

Estimation Theory Final Test_Solution

December 8, 2010

1. A simplified spacecraft tracking problem is formulated by

$$\begin{aligned} \dot{x}_c &= w_c, & w_c &\sim N(0, q) \\ z_c &= x_c + v_c, & v_c &\sim N(0, r). \end{aligned} \tag{T10.2.1-1}$$

(a) Suppose that the measurements are taken every 0.5 second. Show that the discrete model for Eq. (T10.2.1-1) is given by

$$\begin{aligned} x(k+1) &= x(k) + w(k) \\ z(k) &= x(k) + v(k). \end{aligned} \tag{T10.2.1-2}$$

Determine the mean and variance of $w(k)$ and $v(k)$.

(Solution) (10 points)

Given

$$\begin{aligned} \dot{x} &= w \equiv A_c x + G_c w, & A_c &= 0, G_c = 1, w \sim N(0, q) \\ z &= x + v, & v &\sim N(0, r) \end{aligned}$$

the discrete model is given by

$$x(k+1) = e^{A_c T} x(k) + w(k) = x(k) + w(k) \equiv Ax(k) + Gw(k) \tag{T2.1-1}$$

$$z(k) = x(k) + v(k) \equiv Hx(k) + v(k) \tag{T2.1-2}$$

where

$$\begin{aligned} w(k) &\sim N\left(0, \int_0^T e^{A_c \tau} G_c Q G_c^T (e^{A_c \tau})^T d\tau = Tq = 0.5q\right) \\ v(k) &\sim N\left(0, \frac{r}{T} = 2r\right). \end{aligned}$$

(b) Find the steady-state Kalman filter solution for this problem assuming that $w(k)$ and $v(k)$ are white Gaussian and uncorrelated each other.

(Solution) (20 points)

Find P_∞ by solving the ARE

$$\begin{aligned}
P_\infty &= A \left[P_\infty - P_\infty H^T (H P_\infty H^T + R)^{-1} H P_\infty \right] A^T + G Q G^T \\
&= \left[P_\infty - P_\infty (P_\infty + 2r)^{-1} P_\infty \right] + \frac{q}{2} \\
&= P_\infty - \frac{P_\infty^2}{P_\infty + 2r} + \frac{q}{2} \\
\frac{P_\infty^2}{P_\infty + 2r} &= \frac{q}{2} \\
P_\infty^2 - \frac{q}{2} P_\infty - rq &= 0 \\
P_\infty &= \frac{q}{4} \pm \sqrt{\left(\frac{q}{4}\right)^2 + rq}.
\end{aligned}$$

Since $P_\infty > 0$,

$$P_\infty = \frac{q}{4} + \sqrt{\left(\frac{q}{4}\right)^2 + rq} = \frac{1}{4} \left(q + \sqrt{q^2 + 16rq} \right). \quad (\text{T2.1-3})$$

And

$$\begin{aligned}
K_\infty &= P_\infty H (H P_\infty H^T + R)^{-1} \\
&= P_\infty (P_\infty + 2r)^{-1} \\
&= \frac{P_\infty}{P_\infty + 2r} \\
&= \frac{\frac{1}{4} \left(q + \sqrt{q^2 + 16rq} \right)}{\frac{1}{4} \left(q + \sqrt{q^2 + 16rq} \right) + 2r}.
\end{aligned} \quad (\text{T2.1-4})$$

The transfer function for the steady-state Kalman filter, $H_{SSKF}(z)$, is

$$H_{SSKF}(z) = H \left(zI - A [I - K_\infty H] \right)^{-1} A K_\infty = \frac{K_\infty}{z - 1 + K_\infty} \quad (\text{T2.1-5})$$

Let $q = r = 1$ for comparison,

$$H_{SSKF}(z) \approx \frac{0.39}{z - 0.61}. \quad (\text{T2.1-6})$$

(c) Find the causal Wiener filter solution for this problem and compare it with the result of (b).

(Solution) (20 points)

Suppose $w(k)$ and $v(k)$ are white, then

$$S_w(z) = 0.5q$$

From $x(k+1) = x(k) + w(k)$,

$$H_{wto x}(z) = \frac{S_x(z)}{S_w(z)} = \frac{1}{Z^{-1} - 1} \quad (\text{T2.1-7})$$

And

$$\begin{aligned} S_x(z) &= S_w(z)H_{wto x}(z)H_{wto x}(z^{-1}) \\ &= 0.5q \cdot \frac{1}{z^{-1} - 1} \cdot \frac{1}{z - 1} \\ &= \frac{0.5q}{(1 - z^{-1})(1 - z)} \end{aligned} \quad (\text{T2.1-8})$$

Let,

$$s(k) = x(k)$$

$$z(k) = s(k) + v(k)$$

Then,

$$S_s(z) = S_x(z)$$

$$S_z(z) = S_s(z) + S_v(z)$$

$$= \frac{0.5q}{(1 - z)(1 - z^{-1})} + 2r \quad (\text{T2.1-9})$$

$$S_{sz}(z) = S_s(z) = \frac{0.5q}{(1 - z)(1 - z^{-1})}$$

$$H_{Wiener}(z^{-1}) = \frac{1}{S_z^+(z)} \left[\frac{S_{sz}(z)}{S_z^{-1}(z)} \right]_+ \quad (\text{T2.1-10})$$

