7. Minimum Realizations and Coprime Fractions

- ✓ What is Minimal Realization ?
- Minimal Realization and Coprime
- Computing Coprime Fractions
- ✓ Balanced Realization
- Realizations from Markov Parameters
- ✓ Degree of Transfer Matrices
- ✓ Matrix Polynomial Fractions
- ✓ Minimal Realizations-Matrix Case

What is Minimal Realization ?

- > Realization problem ($IOD \rightarrow SVD$)
 - * To apply many design techniques & computational algorithms for dynamical equations.
 - * To simulate before the system is built.
 - * To establish the link between SVD & IOD.
- > Good realization among many realizations.
 - * least possible dimension
 - (minimal dimension)
 - * controllable & observable
 - * easy to analysis (simple form)
 - \Rightarrow minimal realization,

Controllable(controller) canonical form Observale(observer) canonical form Jordan-form **Minimal Realization and Coprime**

Definition: Degree of proper rational transfer function For a proper rational transfer function

$$g(s) = \frac{N(s)}{D(s)},$$

If N(s) and D(s) is coprime,

Degree of g(s) := Degree of D(s).

Question:

What is the degree of
$$\hat{g}(s) = \frac{s+1}{s^2+2s+1}$$
?

Minimal Realization and Coprime

Definition:

Let SISO state equation

 $\dot{x} = Ax + Bu$

y = Cx + Du

be realization of proper & coprime rational ftn g(s).

Then, the state equation is said to be **irreducible** if *f*

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det(sI - A) = k (denomination of g(s))
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dimA=deg g(s)

where k is a nozero constant.

The irreducible state equation is called **minimal realization** of g(s)

Realization of
$$g(s) = \frac{N(s)}{D(s)}$$

 $g(s) = e + \frac{\beta_1 s^{n-1} + \dots + \beta_n}{s^n + \alpha_1 s^{n-1} + \dots + \alpha_n}$
 $g(s) = \frac{\beta_1 s^{n-1} + \dots + \beta_n}{s^n + \alpha_1 s^{n-1} + \dots + \alpha_n} = \frac{N(s)}{D(s)} = \frac{y(s)}{u(s)}$
 $\Rightarrow D(s) y(s) = N(s)u(s)$
 $\Rightarrow y(s) = N(s)D^{-1}(s)u(s)$

Controllable canonical-form realization

Introduce new variable v(t) by $v(s) = D^{-1}(s)u(s)$ $\Rightarrow \frac{D(s)v(s) = u(s)}{y(s) = N(s)v(s)} \left(\frac{v(s)}{u(s)} = \frac{1}{D(s)}\right)$

Realization of Proper Rational Functions

$$\frac{v(s)}{u(s)} = \frac{1}{D(s)} = \frac{1}{s^n + \alpha_1 s^{n-1} + \dots + \alpha_n}$$
$$D(s)v(s) = u(s)$$

Define

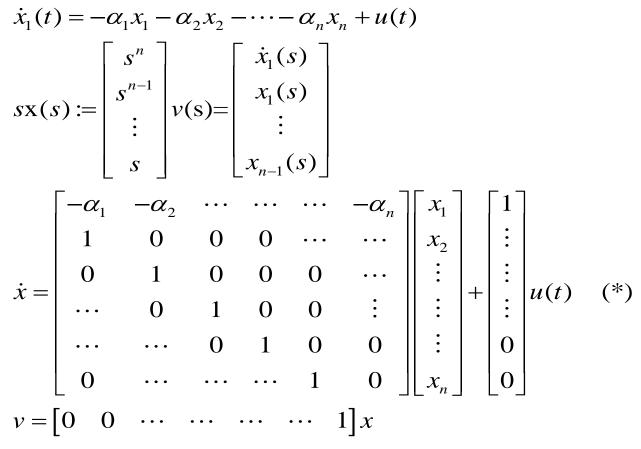
$$\mathbf{x}(s) := \begin{bmatrix} \mathbf{x}_{1}(s) \\ \vdots \\ \mathbf{x}_{n}(s) \end{bmatrix} := \begin{bmatrix} s^{n-1}v(s) \\ s^{n-2}v(s) \\ \vdots \\ v(s) \end{bmatrix} = \begin{bmatrix} s^{n-1} \\ s^{n-2} \\ \vdots \\ 1 \end{bmatrix} v(s),$$

$$\mathbf{x}_{n}(s) = -\alpha_{n}v(s) - \alpha_{n-1}sv(s) - \dots - \alpha_{1}s^{n-1}v(s) + u(s)$$

$$\mathbf{x}_{1}(s) = -\alpha_{n}\mathbf{x}_{n}(s) - \alpha_{n-1}\mathbf{x}_{n-1}(s) - \dots - \alpha_{1}\mathbf{x}_{1}(s) + u(s)$$

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In time domain



Linear Systems

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$$y(s) = N(s)v(s) = \sum_{i=1}^{n} \beta_{i} s^{n-i} v(s)$$

$$= \begin{bmatrix} \beta_{1} & \beta_{2} & \cdots & \beta_{n} \end{bmatrix} \begin{bmatrix} s^{n-1}v(s) \\ s^{n-2}v(s) \\ \vdots \\ v(s) \end{bmatrix}$$

$$= \begin{bmatrix} \beta_{1} & \beta_{2} & \cdots & \beta_{n} \end{bmatrix} \mathbf{x}(s)$$

$$\Rightarrow y(t) = \begin{bmatrix} \beta_{1} & \beta_{2} & \cdots & \beta_{n} \end{bmatrix} \mathbf{x}(t)$$

Controllability

 $\begin{bmatrix} s\mathbf{I} - A & B \end{bmatrix}$ $\begin{bmatrix} s + \alpha_1 & \alpha_2 & \dots & \dots & \alpha_n & \vdots & 1 \\ -1 & s & 0 & 0 & & \vdots & 0 \\ 0 & -1 & s & 0 & 0 & & \vdots & \vdots \\ & 0 & -1 & \dots & 0 & 0 & \vdots & \vdots \\ \dots & 0 & -1 & s & 0 & \vdots & 0 \\ 0 & 0 & \dots & 0 & -1 & s & \vdots & 0 \end{bmatrix} LI \text{ rows} \forall s$ has rank *n* regardless of $C = \begin{bmatrix} \beta_n & \cdots & \beta_1 \end{bmatrix}$ or N(s). Controllable realization from $\frac{N(s)}{D(s)}$ without coprimeness \Rightarrow Controllable Canonical form

Theorem 7.1

Controllable canonical form is observable iff D(s) and N(s) are coprime.

Pf.

 $(A \Rightarrow B) \Leftrightarrow (\neg B \Rightarrow \neg A)$ If D(s) and N(s) are not coprime, there exists a λ_1 such that $N(\lambda_1) = \beta_1 \lambda_1^3 + \beta_2 \lambda_1^2 + \beta_3 \lambda_1 + \beta_4 = 0$ $D(\lambda_1) = \lambda_1^4 + \alpha_1 \lambda_1^3 + \alpha_2 \lambda_1^2 + \alpha_3 \lambda_1 + \alpha_4 = 0.$

Pf. (cont)

Let us define v': = $[\lambda_1^3 \lambda_1^2 \lambda_1 1] \neq 0$, $N(\lambda_1) = cv = 0$ $\mathbf{A}\mathbf{v} = \begin{bmatrix} -\alpha_1 & -\alpha_2 & -\alpha_3 & -\alpha_4 \\ 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} \lambda_1^3 \\ \lambda_1^2 \\ \lambda_1 \\ 1 \end{bmatrix} = \begin{bmatrix} \lambda_1^4 \\ \lambda_1^3 \\ \lambda_1^2 \\ \lambda_1 \end{bmatrix} = \lambda_1 \mathbf{v}$ $A^2 v = AAv = \lambda_1 Av = \lambda_1^2 v, \dots$ $Ov = \begin{vmatrix} c \\ cA \\ cA^2 \\ cA^3 \end{vmatrix} v = \begin{vmatrix} cv \\ \lambda_1 cv \\ \lambda_1^2 cv \\ 2 \end{bmatrix} = 0$

This implies that O does not have full rank, i.e., not observable.

Pf. (cont)

$$(A \Leftarrow B) \Leftrightarrow (\sim A \Rightarrow \sim B)$$

If the state equation is not observable, then

by Theorem 6.O1, there exists λ_1 of A and $v \neq 0$ such that

$$\begin{bmatrix} \mathbf{A} - \lambda_1 \mathbf{I} \\ \mathbf{c} \end{bmatrix} \mathbf{v} = \mathbf{0}.$$

or

$$(A - \lambda_1 I)v = 0 \text{ and } cv = 0.$$

$$N(\lambda_1) = cv = \beta_1 \lambda_1^3 + \beta_2 \lambda_1^2 + \beta_3 \lambda_1 + \beta_4 = 0.$$

$$\lambda_1 \text{ is a root of } N(\lambda_1).$$

$$det(\lambda_1 I - A) = D(\lambda_1) = 0.$$
This implies $N(s)$ and $D(s)$ are not coprime.

Observable canonical form realization

$$y(s) = \frac{N(s)}{D(s)}u(s)$$
$$D(s)y(s) = N(s)u(s)$$

In Time Domain

$$y^{(n)}(t) + \alpha_1 y^{(n-1)}(t) + \dots + \alpha_n y(t)$$

= $\beta_1 u^{(n-1)}(t) + \dots + \beta_n u(t)$ (*)

Taking Laplace Transform with nonzero initial condition,

Taking Laplace Transform (non-zero initial condition)

$$s^{n} y(s) - (s^{n-1} y(0) + s^{n-2} y^{(1)}(0) + \dots + y^{(n-1)}(0) + \alpha_{1} \left\{ s^{n-1} y(s) - \left(s^{n-2} y(0) + \dots + y^{(n-2)}(0) \right) \right\} + \dots + \alpha_{n} y(s) = \beta_{1} \left\{ s^{n-1} u(s) - \left(s^{n-2} u(0) + \dots + u^{(n-2)}(0) \right) \right\} + \beta_{2} \left\{ \dots \right\} + \dots + \beta_{n} u(s)$$

$$D(s)y(s) = N(s)u(s) + \left\{ y(0)s^{n-1} + \left(y^{(1)}(0) + \alpha_1 y(0) - \beta_1 u(0) \right) s^{n-2} + \cdots \left(y^{(n-1)}(0) + \alpha_1 y^{(n-2)}(0) - \beta_1 u^{(n-2)}(0) + \alpha_2 y^{(n-3)}(0) - \beta_2 u^{(n-3)}(0) + \cdots + \alpha_{n-1} y(0) - \beta_{n-1} u(0) \right) \right\}$$

If initial state is known, output for a u(t) is unique.

We choose state as;

$$\begin{aligned} \mathbf{x}_{n}(t) &\coloneqq \mathbf{y}(t) \\ \mathbf{x}_{n-1}(t) &\coloneqq \mathbf{y}^{(1)}(t) + \alpha_{1} \mathbf{y}(t) - \beta_{1} u(t) \\ &\vdots & \vdots \\ \mathbf{x}_{1}(t) &\coloneqq \mathbf{y}^{(n-1)}(t) + \alpha_{1} \mathbf{y}^{(n-2)}(t) - \beta_{1} u^{(n-2)}(t) + \dots + \alpha_{n-1} \mathbf{y}(t) - \beta_{n-1} u(t) \quad (**) \end{aligned}$$

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$$\Rightarrow y = x_n$$

$$x_{n-1} = \dot{x}_n + \alpha_1 x_n - \beta_1 u \Rightarrow \dot{x}_n = x_{n-1} - \alpha_1 x_n + \beta_1 u$$

$$x_{n-2} = \dot{x}_{n-1} + \alpha_2 x_n - \beta_2 u \Rightarrow \dot{x}_{n-1} = x_{n-2} - \alpha_2 x_n + \beta_2 u$$

$$\vdots$$

$$x_1 = \dot{x}_2 + \alpha_{n-1} x_n - \beta_{n-1} u \Rightarrow \dot{x}_2 = x_1 - \alpha_{n-1} x_n + \beta_{n-1} u$$

$$(*) \& (**) \Rightarrow \dot{x}_1 = -\alpha_n x_n + \beta_n u$$

$$(*) \& (**) \Rightarrow \dot{x}_1 = -\alpha_n x_n + \beta_n u$$

$$y = \begin{bmatrix} 0 & \cdots & 0 & -\alpha_n \\ 1 & 0 & \cdots & 0 & -\alpha_n \\ 1 & 0 & \cdots & 0 & -\alpha_n \\ 1 & 0 & \cdots & 0 & -\alpha_n \end{bmatrix} x$$

Linear Systems

Note:

{*A*,*C*} does not depend on {
$$\beta_i$$
}, i.e. *N*(*s*)
 $\rho \begin{bmatrix} C \\ s\mathbf{I} - A \end{bmatrix} = n$

 $\Rightarrow \{A, C\} \text{ is always observable regardless of coprime}$ between N(s) & D(s) (may not be controllable if not coprime) $\Rightarrow \text{Observable}(\text{or observer})$ canonical form.

