1 - \frac{b}{m}x_2 + \frac{b}{m}u \quad \dot{x} = Ax + Bu$$



The right side includes \dot{u} term. To explain the reason we should not include differentiation of u, assume $u = \delta(t)$ (unit impulse function)

$$x_2 = -\frac{k}{m} \int y dt - \frac{b}{m}y + \frac{k}{m}\delta(t)$$

x_2 includes $(k/m)\delta(t)$ term. It means $x_2(0) = \infty$ and cannot be accepted as a state variable.

That's why the standard form is $\dot{x} = Ax + Bu \quad y = Cx + Du$

State Equation and Transfer Function

Method 1: Choose a state variable that includes u

To eliminate \dot{u} term,

$$\ddot{y} = -\frac{k}{m}y - \frac{b}{m}\dot{y} + \frac{b}{m}\dot{u} \rightarrow \ddot{y} - \frac{b}{m}\dot{u} = -\frac{k}{m}y - \frac{b}{m}\dot{y}$$
$$\frac{d}{dt}\left(\dot{y} - \frac{b}{m}u\right) = -\frac{k}{m}y - \frac{b}{m}\left(\dot{y} - \frac{b}{m}u\right) - \left(\frac{b}{m}\right)^2 u$$

So we choose state variables as, $x_1 = y, x_2 = \dot{y} - \frac{b}{m}u$

$$\dot{x}_2 = \ddot{y} - \frac{b}{m}\dot{u} = \left(-\frac{k}{m}y - \frac{b}{m}\dot{y} + \frac{b}{m}\dot{u}\right) - \frac{b}{m}\dot{u} = -\frac{k}{m}x_1 - \frac{b}{m}\left(x_2 + \frac{b}{m}u\right)$$
$$= -\frac{k}{m}x_1 - \frac{b}{m}x_2 - \left(\frac{b}{m}\right)^2 u \rightarrow \dot{u} \text{ term has been eliminated.}$$

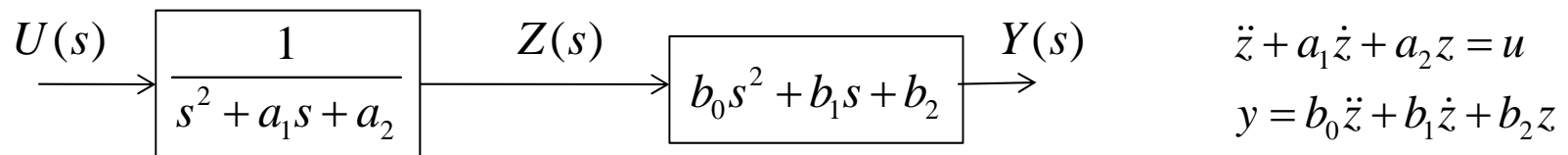
$$\therefore \begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{b}{m} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \frac{b}{m} \\ -\left(\frac{b}{m}\right)^2 \end{bmatrix} u$$

State Equation and Transfer Function

Method 2: Include the input derivatives in the output equation

Consider a second-order system, $\ddot{y} + a_1\dot{y} + a_2y = b_0\ddot{u} + b_1\dot{u} + b_2u$

$$\frac{Y(s)}{U(s)} = \frac{b_0s^2 + b_1s + b_2}{s^2 + a_1s + a_2} \rightarrow \frac{Z(s)}{U(s)} = \frac{1}{s^2 + a_1s + a_2}, \quad \frac{Y(s)}{Z(s)} = b_0s^2 + b_1s + b_2$$



let, $x_1 = z, \quad x_2 = \dot{z}$

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -a_2x_1 - a_1x_2 + u$$

$$b_0\ddot{z} + b_1\dot{z} + b_2z = b_0(-a_2x_1 - a_1x_2 + u) + b_1x_2 + b_2x_1 = y$$

$$\therefore \dot{x}_1 = x_2, \quad \dot{x}_2 = -a_2x_1 - a_1x_2 + u$$

$$y = (b_2 - a_2b_0)x_1 + (b_1 - a_1b_0)x_2 + b_0u$$

State Equation and Transfer Function

Method 2: Include the input derivatives in the output equation

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -a_2 & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u, \quad y = [b_2 - a_2 b_0 \quad \vdots \quad b_1 - a_1 b_0] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + b_0 u$$