(Note: Refer to Eq.(T2.1-7))

To compare with (b), let $q=r=1$, then

$$\begin{aligned} S_z(z) &= \frac{0.5}{(1 - z)(1 - z^{-1})} + 2 \\ &= \frac{0.5 + 2(1 - z)(1 - z^{-1})}{(1 - z)(1 - z^{-1})} \\ &= \frac{4.5 - 2z - 2z^{-1}}{(1 - z)(1 - z^{-1})} \\ &= \frac{1.22(1.64 - z)(1.64 - z^{-1})}{(1 - z)(1 - z^{-1})} \\ &= \frac{1.10(1.64 - z^{-1})}{(1 - z^{-1})} \cdot \frac{1.10(1.64 - z)}{(1 - z)} \\ &\equiv S_z^+(z)S_z^-(z) \end{aligned}$$

$$\begin{aligned}
\frac{S_{SZ}(z)}{S_Z^{-1}(z)} &= \frac{0.5}{(1-z)(1-z^{-1})} \cdot \frac{(1-z)}{1.10(1.64-z)} \\
&= \frac{0.45}{(1-z^{-1})(1.64-z)} \\
&= \frac{0.70}{(1-z^{-1})} - \frac{0.70}{(1-1.64Z^{-1})} \\
\left[\frac{S_{SZ}(z)}{S_Z^{-1}(z)} \right]_+ &= \frac{0.70}{1-z^{-1}} \\
H_{Wiener}(z^{-1}) &= \frac{(1-z^{-1})}{1.10(1.64-z^{-1})} \cdot \frac{0.70}{(1-z^{-1})} \\
&= \frac{0.39}{1-0.61z^{-1}} \\
H_{Wiener}(z) &= \frac{0.39}{z-0.61} \quad (T2.1-11)
\end{aligned}$$

Comparing Eqs. (T2.1-6) and (T2.1-11), we see that $H_{SSKF}(z) = H_{Wiener}(z)$.

(d) Find the steady-state optimal smooth solution for this problem and compare it with the result of (b).

(For (b), (c), and (d), let $q = r = 1$.)

(Solution)

The forward filter is the same as (b), viz,

$$\begin{aligned}
P_\infty &= \frac{1}{4}(q + \sqrt{q^2 + 16rq}) \\
K_\infty &= \frac{P_\infty}{P_\infty + 2r} \quad (T2.1-12)
\end{aligned}$$

The backward filter is given by

$$\begin{aligned}
S_\infty &= A^T \left[I - S_\infty G \left[G^T S_\infty G + Q^{-1} \right]^{-1} G^T \right] S_\infty A + H^T R^{-1} H \\
&= \left[1 - S_\infty \left(S_\infty + \frac{1}{0.5q} \right)^{-1} \right] S_\infty + \frac{1}{2r} \quad (T2.1-13)
\end{aligned}$$

Solve Eq.(T2.1-13) for S_∞ to obtain

$$S_\infty = \frac{1}{4r} \left[1 + \sqrt{1 + \frac{16}{q}} \right] \quad (S_\infty > 0)$$

The steady-state optimal smoother is obtained by

$$\begin{aligned}
K_S &= P_\infty S_\infty \left[I + P_\infty S_\infty \right]^{-1} \\
P_S &= \left[I - K_S \right] P_\infty
\end{aligned}$$

For a simple comparison, let $q=r=1$. Then,

$$\begin{aligned}
P_\infty &= \frac{1}{4}(1 + \sqrt{17}) = 1.28 \\
S_\infty &= \frac{1}{4}(1 + \sqrt{17}) = 1.28 \\
K_s &= 1.28 \times 1.28 \times [1 + 1.28 \times 1.28]^{-1} \\
&= 0.62 \\
P_s &= (1 - 0.62) \times 1.28 = 0.49
\end{aligned}$$

We can see that the steady-state error variance is reduced to 0.49 from 1.28.

2. Consider the following state and measurement equations

$$\begin{aligned}
x_{k+1} &= \begin{bmatrix} 4 & 0 \\ 0 & 1 \end{bmatrix} x_k + w_k, \quad w_k \sim \left(0, \begin{bmatrix} 0 & 0 \\ 0 & 0.5 \end{bmatrix} \right), \quad x_0 \sim (0, P_0) \\
z_k &= Hx_k + v_k, \quad v_k \sim (0, 1).
\end{aligned}$$

(a) Determine if the steady-state Kalman gain K is asymptotically stable when

$$H = [0 \quad 3].$$

(Solution) (10 points)

$$\begin{aligned}
A &= \begin{bmatrix} 4 & 0 \\ 0 & 1 \end{bmatrix}, \quad G = I, \quad Q = \begin{bmatrix} 0 & 0 \\ 0 & 0.5 \end{bmatrix} = \sqrt{Q} \sqrt{Q}^T \\
\sqrt{Q} &= \begin{bmatrix} 0 & 0 \\ 0 & \sqrt{0.5} \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 0.707 \end{bmatrix}
\end{aligned}$$

Reachability test for $(A, G\sqrt{Q})$.

$$\rho \begin{bmatrix} 0 & 0 & 0 & 0 & \dots \\ 0 & 0.707 & 0 & 0.707 & \dots \end{bmatrix} = 1.$$

Therefore, $(A, G\sqrt{Q})$ is non-reachable.