Coprime Fractions

$$g(s) = \frac{N(s)}{D(s)} = \frac{\overline{N}(s)R(s)}{\overline{D}(s)R(s)},$$

If $\overline{D}(s)$ and $\overline{N}(s)$ are coprime, controllable or observable realization of $\overline{g}(s) = \overline{N}(s) / \overline{D}(s)$ is minimal realization.

Theorem 7.2

{A, b, c, d} is a minimal realization of g(s) iff {A, b} is controllable and {A, c} is observable or iff dim A = deg g(s).

Pf. of Theorem 7.2

- (\Rightarrow)
 - If {A, b} is not controllable or {A, c} is not observable,
 the state equation can be reduced by Theorem 6.6 and 6.06.
 Thus {A, b, c, d} is not minimal.

Pf. of Theorem 7.2

(<=)

If {**A**, **b**, **c**, *d*} is controllable and observable, then $\rho(OC) = n$. However, if {**A**, **b**, **c**, *d*} is not minimal, there exists

a realization of g(s) { $\overline{\mathbf{A}}$, $\overline{\mathbf{b}}$, $\overline{\mathbf{c}}$, \overline{d} } with $n_1 < n$. By Theorem 4.1,

$$\mathbf{c}\mathbf{A}^{m}\mathbf{b}=\overline{\mathbf{c}}\overline{\mathbf{A}}^{m}\overline{\mathbf{b}}, \ m=0,1,2,\dots$$

$$OC = \begin{bmatrix} \mathbf{c} \\ \mathbf{cA} \\ \dots \\ \mathbf{cA}^{n-1} \end{bmatrix} \begin{bmatrix} \mathbf{b} & \mathbf{Ab} & \dots & \mathbf{A}^{n-1}\mathbf{b} \end{bmatrix} = \begin{bmatrix} \mathbf{cb} & \mathbf{cAb} & \dots & \mathbf{cA}^{n-1}\mathbf{b} \\ \mathbf{cAb} & & & & \\ \dots & & \dots & \dots \\ \mathbf{cA}^{n-1}\mathbf{b} & \dots & \mathbf{cA}^{2(n-1)}\mathbf{b} \end{bmatrix}$$
$$= \overline{O}_n \overline{C}_n \text{ has rank } n_1 < n.$$

This is contracts to that $\{A, b, c, d\}$ is controllable and observable.

Theorem 7.3

All minimal realization of g(s) are equivalent.

Pf.

Let {**A**, **b**, **c**, *d*} and { $\overline{\mathbf{A}}$, $\overline{\mathbf{b}}$, $\overline{\mathbf{c}}$, \overline{d} } are minimal, OC = $\overline{O}\overline{C}$ and OAC = $\overline{O}\overline{A}\overline{C}$ $\overline{A} = \overline{O}^{-1}OAC\overline{C}^{-1} = PAP^{-1}$, where P = $\overline{O}^{-1}O = \overline{C}C^{-1}(\leftarrow OC = \overline{O}\overline{C})$.

Note:

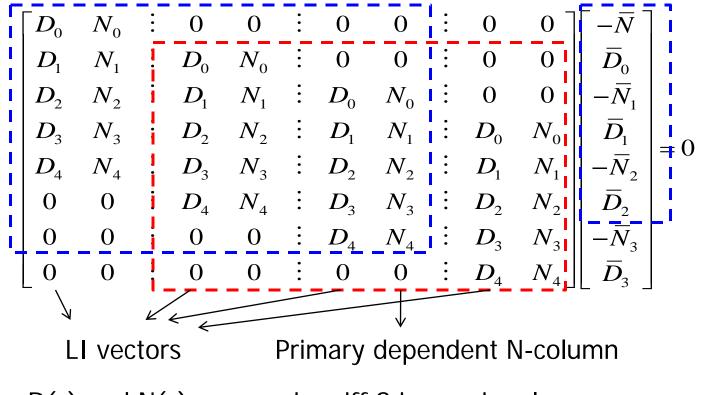
If $\{A, b, c, d\}$ is minimal (controllable and observable), Asymptotically stability \Leftrightarrow BIBO stability

Computing Coprime Fractions

 $g(s) = \frac{N(s)}{D(s)}$ $\frac{N(s)}{D(s)} = \frac{\overline{N}(s)}{\overline{D}(s)}$ If deg D(s) < deg D(s), D(s) and N(s) are not coprime. D(s)(-N(s)) + N(s)D(s) = 0 $D(s) = D_0 + D_1 s + D_2 s^2 + D_3 s^3 + D_4 s^4$ $N(s) = N_0 + N_1 s + N_2 s^2 + N_3 s^3 + N_4 s^4$ $\overline{D}(s) = \overline{D}_0 + \overline{D}_1 s + \overline{D}_2 s^2 + \overline{D}_3 s^3$ $\overline{N}(s) = \overline{N}_0 + \overline{N}_1 s + \overline{N}_2 s^2 + \overline{N}_3 s^3$

By Coefficient Comparison,

:= S (Sylvester resultant)

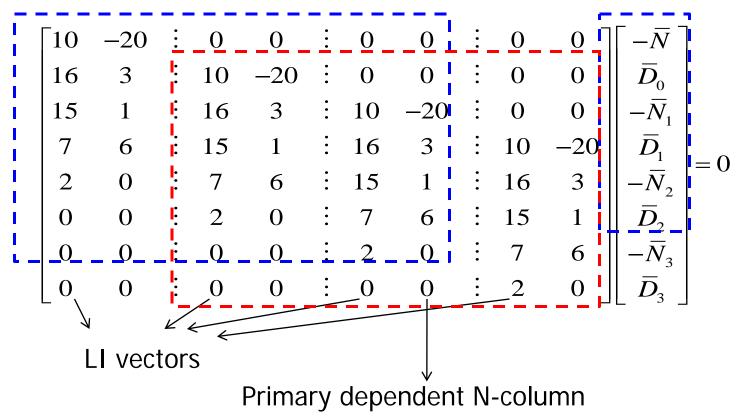


D(s) and N(s) are coprime iff S is nonsingular

Example 7.1

$$\frac{N(s)}{D(s)} = \frac{6s^3 + s^2 + 3s - 20}{2s^4 + 7s^3 + 15s^2 + 16s + 10}$$

Example 7.1 (cont)



Example 7.1 (cont)

This monic null vector equals

$$\begin{bmatrix} -\overline{N}_0 & \overline{D}_0 & -\overline{N}_1 & \overline{D}_1 & -\overline{N}_2 & \overline{D}_2 \end{bmatrix}'$$
$$= \begin{bmatrix} 4 & 2 & -3 & 2 & 0 & 1 \end{bmatrix}.$$

Thus we have $\overline{N}(s) = -4 + 3s + 0 \cdot s^2$ $\overline{D}(s) = 2 + 2s + s^2$ and

$$\frac{6s^3 + s^2 + 3s - 20}{2s^4 + 7s^3 + 15s^2 + 16s + 10} = \frac{3s - 4}{s^2 + 2s + 2}$$

Theorem 7.4

deg g(s) = number of linearly independent *N*-columns =: μ

The coefficients of a coprime fraction $g(s) = \overline{N}(s) / \overline{D}(s)$ is given by

$$\begin{bmatrix} -\overline{N}_0 & \overline{D}_0 & -\overline{N}_1 & \overline{D}_1 & \cdots & -\overline{N}_\mu & \overline{D}_\mu \end{bmatrix}'$$

QR Decomposition for column searching of S

Consider an $n \times m$ matrix M.

Then there exists an $n \times n$ orthonornal matrix \overline{Q} such that

$\overline{\mathbf{Q}}\mathbf{M} = \mathbf{R}$,

where R is an **upper** triangular matrix and

 $\rho \mathbf{M} = \rho \mathbf{R}$ with LI columns in order from left to right.

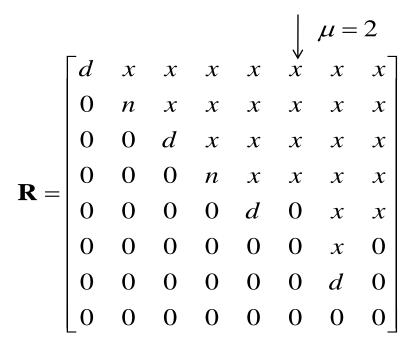
 $\mathbf{M} = \mathbf{Q}\mathbf{R}$, $\overline{\mathbf{Q}}^{-1} = \overline{\mathbf{Q}}' := \mathbf{Q} \leftarrow \mathbf{Q}\mathbf{R}$ decomposition.

Example 7.1

R =	-25.1	3.7	-20.6	10.1	-11.6	11.0	-4.1	5.3	
	0	-20.7	-10.3	4.3	-7.2	2.1	-3.6	6.7	
	0	0	-10.2	-15.6	-20.3	0.8	-16.8	9.6	
	0	0	0	8.9	-3.5	-17.9	-11.2	7.3	
	0	0	0	0	-5.0	0	-12.0	-15.0	
	0	0	0	0	0	0	-2.0	0	
	0	0	0	0	0	0	-4.6	0	
	0	0	0	0	0	0	0	0	
						Ĩ			
	Drimary dependent N. column								

Primary dependent N-column

Primary dependent N-column





Consider

$$g(s) = \frac{\beta_1 s + \beta_2}{s^2 + \alpha_1 s + \alpha_2} =: \frac{N(s)}{D(s)}$$

and its realization

$$\dot{\mathbf{x}} = \begin{bmatrix} -\alpha_1 & -\alpha_2 \\ 1 & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \quad y = \begin{bmatrix} \beta_1 & \beta_2 \end{bmatrix} \mathbf{x}$$

Show that the state equation is observable if and only if the Sylvester resultant of D(s) and N(s) is nonsingular.



Balanced Realization

 $\dot{\mathbf{x}} = \mathbf{A}\mathbf{x} + \mathbf{b}u$ $y = \mathbf{c}\mathbf{x}$

If the system is controllable, observable, and asymptotically stable, there exist $\mathbf{W}_c > 0$, $\mathbf{W}_o > 0$ such that

 $\mathbf{A}\mathbf{W}_{c}+\mathbf{W}_{c}\mathbf{A}^{\prime}=-\mathbf{b}\mathbf{b}^{\prime}$

and

 $\mathbf{A}\mathbf{W}_{o}+\mathbf{W}_{o}\mathbf{A}^{\prime}=-\mathbf{c}\mathbf{c}^{\prime}.$

Different Minimal Realization has different \mathbf{W}_{c} and \mathbf{W}_{o} .

$$\dot{\mathbf{x}} = \begin{bmatrix} -1 & -4/\alpha \\ 4\alpha & -2 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 \\ 2\alpha \end{bmatrix} u \quad \leftarrow g(s) = \frac{3s+18}{s^2+3s+18}$$
$$y = \begin{bmatrix} -1 & -2/\alpha \end{bmatrix} \mathbf{x}$$
$$\mathbf{W}_c = \begin{bmatrix} 0.5 & 0 \\ 0 & \alpha^2 \end{bmatrix} \text{ and } \mathbf{W}_o = \begin{bmatrix} 0.5 & 0 \\ 0 & 1/\alpha^2 \end{bmatrix}$$
$$\mathbf{W}_c \mathbf{W}_o = \begin{bmatrix} 0.25 & 0 \\ 0 & 1 \end{bmatrix}$$
$$\mathbf{W}_c = \mathbf{W}_o = \begin{bmatrix} 0.5 & 0 \\ 0 & 1 \end{bmatrix} \text{ for } \alpha = 1 \text{ \leftarrow Balanced Realization}$$

Balanced Realization

Theorem 7.5

Let {A, b, c} and { \overline{A} , \overline{b} , \overline{c} } be minimal and equivalent, let $W_c W_o$ and $\overline{W}_c \overline{W}_o$ be the product of their controllability and observability Grammians.

 $\mathbf{W}_{c}\mathbf{W}_{o}$ and $\overline{\mathbf{W}}_{c}\overline{\mathbf{W}}_{o}$ are similar and their eigenvalues are all real and positive.