N-th order differential equation,

$$y^{(n)} + a_1 y^{(n-1)} + \dots + a_{n-1} \dot{y} + a_n y = b_0 u^{(n)} + b_1 u^{(n-1)} + \dots + b_{n-1} \dot{u} + b_n u$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_{n-1} \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & 0 & \dots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & \dots & -a_1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_{n-1} \\ x_n \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix} u,$$

$$y = [b_n - a_n b_0 \quad \vdots \quad b_{n-1} - a_{n-1} b_0 \quad \vdots \quad \dots \quad \vdots \quad b_1 - a_1 b_0] \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + b_0 u$$

State Equation and Transfer Function

ex2) Consider this mechanical system again,

$$m\ddot{y} = -ky - b(\dot{y} - \dot{u}), \quad m\ddot{y} + b\dot{y} + ky = b\dot{u}$$

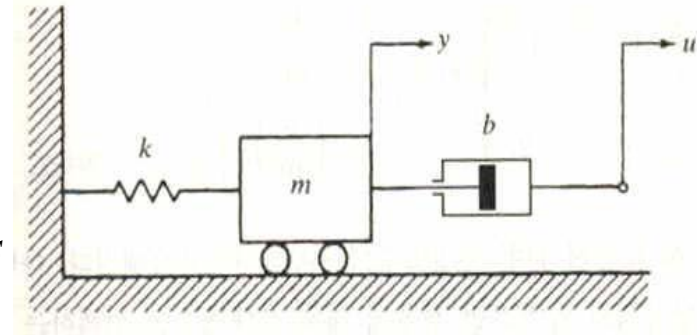
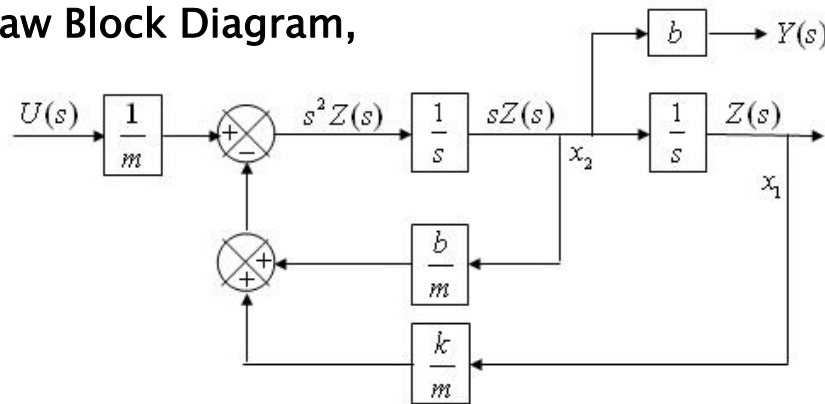
$$\frac{Y(s)}{U(s)} = \frac{bs}{ms^2 + bs + k}, \quad \frac{Z(s)}{U(s)} = \frac{1}{ms^2 + bs + k}, \quad \frac{Y(s)}{Z(s)} = bs$$

$$\frac{Y(s)}{cU(s)} = \frac{bs}{ms^2 + bs + k}, \quad \frac{Y(s)}{b(s)} = \frac{u(s)}{ms^2} = Q(s)$$

$$(ms^2 + bs + k)Z(s) = U(s), \quad bsZ(s) = Y(s)$$

$$s^2Z(s) = \frac{1}{m}U(s) - \frac{b}{m}sZ(s) - \frac{k}{m}Z(s)$$

Draw Block Diagram,



State variables,

$$x_1 = z, \quad x_2 = \dot{z}$$

$$\dot{x}_1 = x_2$$

$$\dot{x}_2 = -\frac{k}{m}x_1 - \frac{b}{m}x_2 + \frac{1}{m}u$$

$$y = bx_2$$

System Matrices:

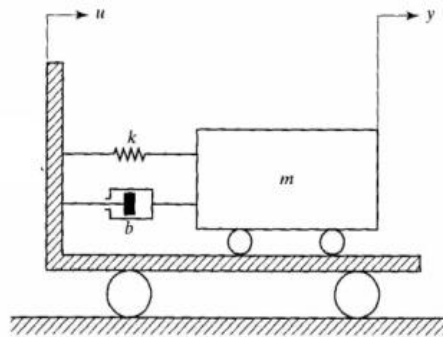
$$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{b}{m} \end{bmatrix} \quad \mathbf{B} = \begin{bmatrix} 0 \\ \frac{1}{m} \end{bmatrix}$$

$$\mathbf{C} = [0 \quad b]$$

$$\mathbf{D} = [0]$$

State Equation and Transfer Function

ex) Consider a spring-mass-damper system



m

$$m \frac{d^2 y}{dt^2} = -b \left(\frac{dy}{dt} - \frac{du}{dt} \right) - k(y - u)$$

$$\text{or } m \frac{d^2 y}{dt^2} + b \frac{dy}{dt} + ky = b \frac{du}{dt} + ku$$

$$\text{Transfer Function} = \frac{Y(s)}{U(s)} = \frac{bs + k}{ms^2 + bs + k}$$

State Eq:

Output Eq:

State variables:

Rewrite equations:

$$b_0 = 0, \quad b_1 = b, \quad b_2 = k \quad b_2 - a_2 b_0 = \frac{k}{m} - \frac{k}{m} + 0 = \frac{k}{m}$$

$$a_1 = \frac{b}{m}, \quad a_2 = \frac{k}{m} \quad b_1 - a_1 b_0 = \frac{b}{m} - \frac{b}{m} + 0 = \frac{b}{m}$$

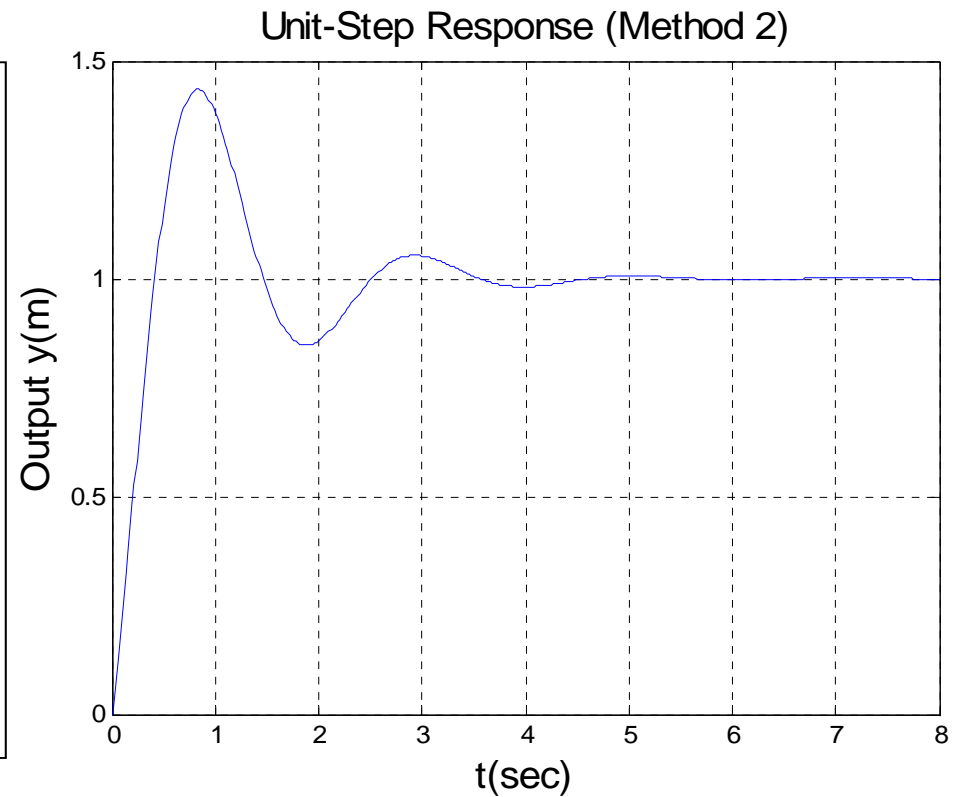
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -\frac{k}{m} & -\frac{b}{m} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u, \quad y = \begin{bmatrix} \frac{k}{m} & \frac{b}{m} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