Now, test detectability for (A, H) . When $H = [0 \quad 3]$,

$$A - LH = \begin{bmatrix} 4 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} l_1 \\ l_2 \end{bmatrix} \begin{bmatrix} 0 & 3 \end{bmatrix} = \begin{bmatrix} 4 & -3l_1 \\ 0 & 1 - 3l_2 \end{bmatrix}.$$

(A, H) is non-detectable since we can not move the eigenvalue 4.

According to Theorem 5-1 and 5-2 in the note, the steady-state error with the Kalman gain K is asymptotically unstable. In addition, it is not guaranteed that for every choice of P_0 there is a bounded limiting solution P .

(b) What if $H = [3 \ 3]$?

(Solution) (10 points)

When $H = [3 \ 3]$,

$$A - LH = \begin{bmatrix} 4 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} l_1 \\ l_2 \end{bmatrix} \begin{bmatrix} 3 & 3 \end{bmatrix} = \begin{bmatrix} 4 - 3l_1 & -3l_1 \\ -3l_2 & 1 - 3l_2 \end{bmatrix}.$$

Two eigenvalues of A may be arbitrarily located by properly choosing l_1 and l_2 . Therefore, (A, H) is detectable. The steady-state error with the Kalman gain K is asymptotically unstable. However, for every choice of P_0 there is a bounded limiting solution P .

3. The equation of motion of a pendulum hanging from a ceiling is given by the differential equation

$$\frac{d^2\theta(t)}{dt^2} + \sin\theta(t) = w(t),$$

where $\theta(t)$ is the angular position of the pendulum at time t . Discrete measurements of $\theta(t)$ are given by $z(n) = \theta(n) + v(n)$, where the sampling interval $T = 0.5$. Suppose that $w(t) \sim N(0, 0.02)$, $v(n) \sim N(0, 1)$, and $\theta(0) \sim N(0, 0.02)$. Give the equations for the EKF that provides an estimate of $\theta(i)$, $i = 1, 2, \dots$, based on the measurements $z(i)$ for $i = 1, 2, \dots, n$.

(Solution) (Kamen's Problem 8.8)

4. Suppose that RV x is uniformly distributed on $[-1, 1]$, and $y = e^{2x}$.

(a) What is the mean of y , \bar{y} ?

(Solution) (5 points)

$$\bar{y} = E\{y\} = E\{e^x\} = \int_{-1}^1 \frac{1}{2} e^{2x} dx = \frac{1}{4} (e^2 - e^{-2}) = 1.813$$

(b) What is the first-order approximation to \bar{y} ?

(Solution) (5 points)

1st order approximation

$$\bar{y} \simeq h(\bar{x}) + \left. \frac{\partial h}{\partial x} \right|_{x=\bar{x}} E\{\tilde{x}\} = 1.$$

(c) What is the second-order approximation to \bar{y} ?

(Solution) (5 points)

2nd order approximation

$$\begin{aligned} \bar{y} &= E\{h(x)\} \simeq h(\bar{x}) + E\left\{D_{\tilde{x}}h + \frac{1}{2!}D_{\tilde{x}}^2h\right\} \\ &= 1 + \left. \frac{1}{2} \frac{\partial^2 h}{\partial x^2} \right|_{x=\bar{x}} E\{\tilde{x}^2\} = 1 + \frac{1}{2} \cdot 4P_x = \frac{5}{3} \approx 1.666. \end{aligned}$$

(d) What is the unscented approximation to \bar{y} ?

(Solution) (5 points)

Since

$$\bar{x} = 0, \quad \sigma_x^2 = \frac{1}{3}$$

and

$$\tilde{x}^{(1)} = \frac{1}{\sqrt{3}}, \quad \tilde{x}^{(2)} = -\frac{1}{\sqrt{3}},$$

we obtain

$$\bar{y}_u = \frac{1}{2} \left(e^{\frac{2}{\sqrt{3}}} + e^{-\frac{2}{\sqrt{3}}} \right) = 1.744..$$

(e) What is the variance of y ? What is the unscented approximation to the variance of y ?

(Solution) (10 points)

$$\text{var}(y) = E\{y^2\} - \bar{y}^2 = E\{e^{4x}\} - (1.813)^2 = 6.8225 - 3.2885 = 3.534..$$

$$\text{var}(y_u) = \frac{1}{2} \sum_{i=1}^2 \left[h(\tilde{x}^{(i)}) - \bar{y}_u \right]^2 = \frac{1}{2} \left[\left(e^{\frac{2}{\sqrt{3}}} - 1.744 \right)^2 + \left(e^{-\frac{2}{\sqrt{3}}} - 1.744 \right)^2 \right] \approx 2.042.$$