Balanced Realization

Pf. of Theorem 7.5

$$\overline{\mathbf{A}} = \mathbf{P}\mathbf{A}\mathbf{P}^{-1} \quad \overline{\mathbf{b}} = \mathbf{P}\mathbf{b} \quad \overline{\mathbf{c}} = \mathbf{c}\mathbf{P}^{-1}$$

$$\overline{\mathbf{A}}\overline{\mathbf{W}}_{c} + \overline{\mathbf{W}}_{c}\overline{\mathbf{A}}' = -\overline{\mathbf{b}}\overline{\mathbf{b}}'$$
and
$$\overline{\mathbf{A}}'\overline{\mathbf{W}}_{o} + \overline{\mathbf{W}}_{o}\overline{\mathbf{A}} = -\overline{\mathbf{c}}'\overline{\mathbf{c}}$$

$$\mathbf{P}\mathbf{A}\mathbf{P}^{-1}\overline{\mathbf{W}}_{c} + \overline{\mathbf{W}}_{c}(\mathbf{P}')^{-1}\mathbf{A}'\mathbf{P}' = -\mathbf{P}\mathbf{b}\mathbf{b}'\mathbf{P}'$$
which implies
$$\mathbf{A}\mathbf{P}^{-1}\overline{\mathbf{W}}_{c}(\mathbf{P}')^{-1} + \mathbf{P}^{-1}\overline{\mathbf{W}}_{c}(\mathbf{P}')^{-1}\mathbf{A}' = -\mathbf{b}\mathbf{b}'$$

$$\mathbf{W}_{c} = \mathbf{P}^{-1}\overline{\mathbf{W}}_{c}(\mathbf{P}')^{-1} \text{ or } \overline{\mathbf{W}}_{c} = \mathbf{P}\mathbf{W}_{c}\mathbf{P}'$$

$$\mathbf{W}_{o} = \mathbf{P}'\overline{\mathbf{W}}_{o}\mathbf{P} \text{ or } \overline{\mathbf{W}}_{o} = (\mathbf{P}')^{-1}\mathbf{W}_{o}\mathbf{P}^{-1}$$

$$\mathbf{W}_{c}\mathbf{W}_{o} = \mathbf{P}^{-1}\overline{\mathbf{W}}_{c}(\mathbf{P}')^{-1}\mathbf{P}'\overline{\mathbf{W}}_{o}\mathbf{P} = \mathbf{P}^{-1}\overline{\mathbf{W}}_{c}\overline{\mathbf{W}}_{o}\mathbf{P} \rightarrow similar$$

Pf. of Theorem 7.5 (cont)

By Theorem 3.6, since \mathbf{W}_c is symmetric and positive definite,

 $\mathbf{W}_{c} = \mathbf{Q}'\mathbf{D}\mathbf{Q} = \mathbf{Q}'\mathbf{D}^{1/2}\mathbf{D}^{1/2}\mathbf{Q} := \mathbf{R}'\mathbf{R},$

where \mathbf{R} is not orthogonal but nonsingular.

 $det(\sigma^{2}\mathbf{I} - \mathbf{W}_{c}\mathbf{W}_{o}) = det(\sigma^{2}\mathbf{I} - \mathbf{R'RW}_{o}) = det(\sigma^{2}\mathbf{I} - \mathbf{RW}_{o}\mathbf{R'})$

This implies that $\mathbf{W}_{c}\mathbf{W}_{o}$ and $\mathbf{RW}_{o}\mathbf{R'}$ have the same eigenvalues. Since $\mathbf{RW}_{o}\mathbf{R'}$ is symmetric and positive definite, all eigenvalues are real and positive. (Q.E.D.)

Note:

 $\mathbf{W}_{\mathbf{c}}\mathbf{W}_{\mathbf{o}}$ of any minimal realization is similar Σ^2 , where $\Sigma = \operatorname{diag}(\sigma_1, \sigma_2, \dots, \sigma_n)$ and $\sigma_1 \ge \sigma_2 \ge \dots \ge \sigma_n > 0$

Balanced Realization

Theorem 7.6

For any *n*-dimensional minimal equation {A, b, c}, there exists an equivalence transformation $\overline{\mathbf{x}} = \mathbf{P}\mathbf{x}$ such that $\overline{\mathbf{W}}_c = \mathbf{\Sigma}\overline{\mathbf{W}}_o = \mathbf{X}$. This is called a *balanced realization*.

 $\mathbf{W}_{c}=\mathbf{R'R}$

 $\mathbf{R'W}_{o}\mathbf{R}$:real and symmetric

 $\rightarrow \mathbf{R'W}_{o}\mathbf{R} = \mathbf{U}\sum^{2}\mathbf{U'} \leftarrow \mathbf{U}: orthonormal$ $\mathbf{P}^{-1} = \mathbf{R'U} \quad {}^{-1/2}\mathbf{P} \text{ or} \mathbf{\Sigma} = \mathbf{U} \quad {}^{1/2}\mathbf{R} \quad {}^{\prime}(\quad {}^{\prime})^{-1}$ $\overline{\mathbf{W}}_{c} = \mathbf{P}\mathbf{W}_{c}\mathbf{P'} \mathbf{R} \quad {}^{1/2}\mathbf{W}_{c}\mathbf{R}\mathbf{W}\mathbf{U}^{\dagger} \quad \mathbf{R}_{c} \mathbf{R}^{-1} \quad {}^{1/2} = (\leftarrow {}_{c} = {}^{\prime})$ $\overline{\mathbf{W}}_{o} = (\mathbf{P'})\mathbf{\Sigma}\mathbf{W}_{o}\mathbf{P}^{-1}\mathbf{R}\mathbf{W} \quad \mathbf{R'}^{2}\mathbf{U}\mathbf{\Sigma} \quad {}_{o}\mathbf{\Sigma} \quad {}^{-1/2}\mathbf{R}\mathbf{W} \quad \mathbf{R'} \leftarrow (\mathbf{U}\mathbf{\Sigma} \quad \mathbf{U} \quad {}^{\prime} = {}^{2} \quad {}^{\prime})$

Balanced Realization

Note:

If $\overline{\mathbf{W}}_{c} = \mathbf{\Sigma}$, $\overline{\mathbf{W}}_{o} = {}^{2}$, it is called input-normal realization. If $\overline{\mathbf{W}}_{c} \mathbf{\Sigma} = {}^{2}\mathbf{W}^{-}\mathbf{J} = ,$ it is called output-normal realization. Balanced realization can be used in *system reduction*.

$$\begin{bmatrix} \dot{\mathbf{x}}_1 \\ \dot{\mathbf{x}}_2 \end{bmatrix} = \begin{bmatrix} \mathbf{A}_{11} & \mathbf{A}_{12} \\ \mathbf{A}_{21} & \mathbf{A}_{22} \end{bmatrix} \begin{bmatrix} \mathbf{x}_1 \\ \mathbf{x}_2 \end{bmatrix} + \begin{bmatrix} \mathbf{b}_1 \\ \mathbf{b}_2 \end{bmatrix} u$$
$$y = \begin{bmatrix} \mathbf{c}_1 & \mathbf{c}_2 \end{bmatrix} \mathbf{x}$$
$$\mathbf{W}_c = \mathbf{\Sigma} \mathbf{X}_o = \operatorname{diag}(1, 1)$$
$$\dot{\mathbf{x}}_1 = \mathbf{A}_{11} \mathbf{x}_1 + \mathbf{b}_1 u$$
$$y = \mathbf{c}_1 \mathbf{x}.$$

If Σ_2 is much smaller than Σ_1 ,

the reduced one is close to the original one.

Realization from the Hankel matrix

$$g(s) = \frac{\beta_0 s^n + \beta_1 s^{n-1} + \dots + \beta_n}{s^n + \alpha_1 s^{n-1} + \dots + \alpha_n}$$

= $h(0) + h(1)s^{-1} + \dots + h(n)s^{-n} + \dots$
(infinite series)
 $\{h(i), i = 0, 1, \dots\}$: Markov Parameters
 $h(0) = \beta_0$
 $h(1) = -\alpha_1 h(0) + \beta_1$
 $h(2) = -\alpha_1 h(1) - \alpha_2 h(0) + \beta_2$
:
 $h(n) = -\alpha_1 h(n-1) - \alpha_2 h(n-2) \dots - \alpha_n h(0) + \beta_n$
 $h(n+i) = -\alpha_1 h(n+i-1) - \alpha_2 h(n+i-2) \dots - \alpha_n h(i), i = 1, 2, \dots$

• •

Hankel matrix (m×n order)

$$\mathbf{T}(\mathbf{m},\mathbf{n}) := \begin{bmatrix} h(1) & h(2) & \cdots & h(n) \\ h(2) & h(3) & & \\ \vdots & & \\ h(m) & h(m+1) & \cdots & h(m+n-1) \end{bmatrix}$$

(!! h(0) is not involved)

Theorem

Proper transfer function g(s) has degree *n* iff $\rho \mathbf{T}(n,n) = \rho \mathbf{T}(n+k,n+l) = n \forall k,l=1,2,3$ Pf) (\Rightarrow) If deg g(s) = n $h(n+i) = \sum_{i=1}^{n} -\alpha_{j}h(n+i-j) \ (\beta_{k} \text{ are not involved})$ for i = 1, 2, ... \rightarrow (*n*+1)th row of **T**(*n*+1, ∞) can be written as a L.C. of *n* rows of $\mathbf{T}(n,\infty)$ $\rightarrow \rho \mathbf{T}(n,\infty) = \rho \mathbf{T}(n+1,\infty)$ $\rightarrow \rho \mathbf{T}(n+i,\infty) = \rho \mathbf{T}(n+i+1,\infty), \ i=1,2,\cdots$ $\rightarrow \rho \mathbf{T}(n,\infty) = \rho \mathbf{T}(\infty,\infty)$

$$\rightarrow \rho \mathbf{T}(n,\infty) = n(\because o/w \exists \overline{n} < n \text{ satisfying})$$

$$\rightarrow \rho \mathbf{T}(n,n) = \rho \mathbf{T}(n+k,n+l) = n \forall k,l$$

$$(\Leftarrow) \ \rho \mathbf{T}(n,n) = \rho \mathbf{T}(n+k,n+l) = n$$

$$\text{ implies } \exists \left\{\alpha_{j}\right\} \quad \Rightarrow$$

$$h(n+i) = \sum_{j=1}^{n} -\alpha_{j} \cdot h(n+i-j)$$

$$\text{ If we find } \left\{\beta_{i}\right\} \text{ using } (*)$$

$$g(s) = \sum_{i=0}^{\infty} h(i)s^{-i} = \frac{\beta_{0}s^{n} + \dots + \beta_{n}}{s^{n} + \alpha_{1}s^{n-1} + \dots + \alpha_{n}}$$

$$\rightarrow \deg g(s) = n$$

Row Searching Algorithm, Appendix A in 2nd Ed.

$$T = \begin{bmatrix} -1 & -1 & 2 & -2 & 1 \\ 1 & -1 & 4 & -2 & 4 \\ -1 & -3 & 8 & -6 & 6 \\ 5 & 1 & -4 & 10 & 1 \\ 7 & 1 & -2 & 10 & 4 \end{bmatrix}$$

$$k_{1}T = \begin{bmatrix} 1 & & & \\ -2 & 1 & & \\ -4 & 1 & & \\ 2 & & 1 & \\ 1 & & & 1 \end{bmatrix} T = \begin{bmatrix} -1 & -1 & 2 & -2 & 1 \\ 3 & 1 & 0 & 2 & 2 \\ 3 & -1 & 0 & 2 & 2 \\ 3 & 1 & 0 & 6 & 3 \\ 6 & 0 & 0 & 8 & 5 \end{bmatrix} := T_{1}$$

$$k_{2}T_{1} = \begin{bmatrix} 1 & 0 & & \\ 0 & 1 & & \\ & -1 & 1 & \\ & -2 & & 1 \end{bmatrix} T_{1} = \begin{bmatrix} -1 & -1 & 2 & -2 & 1 \\ 3 & 1 & 0 & 2 & 2 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & -2 & 0 & 4 & 1 \\ 0 & -2 & 0 & 4 & 1 \end{bmatrix} := T_{2}$$

$$k_{3} = I$$

$$k_{4}T_{2} = \begin{bmatrix} 1 & & & \\ & \ddots & & \\ & & 1 & \\ & & -1 & 1 \end{bmatrix} T_{2} = \begin{bmatrix} -1 & -1 & 2 & -2 & 1 \\ 3 & 1 & 0 & 2 & 2 \\ 0 & 0 & 0 & 0 & 0 \\ 0 & -2 & 0 & 4 & 1 \\ 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