Matlab Example

If, $m=10\text{kg}$, $b=20\text{N-s/m}$, $k=100\text{N/m}$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ -10 & -2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u, \quad y = \begin{bmatrix} 10 & 2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

```
t=0:0.02:8;
A=[0 1;-10 -2];
B=[0;1];
C=[10 2];
D=[0];
sys=ss(A,B,C,D);
[y,t]=step(sys,t);
plot(t,y)
grid
title('Unit-Step Response (Method 2)','FontSize',15)
xlabel('t(sec)','FontSize',15)
ylabel('Output y(m)','FontSize',15)
```



Transformation of Mathematical Models with MATLAB

$$\frac{Y(s)}{U(s)} = \frac{\text{numerator polynomial in } s}{\text{denominator polynomial in } s} = \frac{\text{num}}{\text{den}}$$

MATLAB command, `[A, B, C, D] = tf2ss(num,den)` gives a state space representation.

ex) Consider,
$$\frac{Y(s)}{U(s)} = \frac{s}{s^3 + 14s^2 + 56s + 160}$$

One of many possible state-space representations is,

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -14 & -56 & 160 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix} u$$

$$y = [0 \quad 1 \quad 0] \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + [0]u$$

```
>> num=[0 0 1 0];
>> den=[1 14 56 160];
>> [A,B,C,D]=tf2ss(num,den)
A =
   -14   -56  -160
    1     0     0
    0     1     0
B =
    1
    0
    0
C =
    0     1     0
D =
    0
```

Transformation of a State-Space Models into Another One

- $\dot{x} = Ax + Bu$
 $y = Cx + Du$ can be written as,

$$\begin{aligned} P\dot{\hat{x}} &= AP\hat{x} + Bu & \text{or} & & \dot{\hat{x}} &= P^{-1}AP\hat{x} + P^{-1}Bu \\ y &= CP\hat{x} + Du & & & y &= CP\hat{x} + Du \end{aligned}$$

– Since infinitely many $n \times n$ matrices can be a transformation matrix P , there are infinitely many state–space models for a given system.

- Eigenvalues of an $n \times n$ matrix A are the roots of the characteristic equation.

$$|\lambda I - A| = 0$$

The eigenvalues are also called the characteristic roots.

$$\begin{aligned} \text{ex) } A &= \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -6 & -11 & -6 \end{bmatrix}, & |\lambda I - A| &= \begin{vmatrix} \lambda & -1 & 0 \\ 0 & \lambda & -1 \\ 6 & 11 & \lambda + 6 \end{vmatrix} = \lambda^3 + 6\lambda^2 + 11\lambda + 6 \\ & & & & & = (\lambda + 1)(\lambda + 2)(\lambda + 3) = 0 \end{aligned}$$

Diagonalization of State Matrix A

Consider an $n \times n$ state matrix A :

$$A = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 \\ 0 & 0 & 1 & \cdots & 0 \\ \vdots & \vdots & \vdots & & \vdots \\ 0 & 0 & 0 & \cdots & 1 \\ -a_n & -a_{n-1} & -a_{n-2} & \cdots & -a_1 \end{bmatrix}$$