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$$k := k_{4}k_{3}k_{2}k_{1}$$

$$kT = \begin{bmatrix} \overline{a}_{1} \\ \overline{a}_{2} \\ \vdots \\ \vdots \\ \overline{a}_{n} \end{bmatrix} \Rightarrow \text{If } i\text{-th row is}$$
dependent row, $\overline{a}_{i} = 0$

$$k = \begin{bmatrix} 1 & 0 & 0 & 0 \\ \vdots & \overline{a}_{n} \end{bmatrix} \Rightarrow \overline{a}_{i} = \sum_{j=i-1}^{j=i-1} k_{ij}a_{j} + a_{i} = 0$$

$$k = \begin{bmatrix} 1 & 0 & 0 & 0 \\ \vdots & \overline{a}_{i} & \ddots & 0 \\ \vdots & \ddots & \ddots & 1 \end{bmatrix} \Rightarrow a_{i} = -\sum_{j=i-1}^{j=i-1} k_{ij}a_{j}$$

$$\begin{bmatrix} k_{i1} & k_{i2} & \cdots & k_{i(i-1)} & 1 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} a_{1} \\ a_{2} \\ \vdots \\ a_{n} \end{bmatrix} = 0$$

Linear Systems

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Consider s.e.
$$\{A, B, C, D\}$$

 $g(s) = D + C(sI - A)^{-1}B = D + s^{-1}C(I - s^{-1}A)^{-1}B$
 $= D + CBs^{-1} + CABs^{-2} + CA^2Bs^{-3} \cdots$
 $\Rightarrow \{A, B, C, D\}$ is a realization of $g(s)$ iff
 $D = h(0) \& h(i) = CA^{i-1}B$ $i = 1, 2, \cdots$

Realization:

$$\{A, B, C, D\} \leftarrow h(i) = CA^{i-1}B \leftarrow g(s) = \sum h(i)s^{-i} \leftarrow g(s)$$

Here,
$$g(s) \xrightarrow{H(n,n)} \{A, B, C, D\}$$
 realization
Let $g(s) = \frac{N(s)}{D(s)}$, deg $D(s) = n$
(may not be coprime) (deg $g(s) \le n$)
Determine deg $g(s)$ (Hankel matrix Rank check)
 $\mathbf{T}(n+1,n) = \begin{bmatrix} h(1) & h(n) \\ \vdots & \vdots \\ h(n+1) & h(2n) \end{bmatrix} \Big\} \sigma LI$ rows
 $n+1-\sigma LD$ rows

where σ can be determined by row searching algorithm, $\begin{bmatrix} h(\sigma+1) & \dots & h(2\sigma) \end{bmatrix}$ is primary dependent row.

Note: If
$$D(s) \& N(s)$$
 are coprime, $\sigma = n$
otherwise $\sigma < n$.
 $\rightarrow [a_1 \quad a_2 \quad \cdots \quad a_{\sigma} \quad 1 \quad 0 \quad \cdots \quad 0] \mathbf{T}(n+1,n) = 0$
If $\sigma = n$, $h(n+1) = -\alpha_1 h(n) - \alpha_2 h(n-1) - \dots - \alpha_n h(1)$
 $\rightarrow a_i = \alpha_{n-i}, i = 1, \cdots, n$
If $\sigma < n$, $h(\sigma + 1) = -\sum_{i=1}^{\sigma} a_i h(i)$
 $\rightarrow a_i \neq \alpha_{n-i}, i = 1, \cdots, n$

Claim: $A = \begin{bmatrix} 0 & 1 & & & 0 \\ \vdots & 0 & 1 & & & \\ \vdots & \vdots & 0 & & & \\ \vdots & \vdots & \ddots & & & \\ 0 & 0 & & 0 & 1 \\ -a_1 & -a_2 & & & -a_{\sigma} \end{bmatrix}, \quad B = \begin{bmatrix} h(1) \\ h(2) \\ \vdots \\ \vdots \\ \vdots \\ h(\sigma) \end{bmatrix}$ $C = \begin{bmatrix} 1 & 0 & \cdots & \cdots & 0 \end{bmatrix}, \qquad D = h(0)$

is controllable & observable.

Since
$$h(\sigma + i) = -a_1 h(\sigma + i - 1) - a_2 h(\sigma + i - 2) \cdots - a_{\sigma} h(i)$$

 $i = 1, 2, 3 \cdots$
 $AB = \begin{bmatrix} h(2) \\ \vdots \\ h(\sigma + i) \end{bmatrix}, A^2 B = \begin{bmatrix} h(3) \\ h(4) \\ \vdots \\ h(\sigma + 2) \end{bmatrix} \cdots A^k B = \begin{bmatrix} h(k+1) \\ h(k+2) \\ \vdots \\ h(k+\sigma) \end{bmatrix}$
 $(C = \begin{bmatrix} 1 & 0 & \cdots \end{bmatrix})$
 $\Rightarrow CB = h(1), CAB = h(2), \cdots CA^2 B = h(3) \cdots$
 $\Rightarrow \{A, B, C, D\}$ is realization of $g(s)$

Controllability matrix

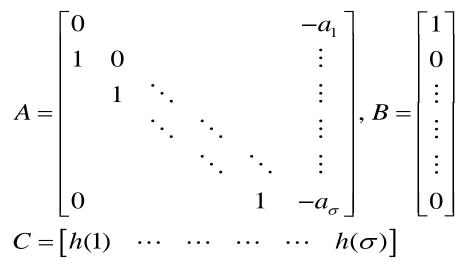
$$\begin{bmatrix} B & AB & \cdots & A^{\sigma-1}B \end{bmatrix} = \mathbf{T}(\sigma, \sigma) \Rightarrow \text{controllable}$$

Observability matrix

$$\begin{bmatrix} C \\ CA \\ \vdots \\ CA^{\sigma-1} \end{bmatrix} = \begin{bmatrix} 1 & & \\ & 1 & \\ & \ddots & \\ & & & 1 \end{bmatrix} \Rightarrow \text{observable}$$

If A is realized by $n > \sigma$, {A,B} is not controllable, but {A,C} is observable. \rightarrow Observability Realization $\mathbf{T}(\sigma, \sigma) = \mathbf{OC}$.

On the other hand,



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$$\tilde{\mathbf{T}}(\sigma,\sigma) \coloneqq \mathbf{OAC} = \begin{bmatrix} 1 & & & \\ & \dots & & \\ -a_{\sigma} & -a_{\sigma-1} & -a_{\sigma-2} \end{bmatrix} \begin{bmatrix} h(1) & \dots & h(\sigma) \\ \dots & \dots & & \\ h(\sigma) & \dots & h(2\sigma) \end{bmatrix}$$
$$= \begin{bmatrix} h(2) & \dots & h(\sigma+1) \\ \dots & \dots & & \\ h(\sigma+1) & \dots & h(2\sigma+1) \end{bmatrix}$$
$$\mathbf{AT}(\sigma,\sigma) = \tilde{\mathbf{T}}(\sigma,\sigma)$$
$$\mathbf{A} = \tilde{\mathbf{T}}(\sigma,\sigma)^{-1}\mathbf{T}(\sigma,\sigma)$$

If we know σ , we can determine **A**.

Example 7.2

$$g(s) = \frac{4s^2 - 2s - 6}{2s^4 + 2s^3 + 2s^2 + 3s + 1}$$

= 0s⁻¹ + 2s⁻² - 3s⁻³ - 2s⁻⁴ + 2s⁻⁵ + 3.5s⁻⁶ + ...
$$\mathbf{T}(4,4) = \begin{bmatrix} 0 & 2 & -3 & -2 \\ 2 & -3 & -2 & 2 \\ -3 & -2 & 2 & 3.5 \\ -2 & 2 & 3.5 & ... \end{bmatrix}, \quad \rho \mathbf{T}(4,4) = 3 = \sigma = \deg g(s)$$

$$\mathbf{A} = \tilde{\mathbf{T}}(3,3)\mathbf{T}^{-1}(3,3) = \begin{bmatrix} 2 & -3 & -2 \\ -3 & -2 & 2 \\ -2 & 2 & 3.5 \end{bmatrix} \begin{bmatrix} 0 & 2 & -3 \\ 2 & -3 & -2 \\ -3 & -2 & 2 \\ -3 & -2 & 2 \end{bmatrix}^{-1} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -0.5 & -1 & 0 \end{bmatrix}$$

$$b = \begin{bmatrix} 0 & 2 & -3 \end{bmatrix}', \ c = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}$$

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Example 7.2 (cont)

Without calculating $\tilde{\mathbf{T}}(3,3)\mathbf{T}^{-1}(3,3)$, by row searching algorothm

$$\begin{bmatrix} 1 & 0 & 0 & 0 \\ \dots & 1 & 0 & 0 \\ a_3 & a_2 & a_1 & 1 \end{bmatrix} \mathbf{T}(4,3) = \begin{bmatrix} c_1 \neq 0 \\ c_2 \neq 0 \\ c_3 \neq 0 \\ 0 \end{bmatrix}$$
$$\begin{bmatrix} a_3 & a_2 & a_1 & 1 \end{bmatrix} \mathbf{T}(4,3) = 0, \text{ by transpose,}$$
$$\mathbf{T}(3,4)\mathbf{a} = \begin{bmatrix} 0 & 2 & -3 & -2 \\ 2 & -3 & -2 & 2 \\ -3 & -2 & 2 & 3.5 \end{bmatrix} \begin{bmatrix} a_3 \\ a_2 \\ a_1 \\ 1 \end{bmatrix} = 0$$

a is null vector of T(3, 4).

Balanced Form

 $\mathbf{T}(\sigma, \sigma) = \mathbf{OC}.$ $\mathbf{T}(\sigma, \mathbf{\Delta E}) = \mathbf{K} \mathbf{\Lambda} \quad \mathbf{A} = \mathbf{L}^{1/2} \quad \mathbf{I}^{1/2} \quad \mathbf{I}^{1/2}$ $\mathbf{O} = \mathbf{K} \quad \mathbf{I}^{1/2} \text{ and } \mathbf{A} = \mathbf{L}^{1/2} \quad \mathbf{I}^{1/2}.$ $\mathbf{\tilde{T}}(\sigma, \sigma) = \mathbf{OAC} \rightarrow \mathbf{A} = \mathbf{O}^{-1} \mathbf{\tilde{T}}(\sigma, \sigma) \mathbf{C}^{-1}$ $\mathbf{AA} = \mathbf{K}^{2} \mathbf{T} \quad \mathbf{T} \quad (\sigma, \mathbf{E}) \mathbf{\Lambda} \quad \mathbf{I}^{-1/2}.$

CC'A= $\mathbf{L}^{1/2}\mathbf{L}\mathbf{\Lambda}$ $^{1/2}\mathbf{\Lambda}$ = O'OA= $\mathbf{K}^{1/2}\mathbf{K}\mathbf{\Lambda}$ $^{1/2}\mathbf{\Lambda}$ = \rightarrow Balanced Realization

Summary

Realizations of g(s) = $\frac{\beta_1 s^2 + \beta_2 s + \beta_3}{s^3 + \alpha_1 s^2 + \alpha_2 s + \alpha_3}$ (assume coprime)	
Controllable form	Observable form
$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 0 & 1 \\ -\alpha_3 & -\alpha_2 & -\alpha_1 \end{bmatrix}, \ \mathbf{B} = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$ $\mathbf{C} = \begin{bmatrix} \beta_3 & \beta_2 & \beta_1 \end{bmatrix}, D = h(0)$	$\mathbf{A} = \begin{bmatrix} 0 & 0 & -\alpha_3 \\ 1 & 0 & -\alpha_2 \\ 0 & 1 & -\alpha_1 \end{bmatrix}, \qquad \mathbf{B} = \begin{bmatrix} \beta_3 \\ \beta_2 \\ \beta_1 \end{bmatrix}$ $\mathbf{C} = \begin{bmatrix} 0 & 0 & 1 \end{bmatrix}, \qquad D = h(0)$
Controllability form	Observability form
$\mathbf{A} = \begin{bmatrix} 0 & 0 & -\alpha_3 \\ 1 & 0 & -\alpha_2 \\ 0 & 1 & -\alpha_1 \end{bmatrix}, \qquad \mathbf{B} = \begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}$ $\mathbf{C} = \begin{bmatrix} h(1) & h(2) & h(3) \end{bmatrix}, D = h(0)$	$\mathbf{A} = \begin{bmatrix} 0 & 1 \\ 0 & 1 \\ -\alpha_3 & -\alpha_2 & -\alpha_1 \end{bmatrix}, \mathbf{B} = \begin{bmatrix} h(1) \\ h(2) \\ h(3) \end{bmatrix}$ $\mathbf{C} = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix}, D = h(0)$

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Show that the two state equations

$$\dot{\mathbf{x}} = \begin{bmatrix} 2 & 1 \\ 0 & 1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 \\ 0 \end{bmatrix} u \quad y = \begin{bmatrix} 2 & 2 \end{bmatrix} \mathbf{x}$$

and

$$\dot{\mathbf{x}} = \begin{bmatrix} 2 & 0 \\ -1 & -1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 \\ 2 \end{bmatrix} u \quad y = \begin{bmatrix} 2 & 0 \end{bmatrix} \mathbf{x}$$

are realizations of $(2s+2)/(s^2 - s - 2)$. Are they minimal realization?

Are they algebraically equavalent?

Definition: MIMO case

Degree of a proper rational matrix $\hat{G}(s)$ is defined as the degree of Least Common Denominator (LCD) of all coprime minors of $\hat{G}(s)$.

Example

$$\hat{G}_{1}(s) = \begin{bmatrix} \frac{1}{s+1} & \frac{1}{s+1} \\ \frac{1}{s+1} & \frac{1}{s+1} \end{bmatrix}$$

The minors of order 1 : $\frac{1}{s+1}, \frac{1}{s+1}, \dots$ The minors of order 2 : 0 LCM of denominators $= s+1 = \Delta(s)$ $\Rightarrow \delta(\hat{G}_1) = 1$

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Example

$$\hat{G}_{2}(s) = \begin{bmatrix} \frac{2}{s+1} & \frac{1}{s+1} \\ \frac{1}{s+1} & \frac{1}{s+1} \end{bmatrix}$$

The minors of order 1 :
$$\frac{1}{(s+1)}$$

The minors of order 2 : $\frac{1}{(s+1)^2}$
 \Rightarrow LCM of denominators $= (s+1)^2$
 $\Rightarrow \delta(\hat{G}_2) = 2$

Example 7.5

$$\hat{\mathbf{G}}(s) = \begin{bmatrix} \frac{s}{s+1} & \frac{1}{(s+1)(s+2)} & \frac{1}{s+3} \\ \frac{-1}{s+1} & \frac{1}{(s+1)(s+2)} & \frac{1}{s} \end{bmatrix}$$

 1×1 minors : entries

$$2 \times 2 \text{ minors}: \left\{ \begin{array}{c} (s \neq 1) \\ \hline (s + 1)(s \pm 1)(s + 2) \\ \hline (s + 1)(s \pm 4) \\ \hline s(s \pm 4) \\ \hline s(s \pm 1)(s \pm 3) \\ \hline \hline s(s \pm 1)(s \pm 2)(s \pm 3) \end{array} \right\} \Leftarrow (\text{all should be coprime})$$

$$= \sum_{i=1}^{n} \delta(\hat{\mathbf{G}}) = 4$$

Minimal Realizations-Matrix Case

Theorem 7.M2

 $\{A, B, C, D\}$ is a minimal realization of G(s) iff

 $\{A,B\}$ is controllable and $\{A,C\}$ is observable or iff

 $\dim \mathbf{A} = \deg \mathbf{G}(s).$

Pf. of Theorem 7.M2

- (\Rightarrow) If not controllable or observable, There exists a zero state
 - equivalent equation with lesser dimension which is not minimal.
- (\Leftarrow) If not minimal, $\exists \{\overline{A}, \overline{B}, \overline{C}, \overline{D}\}\$ with $\overline{n} < n$, Theorem 4.1 implies

$$\mathbf{C}\mathbf{A}^{m}\mathbf{B} = \overline{\mathbf{C}}\overline{\mathbf{A}}^{m}\overline{\mathbf{B}}$$
 for $m = 0, 1, 2, ...$

$$OC = \overline{O}_n \overline{C}_n \qquad (*)$$

where $O, C, \overline{O}_n, \overline{C}_n$ are, respectly, $nq \times n, n \times np, nq \times \overline{n}$, and $\overline{n} \times np$. Using Sylvester inequility

$$\rho(\overline{O}_n) + \rho(\overline{C}_n) - \overline{n} \le \rho(\overline{O}_n\overline{C}_n) \le \min(\rho(\overline{O}_n), \rho(\overline{C}_n))$$

which is proved in [6], and $\rho(\overline{O}_n) = \rho(\overline{C}_n) = \overline{n}$, we have $\rho(\overline{O}_n\overline{C}_n) = \overline{n}$. From (*), $\rho(OC) = \rho(\overline{O}_n\overline{C}_n) = \overline{n} < n$.

This implies {A, B, C, D} is not controllable or observable.

The remaining part will be given in the remainder of this chapter.

Theorem 7.M3

All minimal realizations of G(s) are equivalent.

Pf.

Consider two minimal realizations {A, B, C, D} and { \overline{A} , \overline{B} , \overline{C} , \overline{D} }.

$$OC = \overline{OC}$$

$$\overline{O'OCCC'} = \overline{O'OCCC'}$$

$$(\overline{O'O})^{-1}\overline{O'O} = \overline{CC'}(C\overline{C'})^{-1} := P$$

$$OAC = \overline{OAC}$$

$$\overline{O'OACC'} = \overline{O'OACC'}$$

$$\overline{A} = (\overline{O'O})^{-1}\overline{O'OACC'}(\overline{CC'})^{-1} = PAP^{-1}$$

This shows {A, B, C, D} and {A, B, C, D} are equivalent.

Exmaple 7.6

$$\mathbf{G}(s) = \begin{bmatrix} \frac{4s - 10}{2s + 1} & \frac{3}{s + 2} \\ \frac{1}{(2s + 1)(s + 2)} & \frac{1}{(s + 2)^2} \end{bmatrix}, \Delta(s) = (2s + 1)(s + 2)^2$$

Minimal realization has 3-dimension.

6-dim. in (4.39) and 4-dim. in (4.44) are not minimal.

By Matlab, [am,bm,cm,dm]=minreal(a,b,c,d);

$$\dot{\mathbf{x}} = \begin{bmatrix} -0.8625 & -4.0897 & 3.2544 \\ 0.2921 & -3.0508 & 1.2709 \\ -0.0944 & 0.3377 & -0.5867 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 0.3218 & -0.5305 \\ 0.0459 & -0.4983 \\ -0.1688 & 0.0840 \end{bmatrix} \mathbf{u}$$
$$\mathbf{y} = \begin{bmatrix} 0 & -0.0339 & 35.5281 \\ 0 & -2.1031 & -0.5720 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 2 & 0 \\ 0 & 0 \end{bmatrix} \mathbf{u}$$

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Matrix Polynomial Fractions $\mathbf{G}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s)$

Example 7.5 can be expressed as a right fraction

$$\mathbf{G}(s) = \begin{bmatrix} s & 1 & s \\ -1 & 1 & s+3 \end{bmatrix} \begin{bmatrix} s+1 & 0 & 0 \\ 0 & (s+1)(s+2) & 0 \\ 0 & 0 & s(s+3) \end{bmatrix}^{-1}$$

$$\mathbf{G}(s) = \overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$$
 is called a *left fraction*.

$$\mathbf{G}(s) = \left[\mathbf{N}(s)\mathbf{R}(s)\right]\left[\mathbf{N}(s)\mathbf{R}(s)\right]^{-1}$$
$$= \mathbf{N}(s)\mathbf{R}(s)\mathbf{R}^{-1}(s)\mathbf{D}^{-1}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s)$$

→ Right (left) fraction is not unique. → Right (left) coprime fraction is needed. If $\mathbf{D}(s) = \hat{\mathbf{D}}(s)\mathbf{R}(s)$ and $\mathbf{N}(s) = \mathbf{N}(s)\mathbf{R}(s)$, $\mathbf{R}(s)$ is called *common right devider*. A(s) = B(s)C(s)
B: left devider of A
C: right devider of A
A: right multiple of B(s)
A: left multiple of C(s)

Definition 7.2 A square polynomial matrix M(s) is called a unimodular matrix if its determinant is nonzero and independent of s

Examples of unimodular matrix

$$\begin{bmatrix} 2s & s^2 + s + 1 \\ 2 & s + 1 \end{bmatrix}, \begin{bmatrix} -2 & s^{10} + s + 1 \\ 0 & 3 \end{bmatrix}, \begin{bmatrix} s & s + 1 \\ s - 1 & s \end{bmatrix}$$

Products of unimodular matrices are clearly unimodular.

$$\det \mathbf{M}_1(s) \det \mathbf{M}_2(s) = \det \left[\mathbf{M}_1(s) \mathbf{M}_2(s) \right] = c \neq 0$$

Inverse of unimodular matrix is unimodular.

$$\det \mathbf{M}(s) \det \mathbf{M}^{-1}(s) = \det \left[\mathbf{M}(s) \mathbf{M}^{-1}(s) \right] = \det \mathbf{I} = 1$$

Linear Systems

Definition 7.3 A square polynomial matrix R(s) is a *greatest common right divider (gcrd)* of D(s) and N(s) if
1) R(s) is *common right divider (crd)* of D(s) and N(s)
2) R(s) is left multiple of every crd of D(s) and N(s)
If a gcrd is a unimodular, D(s) and N(s) are right coprime.

Left coprime can be defined in a similar manner.

Greatest common right(left) devider M(*s*) is unimodular in $\mathbf{N}_r(s) = \overline{\mathbf{N}}_r(s)\mathbf{M}(s), \ \mathbf{D}_r(s) = \overline{\mathbf{D}}_r(s)\mathbf{M}(s)$ or $\mathbf{N}_l(s) = \mathbf{M}(s)\overline{\mathbf{N}}_l(s), \ \mathbf{D}_l(s) = \mathbf{M}(s)\overline{\mathbf{D}}_l(s),$ where det M(*s*) is independent of *s*.

Definition 7.4:

 $\mathbf{G}(s) = \mathbf{N}_{r}(s)\mathbf{D}_{r}^{-1}(s) = \mathbf{D}_{l}^{-1}(s)\mathbf{N}_{l}(s)$ right coprime $\Rightarrow \text{Characteristic polynomial} = \det \mathbf{D}_{r}(s) = \det \mathbf{D}_{l}(s)$ $\Rightarrow \deg \mathbf{G}(s) = \deg \det \mathbf{D}_{r}(s) = \deg \det \mathbf{D}_{l}(s)$

$$\mathbf{G}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s) = [\mathbf{N}(s)\mathbf{R}(s)][\mathbf{D}(s)\mathbf{R}(s)]^{-1}$$

Define $\mathbf{D}_1(s) = \mathbf{D}(s)\mathbf{R}(s)$, $\mathbf{N}_1(s) = \mathbf{N}(s)\mathbf{R}(s)$
det $\mathbf{D}_1(s) = \det[\mathbf{D}(s)\mathbf{R}(s)] = \det\mathbf{D}(s)\det\mathbf{R}(s)$
deg det $\mathbf{D}_1(s) = \deg\det\mathbf{D}(s) + \deg\det\mathbf{R}(s)$
If $\mathbf{R}(s)$ is unimodular, deg det $\mathbf{D}_1(s) = \deg\det\mathbf{D}(s)$
Then $\mathbf{D}_1(s)$ and $\mathbf{N}_1(s)$ are right coprime.