If matrix A has distinct eigenvalues and the state vector x is transformed into another state vector z by use of a transformation matrix P ,

$$x = Pz, \text{ where } P = \begin{bmatrix} 1 & 1 & \cdots & 1 \\ \lambda_1 & \lambda_2 & \cdots & \lambda_n \\ \lambda_1^2 & \lambda_2^2 & \cdots & \lambda_n^2 \\ \vdots & \vdots & & \vdots \\ \lambda_1^{n-1} & \lambda_2^{n-1} & \cdots & \lambda_n^{n-1} \end{bmatrix}$$

$$P^{-1}AP = \begin{bmatrix} \lambda_1 & & & 0 \\ & \lambda_2 & & \\ & & \ddots & \\ 0 & & & \lambda_n \end{bmatrix}$$

$P^{-1}AP$ is a canonical matrix and each column of P is an eigenvector of matrix A

Jordan Canonical Form

If matrix A involves multiple eigenvalues, diagonalization is not possible but matrix A can be transformed into a Jordan Canonical Form.

Consider the 3 x 3 matrix A :
$$A = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_3 & -a_2 & -a_1 \end{bmatrix}$$

Assume that matrix A has eigenvalues $\lambda_1, \lambda_2, \lambda_3$ where $\lambda_1 = \lambda_2 \neq \lambda_3$

$$(\lambda I - A)v_i = 0 \quad A: 3 \times 3 \text{ matrix}, \quad \lambda_1, \lambda_2, \lambda_3 \quad v_1, v_2, v_3$$

Case 1. $\text{rank}(\lambda_1 I - A) = 1$ can determine two eigenvectors v_1, v_2 .

$$Av_1 = \lambda_1 v_1, \quad Av_2 = \lambda_1 v_2, \quad Av_3 = \lambda_3 v_3$$

$$A \begin{bmatrix} v_1 & v_2 & v_3 \end{bmatrix} = \begin{bmatrix} v_1 & v_2 & v_3 \end{bmatrix} \begin{bmatrix} \lambda_1 & 0 & 0 \\ 0 & \lambda_1 & 0 \\ 0 & 0 & \lambda_3 \end{bmatrix}, \quad P^{-1}AP = \begin{bmatrix} \lambda_1 & 0 & 0 \\ 0 & \lambda_1 & 0 \\ 0 & 0 & \lambda_3 \end{bmatrix}$$

Jordan Canonical Form

Case 2. $\text{rank}(\lambda_1 I - A) = 2$ can determine one eigenvector v_1 .

$$A v_1 = \lambda_1 v_1, \quad A v_3 = \lambda_3 v_3$$

Find v_2 such that $(A - \lambda_1 I)v_2 = v_1$ $A v_2 = v_1 + \lambda_1 v_2$

$$A[v_1 \ v_2 \ v_3] = [v_1 \ v_2 \ v_3] \begin{bmatrix} \lambda_1 & 1 & 0 \\ 0 & \lambda_1 & 0 \\ 0 & 0 & \lambda_3 \end{bmatrix}$$

$$x = S z \quad \text{where} \quad S = \begin{bmatrix} 1 & 0 & 1 \\ \lambda_1 & 1 & \lambda_3 \\ \lambda_1^2 & 2\lambda_1 & \lambda_3^2 \end{bmatrix} \quad \text{will yield} \quad S^{-1} A S = \begin{bmatrix} \lambda_1 & 1 & 0 \\ 0 & \lambda_1 & 0 \\ 0 & 0 & \lambda_3 \end{bmatrix} = J$$

This is in the Jordan Canonical Form.