Column and Row Reducedness

Degree of polynimial vector: the highest power in all entries. $\delta_{ci} \mathbf{M}(s) = \text{degree of ith column of } \mathbf{M}(s) : column degree$ $\delta_{ri} \mathbf{M}(s) = \text{degree of ith row of } \mathbf{M}(s) : row degree$ $\mathbf{M}(s) = \begin{bmatrix} s+1 & s^3-2s+5 & -1 \\ s-1 & s^2 & 0 \end{bmatrix}$ $\rightarrow \delta_{c1} = 1, \ \delta_{c2} = 3, \ \delta_{c3} = 0, \ \delta_{r1} = 3, \ \delta_{r2} = 2$

Definition 7.5

A nonsingular polynomial matrix $\mathbf{M}(s)$ degree is column reduced if

deg det $\mathbf{M}(s) = sum of all column degrees.$

It is row reduced if

deg det $\mathbf{M}(s) = sum of all row degrees.$

Example:

$$\mathbf{M}(s) = \begin{bmatrix} 3s^2 + 2s & 2s + 1 \\ s^2 + s - 3 & s \end{bmatrix}$$

$$\Delta(s) = s^3 - s^2 + 5s + 3 \Rightarrow \deg \Delta(s) = \delta_{c1} + \delta_{c2} = 2 + 1$$

$$\rightarrow column \ reduced$$

$$\Rightarrow \deg \Delta(s) \neq \delta_{r1} + \delta_{r2} = 2 + 2$$

$$\rightarrow not \ row \ reduced$$

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 $\mathbf{M}(s)$ can be expressed as

$$\mathbf{M}(s) = \mathbf{M}_{hc}\mathbf{H}_{c}(s) + \mathbf{M}_{lc}(s)$$

Example :

$$\mathbf{M}(s) = \begin{bmatrix} 3 & 2 \\ 1 & 1 \end{bmatrix} \begin{bmatrix} s^2 & 0 \\ 0 & s \end{bmatrix} + \begin{bmatrix} 2s & 1 \\ s-3 & 0 \end{bmatrix}$$

 \searrow nonsingular \leftrightarrow column reduced

or can be expressed as

$$\mathbf{M}(s) = \mathbf{H}_{r}(s)\mathbf{M}_{hr} + \mathbf{M}_{lr}(s)$$

Example :

$$\mathbf{M}(s) = \begin{bmatrix} s^2 & 0 \\ 0 & s^2 \end{bmatrix} \begin{bmatrix} 3 & 0 \\ 1 & 0 \end{bmatrix} + \begin{bmatrix} 2s & 2s+1 \\ s-3 & s \end{bmatrix}$$

 \searrow singular \leftrightarrow not row reduced

$$\mathbf{G}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s) = \overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$$

right coprime left coprime $\mathbf{D}(s)$: column reduced, $\overline{\mathbf{D}}(s)$: row reduced

 $\deg \mathbf{G}(s) = \operatorname{sum} \operatorname{of} \operatorname{column} \operatorname{degrees} \operatorname{of} \mathbf{D}(s)$

= sum of row degrees of $\overline{\mathbf{D}}(s)$

If G(s) is strictly proper, then

 $\delta_{ci} \mathbf{N}(s) < \delta_{ci} \mathbf{D}(s), \ i = 1, 2, \dots$

The converse is not necessarily true, ex,

$$\mathbf{N}(s)\mathbf{D}^{-1}(s) = \begin{bmatrix} 1 & 2 \end{bmatrix} \begin{bmatrix} s^2 & s-1 \\ s+1 & 1 \end{bmatrix}^{-1} = \begin{bmatrix} \frac{-2s-1}{1} & \frac{2s^2-s+1}{1} \end{bmatrix}$$

The reason is that $\mathbf{D}(s)$ is not column reduced.

Theorem 7.8

If $\mathbf{D}(s)$ is column reduced, then

 $N(s)D^{-1}(s)$ is proper (strictly proper) iff

 $\delta_{ci} \mathbf{N}(s) \le \delta_{ci} \mathbf{D}(s) \quad \left[\delta_{ci} \mathbf{N}(s) < \delta_{ci} \mathbf{D}(s) \right] \text{ for } i = 1, 2, 3, \dots$

Pf.

Necessity part follows from the preceding examples.

To show sufficiency,

$$\mathbf{D}(s) = \mathbf{D}_{hc} \mathbf{H}_{c}(s) + \mathbf{D}_{lc}(s) = \left[\mathbf{D}_{hc} + \mathbf{D}_{lc}(s)\mathbf{H}_{c}^{-1}(s)\right]\mathbf{H}_{c}(s)$$

$$\mathbf{N}(s) = \mathbf{N}_{hc} \mathbf{H}_{c}(s) + \mathbf{N}_{lc}(s) = \left[\mathbf{N}_{hc} + \mathbf{N}_{lc}(s)\mathbf{H}_{c}^{-1}(s)\right]\mathbf{H}_{c}(s)$$

$$\mathbf{G}(s) \coloneqq \mathbf{N}(s)\mathbf{D}^{-1}(s) = \left[\mathbf{N}_{hc} + \mathbf{N}_{lc}(s)\mathbf{H}_{c}^{-1}(s)\right]\left[\mathbf{D}_{hc} + \mathbf{D}_{lc}(s)\mathbf{H}_{c}^{-1}(s)\right]^{-1}$$

$$\lim_{x \to \infty} \mathbf{G}(s) = \mathbf{N}_{hc}\mathbf{D}_{hc}^{-1}$$

$$\mathbf{D}_{hc}^{-1} \text{ is nonsingular since column reduced}$$

$$\Rightarrow proper$$

$$\mathbf{N}_{hc} = 0 \text{ for } \delta_{ci}\mathbf{N}(s) < \delta_{ci}\mathbf{D}(s)$$

Corollary 7.8

If $\mathbf{D}(s)$ is row reduced, then $\overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$ is proper (strictly proper) iff $\delta_{ri}\overline{\mathbf{N}}(s) \le \delta_{ri}\overline{\mathbf{D}}(s) \quad \left[\delta_{ri}\overline{\mathbf{N}}(s) < \delta_{ri}\overline{\mathbf{D}}(s)\right]$ for i = 1, 2, 3, ... HW 7-3

Find the characteristic polynomials and degrees of the following proper rational matrix of

$$\mathbf{G}(s) = \begin{bmatrix} \frac{1}{(s+1)^2} & \frac{s+3}{s+2} & \frac{1}{s+5} \\ \frac{1}{(s+3)^2} & \frac{s+1}{s+4} & \frac{1}{s} \end{bmatrix}$$

Use two methods: minors and column degrees.

You may use Matlab for coprime fraction.

Computing Matrix Coprime Fractions

For the given left fraction $\overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$,

not necessarily left coprime,

we can find the right coprime fraction $N(s)D^{-1}(s)$

$$\mathbf{G}(s) = \overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s)$$
$$\overline{\mathbf{N}}(s)\mathbf{D}(s) = \overline{\mathbf{D}}(s)\mathbf{N}(s)$$
$$\overline{\mathbf{D}}(s)(-\mathbf{N}(s)) + \overline{\mathbf{N}}(s)\mathbf{D}(s) = \mathbf{0}$$

where

$$\overline{\mathbf{D}}(s) = \overline{\mathbf{D}}_0 + \overline{\mathbf{D}}_1 s + \overline{\mathbf{D}}_2 s^2 + \overline{\mathbf{D}}_3 s^3 + \overline{\mathbf{D}}_4 s^4$$

$$\overline{\mathbf{N}}(s) = \overline{\mathbf{N}}_0 + \overline{\mathbf{N}}_1 s + \overline{\mathbf{N}}_2 s^2 + \overline{\mathbf{N}}_3 s^3 + \overline{\mathbf{N}}_4 s^4$$

$$\mathbf{D}(s) = \mathbf{D}_0 + \mathbf{D}_1 s + \mathbf{D}_2 s^2 + \mathbf{D}_3 s^3$$

$$\mathbf{N}(s) = \mathbf{N}_0 + \mathbf{N}_1 s + \mathbf{N}_2 s^2 + \mathbf{N}_3 s^3$$

$$\mathbf{SM} := \begin{bmatrix} \mathbf{\bar{D}}_{0} & \mathbf{\bar{N}}_{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} \\ \mathbf{\bar{D}}_{1} & \mathbf{\bar{N}}_{1} & \vdots & \mathbf{\bar{D}}_{0} & \mathbf{\bar{N}}_{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} \\ \mathbf{\bar{D}}_{2} & \mathbf{\bar{N}}_{2} & \vdots & \mathbf{\bar{D}}_{1} & \mathbf{\bar{N}}_{1} & \vdots & \mathbf{\bar{D}}_{0} & \mathbf{\bar{N}}_{0} & \vdots & \mathbf{0} & \mathbf{0} \\ \mathbf{\bar{D}}_{3} & \mathbf{\bar{N}}_{3} & \vdots & \mathbf{\bar{D}}_{2} & \mathbf{\bar{N}}_{2} & \vdots & \mathbf{\bar{D}}_{1} & \mathbf{\bar{N}}_{1} & \vdots & \mathbf{\bar{N}}_{0} & \mathbf{\bar{N}}_{0} \\ \mathbf{\bar{D}}_{4} & \mathbf{\bar{N}}_{4} & \vdots & \mathbf{\bar{D}}_{3} & \mathbf{\bar{N}}_{3} & \vdots & \mathbf{\bar{D}}_{2} & \mathbf{\bar{N}}_{2} & \vdots & \mathbf{\bar{N}}_{1} & \mathbf{\bar{N}}_{1} \\ \mathbf{0} & \mathbf{0} & \vdots & \mathbf{\bar{D}}_{4} & \mathbf{\bar{N}}_{4} & \vdots & \mathbf{\bar{D}}_{3} & \mathbf{\bar{N}}_{3} & \vdots & \mathbf{\bar{N}}_{2} & \mathbf{\bar{N}}_{2} \\ \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{\bar{D}}_{4} & \mathbf{\bar{N}}_{4} & \vdots & \mathbf{\bar{N}}_{3} & \mathbf{\bar{N}}_{3} \\ \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{0} & \mathbf{0} & \vdots & \mathbf{\bar{N}}_{4} & \mathbf{\bar{N}}_{4} \\ \end{bmatrix} \begin{bmatrix} -\mathbf{N}_{0} \\ \mathbf{D}_{0} \\ -\mathbf{N}_{1} \\ \mathbf{D}_{1} \\ -\mathbf{N}_{2} \\ \mathbf{D}_{2} \\ -\mathbf{N}_{3} \\ \mathbf{D}_{3} \end{bmatrix} = \mathbf{0}$$

S: Generalized resultant:
$$8q \times 4(q+p)$$

 $\overline{\mathbf{D}}_i : q \times q, \ \overline{\mathbf{N}}_i : q \times p, \ \mathbf{D}_i : p \times p, \ \mathbf{N}_i : q \times p$

Example 7.7

$$\mathbf{G}(s) = \begin{bmatrix} \frac{4s - 10}{2s + 1} & \frac{3}{s + 2} \\ \frac{1}{(2s + 1)(s + 2)} & \frac{s + 1}{(s + 2)^2} \end{bmatrix}$$

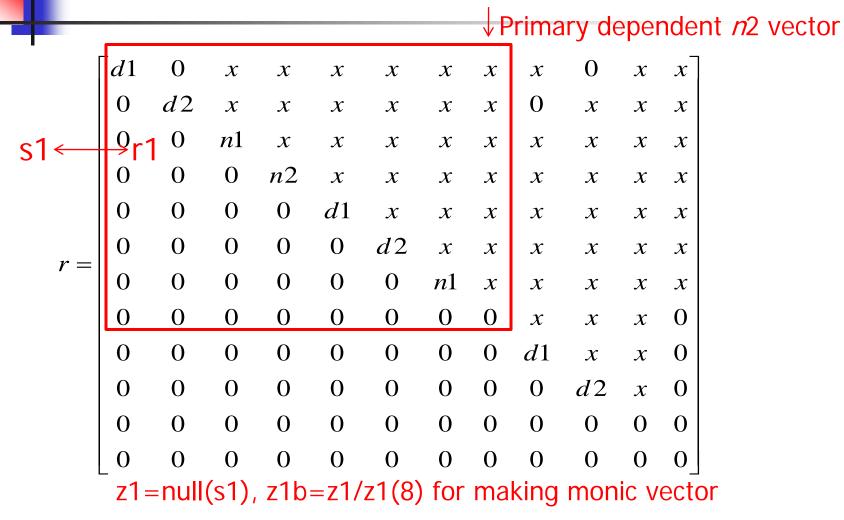
$$\mathbf{G}(s) = \begin{bmatrix} (2s+1)(s+2) & 0\\ 0 & (2s+1)(s+2)^2 \end{bmatrix}^{-1} \\ \times \begin{bmatrix} (4s-10)(s+2) & 3(2s+1)\\ s+2 & (s+1)(2s+1) \end{bmatrix} =: \overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$$

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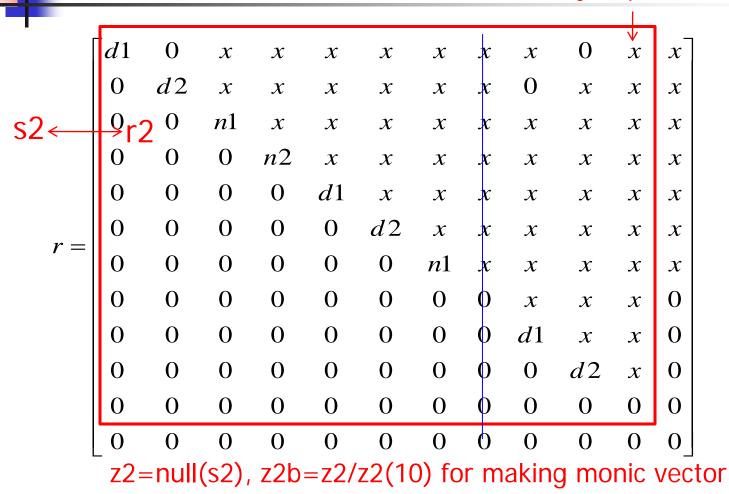
$$\overline{\mathbf{D}}(s) = \begin{bmatrix} 2s^2 + 5s + 2 & 0 \\ 0 & 2s^3 + 9s^2 + 12s + 4 \end{bmatrix}$$
$$= \begin{bmatrix} 2 & 0 \\ 0 & 4 \end{bmatrix} + \begin{bmatrix} 5 & 0 \\ 0 & 12 \end{bmatrix} s + \begin{bmatrix} 2 & 0 \\ 0 & 9 \end{bmatrix} s^2 + \begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix} s^3$$
$$\overline{\mathbf{N}}(s) = \begin{bmatrix} 4s^2 - 2s - 20 & 6s + 3 \\ s + 2 & 2s^2 + 3s + 1 \end{bmatrix}$$
$$= \begin{bmatrix} -20 & 3 \\ 2 & 1 \end{bmatrix} + \begin{bmatrix} -2 & 6 \\ 1 & 3 \end{bmatrix} s + \begin{bmatrix} 4 & 0 \\ 0 & 2 \end{bmatrix} s^2 + \begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix} s^3$$

By Matlab, $[q, r] = qr(S) \rightarrow S = qr$



Matrix Polynomial Fractions

Primary dependent *n*1 vector



$$\begin{bmatrix} -\mathbf{N}_{0} \\ \cdots \\ -\mathbf{D}_{0}^{21} & -n_{0}^{22} \\ d_{0}^{11} & d_{0}^{12} \\ d_{0}^{21} & d_{0}^{22} \\ -n_{1}^{11} & -n_{1}^{12} \\ -n_{1}^{21} & -n_{1}^{22} \\ d_{1}^{11} & d_{1}^{12} \\ d_{2}^{21} & d_{2}^{22} \\ -n_{2}^{21} & -n_{2}^{22} \\ -n_{2}^{21} & -n_{2}^{22} \\ d_{1}^{21} & d_{1}^{22} \\ d_{2}^{21} & d_{2}^{22} \\ \end{bmatrix} = \begin{bmatrix} 10 & 7 \\ -0.5 & -1 \\ 1 & 1 \\ 0 & 2 \\ 1 & -4 \\ 0 & 0 \\ 2.5 \\ 0 & 0 \\ 2.5 \\ 0 & 1 \\ -2 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 0 \\ \end{bmatrix}$$

$$z2b = z2/z2(10)$$

$$\mathbf{D}(s) = \begin{bmatrix} 1 & 1 \\ 0 & 2 \end{bmatrix} + \begin{bmatrix} 2.5 & 2 \\ 0 & 1 \end{bmatrix} s + \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} s^{2}$$

$$= \begin{bmatrix} s^{2} + 2.5s + 1 & 2s + 1 \\ 0 & s + 2 \end{bmatrix}$$

$$\mathbf{N}(s) = \begin{bmatrix} -10 & -7 \\ 0.5 & 1 \end{bmatrix} + \begin{bmatrix} -1 & 4 \\ 0 & 0 \end{bmatrix} s + \begin{bmatrix} 2 & 0 \\ 0 & 0 \end{bmatrix} s^{2}$$

$$= \begin{bmatrix} 2s^{2} - s - 10 & 4s - 7 \\ 0.5 & 1 \end{bmatrix}$$

$$\mathbf{G}(s) = \begin{bmatrix} (2s - 5)(s + 2) & 4s - 7 \\ 0.5 & 1 \end{bmatrix} \begin{bmatrix} (s + 2)(s + 0.5) & 2s + 1 \\ 0 & s + 2 \end{bmatrix}^{-1}$$

$$column \ degrees \ \mu_{1} = 2, \ \mu_{2} = 1, \ deg \ det \ \mathbf{D}(s) = 2 + 1 = 3$$

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Note:

$$\deg \mathbf{G}(s) = \deg \det \mathbf{D}(s) = \sum \mu_i$$

= total number of linearly independent \overline{N} -columns in S

$$\mathbf{P} = \begin{bmatrix} 0 & 0 & 1 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix}$$

 $\mathbf{G}(s) = \widehat{\mathbf{N}}(s)\mathbf{D}^{-1}(s) = [\mathbf{N}(s)\mathbf{P}][\mathbf{D}(s)\mathbf{P}]^{-1} = \mathbf{N}(s)\mathbf{D}^{-1}(s)$ The columns of $\mathbf{N}(s)\mathbf{D}(s)$ can be arbitrarily permutated.

Theorem 7.M4

Let $G(s) = \overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$ be a left fraction, not necessarily left coprime Let $\mu_i, i = 1, 2, ..., p$, be the number of linearly independent $\overline{N}_i - columns$. $\deg \mathbf{G}(s) = \mu_1 + \mu_2 + \dots + \mu_p$

A right coprime fraction $\mathbf{N}(s)\mathbf{D}^{-1}(s)$ can be obtained by computing p monic null vectors using p matrices formed from each primary dependent $\overline{N}_i - column$ and its LHS LI columns. Note:

The column-degree coefficient matrix \mathbf{D}_{hc}

can be a unit upper trangular matrix.

$$\mathbf{D}_{hc} = \begin{bmatrix} 1 & 2 \\ 0 & 1 \end{bmatrix} : column \ echelon \ form$$

 \rightarrow Realization will be nicer.

Dual:	Compute a left coprime fraction $\overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$								
	from a right fraction $N(s)D^{-1}(s)$								
	$\left[-\overline{\mathbf{N}}_{0}\overline{\mathbf{D}}_{0}\vdots-\overline{\mathbf{N}}_{1}\overline{\mathbf{D}}_{1}\vdots-\overline{\mathbf{N}}_{2}\overline{\mathbf{D}}_{2}\vdots-\overline{\mathbf{N}}_{3}\overline{\mathbf{D}}_{3}\right]\mathbf{T}=0$								
		$\begin{bmatrix} \mathbf{D}_0 \end{bmatrix}$	\mathbf{D}_1	\mathbf{D}_2	\mathbf{D}_3	\mathbf{D}_4	0	0	0]
		\mathbf{N}_{0}	\mathbf{N}_1	\mathbf{N}_2	\mathbf{N}_3	\mathbf{N}_4	0	0	0
		•••	•••	•••	•••	•••	•••	•••	•••
		0	\mathbf{D}_0	\mathbf{D}_1	\mathbf{D}_2	\mathbf{D}_3	\mathbf{D}_4	0	0
		0	\mathbf{N}_0	\mathbf{N}_1	\mathbf{N}_2	\mathbf{N}_3	\mathbf{N}_4	0	0
	$\mathbf{T} \coloneqq$	•••	•••	•••	•••	•••	•••	•••	•••
		0	0	\mathbf{D}_0	\mathbf{D}_1	\mathbf{D}_2	\mathbf{D}_3	\mathbf{D}_4	0
		0	0	\mathbf{N}_0	\mathbf{N}_1	\mathbf{N}_2	\mathbf{N}_3	\mathbf{N}_4	0
		•••	•••	•••	•••	•••	•••	•••	•••
		0	0	0	\mathbf{D}_0	\mathbf{D}_1	\mathbf{D}_2	\mathbf{D}_3	\mathbf{D}_4
		0	0	0	\mathbf{N}_0	\mathbf{N}_1	\mathbf{N}_2	\mathbf{N}_3	\mathbf{N}_4

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Corollary 7.M4

- Let $G(s)=N(s)D^{-1}(s)$ be a right fraction, not necessarily right coprime Let $v_i, i = 1, 2, ..., q$, be the number of linearly independent $\overline{N}_i - rows$ in T. $\deg G(s) = v_1 + v_2 + \dots + v_q$
- A left coprime fraction $\overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$ can be obtained by computing q monic null vectors using q matrices formed from each primary dependent $\overline{N}_i rows$ and its preceding LI rows.

Note:

The row echelon form can be also defined.

Realizations from Coprime Fractions

$$\mathbf{G}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s)$$

$$\mathbf{H}(s) \coloneqq \begin{bmatrix} s^{\mu_{1}} & 0\\ 0 & s^{\mu_{2}} \end{bmatrix} = \begin{bmatrix} s^{4} & 0\\ 0 & s^{2} \end{bmatrix}$$

$$\mathbf{L}(s) \coloneqq \begin{bmatrix} s^{\mu_{1}-1} & 0\\ \vdots & \vdots\\ 1 & 0\\ 0 & s^{\mu_{2}-1}\\ \vdots & \vdots\\ 0 & 1 \end{bmatrix} = \begin{bmatrix} s^{3} & 0\\ s^{2} & 0\\ s & 0\\ 1 & 0\\ 0 & s\\ 0 & 1 \end{bmatrix}$$

$$\mathbf{y}(s) = \mathbf{N}(s)\mathbf{D}^{-1}(s)\mathbf{u}(s)$$

$$\mathbf{v}(s) = \mathbf{D}^{-1}(s)\mathbf{u}(s) \rightarrow \mathbf{D}^{-1}(s)\mathbf{v}(s) = \mathbf{u}(s)$$

$$\mathbf{y}(s) = \mathbf{N}(s)\mathbf{v}(s)$$

)

Define state variables

$$\mathbf{x}(s) = \mathbf{L}(s)\mathbf{v}(s) = \begin{bmatrix} s^{\mu_{1}-1} & 0\\ \vdots & \vdots\\ 1 & 0\\ 0 & s^{\mu_{2}-1}\\ \vdots & \vdots\\ 0 & 1 \end{bmatrix} \begin{bmatrix} v_{1}(s)\\ v_{2}(s) \end{bmatrix}$$
$$= \begin{bmatrix} s^{3}v_{1}(s)\\ s^{2}v_{1}(s)\\ sv_{1}(s)\\ v_{1}(s)\\ sv_{2}(s)\\ v_{2}(s) \end{bmatrix} = \begin{bmatrix} x_{1}(s)\\ x_{2}(s)\\ x_{3}(s)\\ x_{3}(s)\\ x_{4}(s)\\ x_{5}(s)\\ x_{6}(s) \end{bmatrix} \rightarrow \mathbf{x}(t) = \begin{bmatrix} \ddot{v}_{1}\\ \ddot{v}_{1}\\ \dot{v}_{1}\\ \dot{v}_{2}\\ v_{2} \end{bmatrix}$$