Canonical Forms

- **Diagonal (or Jordan) Canonical Form** (Partial Fraction Expansion)

Case 3. Complex Roots

$$AP = P\Lambda$$

Note : Complex Roots, Complex State x

$$\dot{x} = \Lambda x + bu$$

$$y = Cx$$

→ Complex case의 diagonalization 방법 이용

$$\Lambda K = KJ$$
$$\Lambda = \begin{bmatrix} \sigma + j\omega & 0 \\ 0 & \sigma - j\omega \end{bmatrix} \quad K = \begin{bmatrix} \frac{1}{2} & -\frac{j}{2} \\ \frac{1}{2} & \frac{j}{2} \end{bmatrix} \quad K^{-1} = \frac{2}{j} \begin{bmatrix} \frac{j}{2} & \frac{j}{2} \\ -\frac{1}{2} & \frac{1}{2} \end{bmatrix}$$

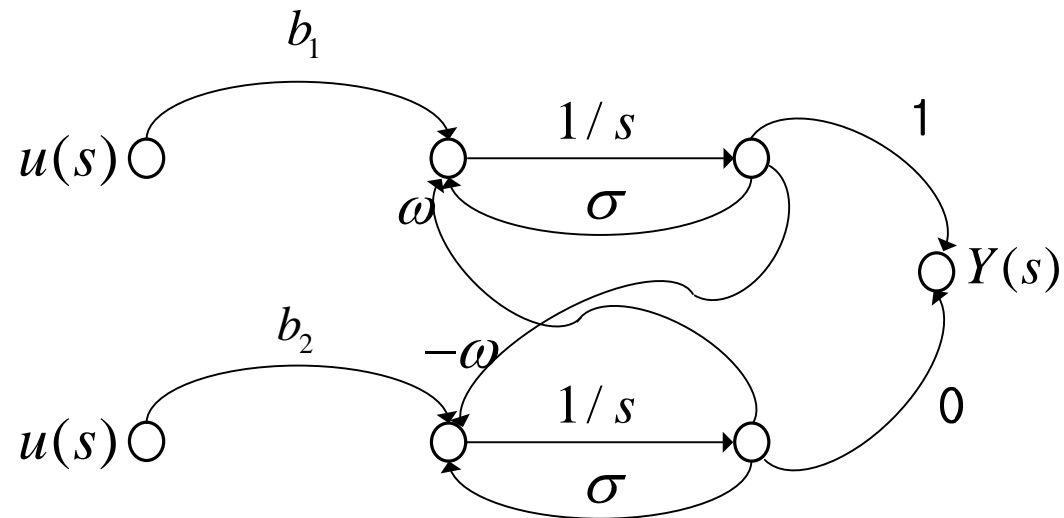
$$J = K^{-1}\Lambda K$$
$$= K^{-1}P^{-1}APK \quad J = \begin{bmatrix} \sigma & \omega \\ -\omega & \sigma \end{bmatrix}$$

Canonical Forms

- **Diagonal (or Jordan) Canonical Form** (Partial Fraction Expansion)

Case 3. Complex Roots

Ex)
$$\dot{z} = \begin{bmatrix} \sigma & \omega \\ -\omega & \sigma \end{bmatrix} z + \begin{bmatrix} b_1 \\ b_2 \end{bmatrix} u$$
$$y = [1 \quad 0] z$$



Canonical Forms

- **Diagonal (or Jordan) Canonical Form** (Partial Fraction Expansion)

Case 3. Complex Roots

Step 1 $\dot{x} = Ax + Bu$

Step 2 let $x = P\xi$
 $\dot{\xi} = \underbrace{P^{-1}AP}_{\Lambda} \xi + P^{-1}Bu$

: diagonal

Step 3 let $\xi = Kz$
 $\dot{z} = \underbrace{K^{-1}\Lambda K}_J z + K^{-1}P^{-1}Bu$
 $= \begin{bmatrix} \sigma & \omega \\ -\omega & \sigma \end{bmatrix} z + \begin{bmatrix} b_1 \\ b_2 \end{bmatrix} u$

$$\begin{cases} x = P\xi = PKz \\ \dot{z} = \underbrace{K^{-1}\Lambda K}_J z + K^{-1}P^{-1}bu \\ y = CPKz \end{cases}$$

End of Lecture 5(II)