Linear Systems

$$\begin{aligned} x_{1}(t) &= v_{1}^{(3)}(t) \quad x_{2}(t) = \ddot{v}_{1}(t) \quad x_{3}(t) = \dot{v}_{1}(t) \quad x_{4}(t) = v_{1}(t) \\ x_{5}(t) &= \dot{v}_{2}(t) \quad x_{6}(t) = v_{2}(t) \\ \dot{x}_{2} &= x_{1} \quad \dot{x}_{3} = x_{2} \quad \dot{x}_{4} = x_{3} \quad \dot{x}_{6} = x_{5} \\ \text{To develope } x_{1} \text{ and } x_{5}, \\ \mathbf{D}(s) &= \mathbf{D}_{hc} \mathbf{H}(s) + \mathbf{D}_{lc} \mathbf{L}(s) \\ \left[\mathbf{D}_{hc} \mathbf{H}(s) + \mathbf{D}_{lc} \mathbf{L}(s)\right] \mathbf{v}(s) = \mathbf{u}(s) \\ \mathbf{H}(s) \mathbf{v}(s) + \mathbf{D}_{hc}^{-1} \mathbf{D}_{lc} \mathbf{L}(s) \mathbf{v}(s) = \mathbf{D}_{hc}^{-1} \mathbf{u}(s) \\ \mathbf{H}(s) \mathbf{v}(s) = -\mathbf{D}_{hc}^{-1} \mathbf{D}_{lc} \mathbf{x}(s) + \mathbf{D}_{hc}^{-1} \mathbf{u}(s) \\ \mathbf{D}_{hc}^{-1} \mathbf{D}_{lc} = \vdots \begin{bmatrix} \alpha_{111} & \alpha_{112} & \alpha_{113} & \alpha_{114} & \alpha_{121} & \alpha_{122} \\ \alpha_{211} & \alpha_{212} & \alpha_{213} & \alpha_{214} & \alpha_{221} & \alpha_{222} \end{bmatrix} \\ \mathbf{D}_{hc}^{-1} = \vdots \begin{bmatrix} 1 & b_{12} \\ 0 & 1 \end{bmatrix} \end{aligned}$$

$$\begin{bmatrix} sx_{1}(s) \\ sx_{5}(s) \end{bmatrix} = -\begin{bmatrix} \alpha_{111} & \alpha_{112} & \alpha_{113} & \alpha_{114} & \alpha_{121} & \alpha_{122} \\ \alpha_{211} & \alpha_{212} & \alpha_{213} & \alpha_{214} & \alpha_{221} & \alpha_{222} \end{bmatrix} \mathbf{x}(s) + \begin{bmatrix} 1 & b_{12} \\ 0 & 1 \end{bmatrix} \mathbf{u}(s) \begin{bmatrix} \dot{x}_{1} \\ \dot{x}_{5} \end{bmatrix} = -\begin{bmatrix} \alpha_{111} & \alpha_{112} & \alpha_{113} & \alpha_{114} & \alpha_{121} & \alpha_{122} \\ \alpha_{211} & \alpha_{212} & \alpha_{213} & \alpha_{214} & \alpha_{221} & \alpha_{222} \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 & b_{12} \\ 0 & 1 \end{bmatrix} \mathbf{u} \\\mathbf{N}(s) = \begin{bmatrix} \beta_{111} & \beta_{112} & \beta_{113} & \beta_{114} & \beta_{121} & \beta_{122} \\ \beta_{211} & \beta_{212} & \beta_{213} & \beta_{214} & \beta_{221} & \beta_{222} \end{bmatrix} \mathbf{L}(s) \\\mathbf{y}(s) = \begin{bmatrix} \beta_{111} & \beta_{112} & \beta_{113} & \beta_{114} & \beta_{121} & \beta_{122} \\ \beta_{211} & \beta_{212} & \beta_{213} & \beta_{214} & \beta_{221} & \beta_{222} \end{bmatrix} \mathbf{x}(s)$$

$$\dot{\mathbf{x}} = \begin{bmatrix} -\alpha_{111} & -\alpha_{112} & -\alpha_{113} & -\alpha_{114} & \vdots & -\alpha_{121} & -\alpha_{122} \\ 1 & 0 & 0 & 0 & \vdots & 0 & 0 \\ 0 & 1 & 0 & 0 & \vdots & 0 & 0 \\ 0 & 0 & 1 & 0 & \vdots & 0 & 0 \\ \dots & \dots & \dots & \dots & \dots & \dots & \dots \\ -\alpha_{211} & -\alpha_{212} & -\alpha_{213} & -\alpha_{214} & \vdots & -\alpha_{221} & -\alpha_{222} \\ 0 & 0 & 0 & 0 & \vdots & 1 & 0 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 & b_{12} \\ 0 & 0 \\ 0 & 0 \\ \dots & \dots \\ 0 & 1 \\ 0 & 0 \end{bmatrix} \mathbf{u}$$
$$\mathbf{y} = \begin{bmatrix} \beta_{111} & \beta_{112} & \beta_{113} & \beta_{114} & \vdots & \beta_{121} & \beta_{122} \\ \beta_{211} & \beta_{212} & \beta_{213} & \beta_{214} & \vdots & \beta_{221} & \beta_{222} \end{bmatrix} \mathbf{x}$$

Controllable canonical form

Example 7.8

$$\mathbf{G}(s) = \begin{bmatrix} \frac{4s-10}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{s+1}{(s+2)^2} \end{bmatrix} =: \mathbf{G}(\infty) + \mathbf{G}_{sp}(s)$$
$$= \begin{bmatrix} 2 & 0 \\ 0 & 0 \end{bmatrix} + \begin{bmatrix} \frac{-12}{2s+1} & \frac{3}{s+2} \\ \frac{1}{(2s+1)(s+2)} & \frac{s+1}{(s+2)^2} \end{bmatrix}$$
$$\mathbf{G}_{sp}(s) = \begin{bmatrix} -6s-12 & -9 \\ 0.5 & 1 \end{bmatrix} \begin{bmatrix} s^2+2.5s+1 & 2s+1 \\ 0 & s+2 \end{bmatrix}^{-1}$$

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$$\mathbf{H}(s) = \begin{bmatrix} s^2 & 0 \\ 0 & s \end{bmatrix} \qquad \mathbf{L}(s) = \begin{bmatrix} s & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}$$
$$\mathbf{D}(s) = \begin{bmatrix} 1 & 2 \\ 0 & 1 \end{bmatrix} \mathbf{H}(s) + \begin{bmatrix} 2.5 & 1 & 1 \\ 0 & 0 & 2 \end{bmatrix} \mathbf{L}(s)$$
$$\mathbf{N}(s) = \begin{bmatrix} -6 & -12 & -9 \\ 0 & 0.5 & 1 \end{bmatrix} \mathbf{L}(s)$$
$$\mathbf{D}_{hc}^{-1} = \begin{bmatrix} 1 & 2 \\ 0 & 1 \end{bmatrix}^{-1} = \begin{bmatrix} 1 & -2 \\ 0 & 1 \end{bmatrix}$$
$$\mathbf{D}_{hc}^{-1} \mathbf{D}_{lc} = \begin{bmatrix} 1 & -2 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 2.5 & 1 & 1 \\ 0 & 0 & 2 \end{bmatrix} = \begin{bmatrix} 2.5 & 1 & -3 \\ 0 & 0 & 2 \end{bmatrix}$$

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$$\dot{\mathbf{x}} = \begin{bmatrix} -2.5 & -1 & \vdots & 3\\ 1 & 0 & \vdots & 0\\ \dots & \dots & \dots & \dots\\ 0 & 0 & \vdots & -2 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 1 & -2\\ 0 & 0\\ \dots & \dots\\ 0 & 1 \end{bmatrix} \mathbf{u}$$
$$\mathbf{y} = \begin{bmatrix} -6 & -12 & \vdots & -9\\ 0 & 0.5 & \vdots & 1 \end{bmatrix} \mathbf{x} + \begin{bmatrix} 2 & 0\\ 0 & 0 \end{bmatrix} \mathbf{u}$$

Observable canonical form can be obtained by using Left coprime fraction.

Note:

Let {A, B, C, D} be a minimal realization

$$\mathbf{C}(s\mathbf{I}-\mathbf{A})^{-1}\mathbf{B}+\mathbf{D}=\mathbf{N}(s)\mathbf{D}^{-1}(s)=\overline{\mathbf{D}}^{-1}(s)\overline{\mathbf{N}}(s)$$

which implies

$$\frac{1}{\det(s\mathbf{I} - \mathbf{A})} \mathbf{C} \Big[\mathrm{Adj}(s\mathbf{I} - \mathbf{A}) \Big] \mathbf{B} + \mathbf{D} = \frac{1}{\det \mathbf{D}(s)} \mathbf{N}(s) \Big[\mathrm{Adj}(\mathbf{D}(s)) \Big]$$
$$= \frac{1}{\det \mathbf{\overline{D}}(s)} \Big[\mathrm{Adj}(\mathbf{\overline{D}}(s)) \Big] \mathbf{\overline{N}}(s)$$

*deg $\mathbf{G}(s)$ = deg det $\mathbf{D}(s)$ = deg det $\overline{\mathbf{D}}(s)$ = dim $\mathbf{A} \leftarrow$ Proof of Theorem 7.M2 *characteristic polynomial of $\mathbf{G}(s) = k_1 \det \mathbf{D}(s) = k_2 \det \overline{\mathbf{D}}(s) = k_3 \det(s\mathbf{I} - \mathbf{A})$ *the set of column degrees of $\mathbf{D}(s)$ = the set of controllable indices of (\mathbf{A}, \mathbf{B}) *the set of row degrees of $\overline{\mathbf{D}}(s)$ = the set of observability indices of (\mathbf{A}, \mathbf{C})

Linear Systems



Find a right coprime fraction of

$$\mathbf{G}(s) = \begin{bmatrix} \frac{s^2 + 1}{s^3} & \frac{2s + 1}{s^2} \\ \frac{s + 2}{s^2} & \frac{2}{s} \end{bmatrix}$$

and then a minimal realization.

Realizations from Matrix Markov Parameters

Realizations from Matrix Markov Parameters $\mathbf{G}(s) = \mathbf{H}(1)s^{-1} + \mathbf{H}(2)s^{-2} + \mathbf{H}(3)s^{-3} + \cdots$ **H**(1) $\mathbf{H}(2)$ $\mathbf{H}(3)$ \cdots $\mathbf{H}(r)$ $\mathbf{H}(2) \quad \mathbf{H}(3) \quad \mathbf{H}(4) \quad \cdots \quad \mathbf{H}(r+1)$ $\mathbf{T} = \begin{vmatrix} \mathbf{H}(3) & \mathbf{H}(4) & \mathbf{H}(5) & \cdots & \mathbf{H}(r+2) \end{vmatrix}$ · · · · H(r) H(r+1) H(r+2) ··· H(2r-1) $\mathbf{H}(3)$ $\mathbf{H}(4)$ \cdots $\mathbf{H}(r+1)$ **H**(2) $\mathbf{H}(4) \qquad \mathbf{H}(5) \qquad \cdots \qquad \mathbf{H}(r+2)$ **H**(3) **H**(4) **H**(5) **H**(6) \cdots **H**(r+3) $\tilde{\mathbf{T}} =$ $\mathbf{T} = OC$ and $\tilde{\mathbf{T}} = O\mathbf{A}C \rightarrow O'\tilde{\mathbf{T}}C' = O'O\mathbf{A}CC'$ $\rightarrow \mathbf{A} = (O'O)^{-1}O'\tilde{\mathbf{T}}C'(CC')^{-1}$ $= O^+ \tilde{\mathbf{T}} C^+$

Realizations from Matrix Markov Parameters

Theorem 7.M7

A strictly proper rational matrix G(s) has degree *n* iff the matrix **T** has rank *n*.

Realizations from Matrix Markov Parameters

By singular value decomposition,

$$\mathbf{T} = \mathbf{K} \begin{bmatrix} \mathbf{A} \mathbf{0} \\ \mathbf{0} & \mathbf{0} \end{bmatrix} \mathbf{L}'$$

Choosing nonsingular part $(n - \dim)$,
$$\mathbf{T} = \mathbf{A} \mathbf{K} \mathbf{L}^{-1/2} \quad \mathbf{L}^{-1/2-'} =: OC$$

 $O = \mathbf{K} \mathbf{A}^{1/2} \mathbf{L}$ and $C = \frac{1/2^{-'}}{2^{-'}}$
 $O^{+} = \left[(\mathbf{A} \mathbf{K}^{*} \mathbf{K} \mathbf{A}^{*} \mathbf{A} \mathbf{K}^{-1/2} \right]^{-1} (\frac{1/2}{2^{-'}})^{-'}$
 $O^{+} = \mathbf{A} \mathbf{K}^{1/2-'}$
 $O^{+} = \mathbf{A} \mathbf{K}^{1/2-'}$
 $\mathbf{A} = O^{+} \mathbf{\tilde{T}} C^{+}$
 $\mathbf{B} = \text{first } p \text{ columns of } C$
 $\mathbf{C} = \text{first } q \text{ rows of } O$
 $O'O = \mathbf{A} \mathbf{K}^{*} \mathbf{K} \mathbf{A} \mathbf{A}^{-1/2} =$
 $CC' = \mathbf{A} \mathbf{L}^{2} \mathbf{\tilde{L}} \mathbf{A} \mathbf{A}^{-1/2} =$
 $\mathbf{C} C' = \mathbf{A} \mathbf{L}^{2} \mathbf{\tilde{L}} \mathbf{A} \mathbf{A}^{-1/2} =$
 $\mathbf{C} = \text{Balanced realization}$

4

Summary

Degree of transfer function Coprimeness and minimal realization Computing Coprime fraction (Sylvester matrix) Controllable form, Observable form Controllability form, Observability form (from Henkel Matrix) Balanced realization

Degree of transfer function matrix, Unimodular Greatest common right divisor, Left(right) multiple column(row) degree, column(row) reduced, Coprimeness of transfer function matrix, Computing Right(Left) coprime fraction Minimum realizations(controllable/Observable/Balanced-Henkel)