

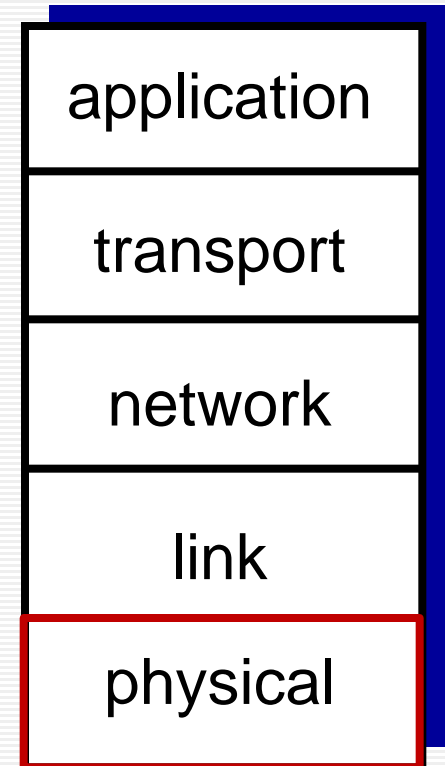
# Review

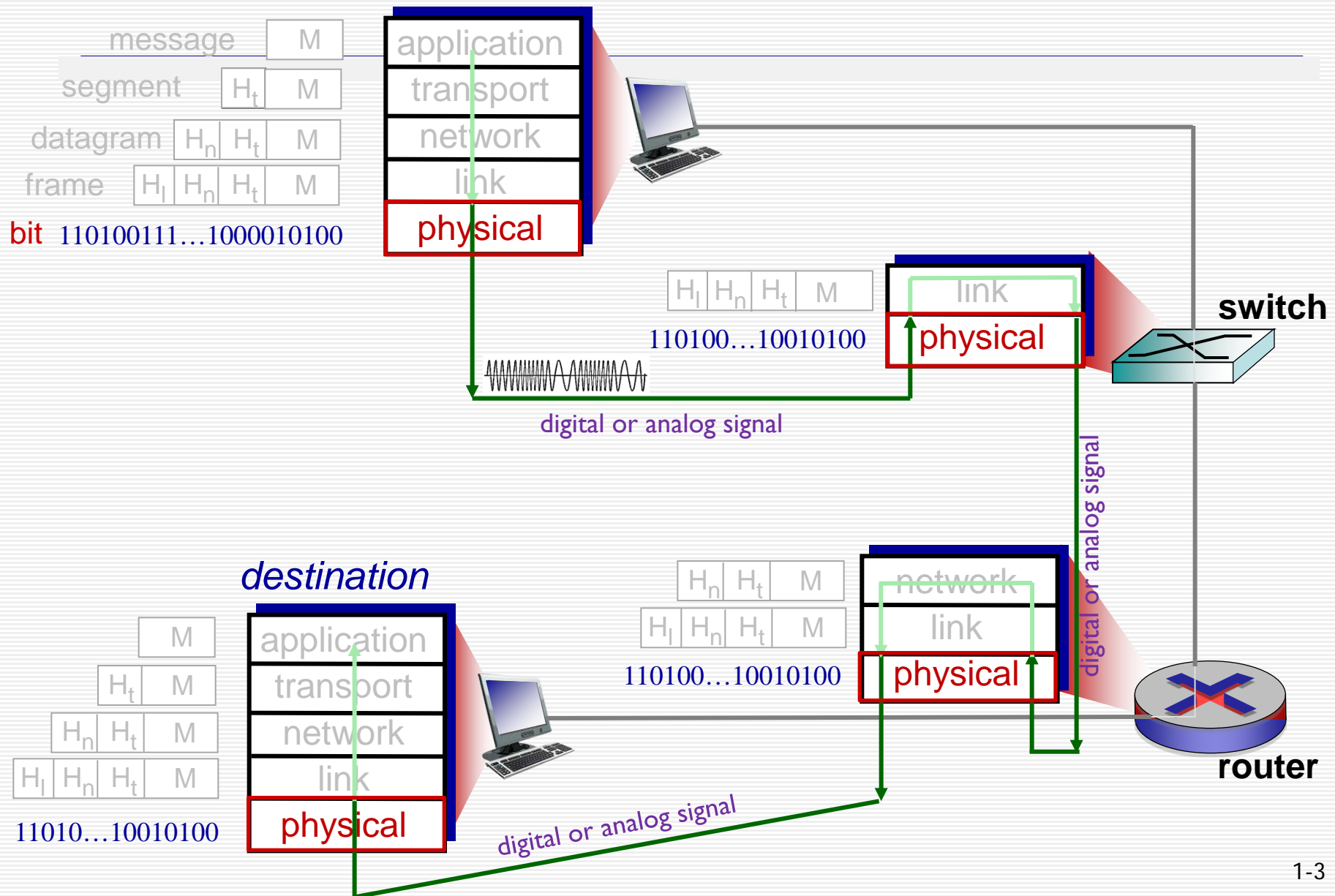
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# Protocol stack

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- *application*: supporting network applications
- *transport*: data transfer between applications (peer processes)
  - TCP, UDP
- *network*: routing of packets from source host to destination host
  - IP, routing protocols
- *link*: data transfer between neighboring network elements
  - Ethernet, 802.11 (WiFi), Cellular
- *physical*: signal representing bits on the medium (data encoding/decoding)

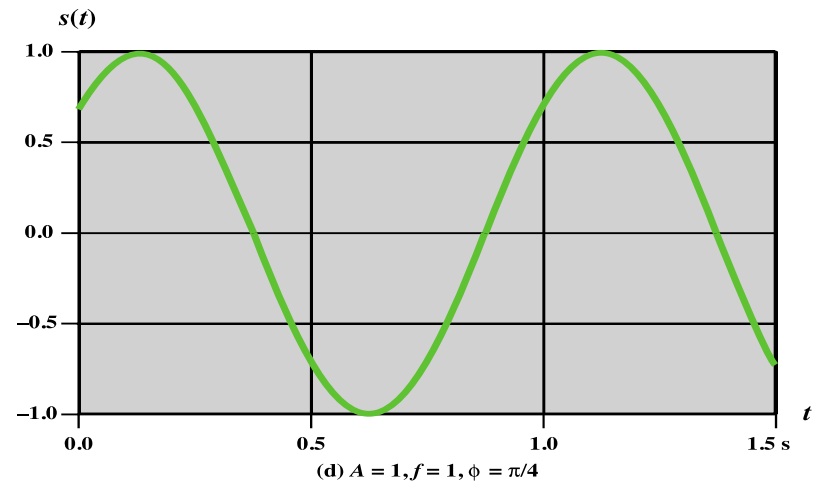
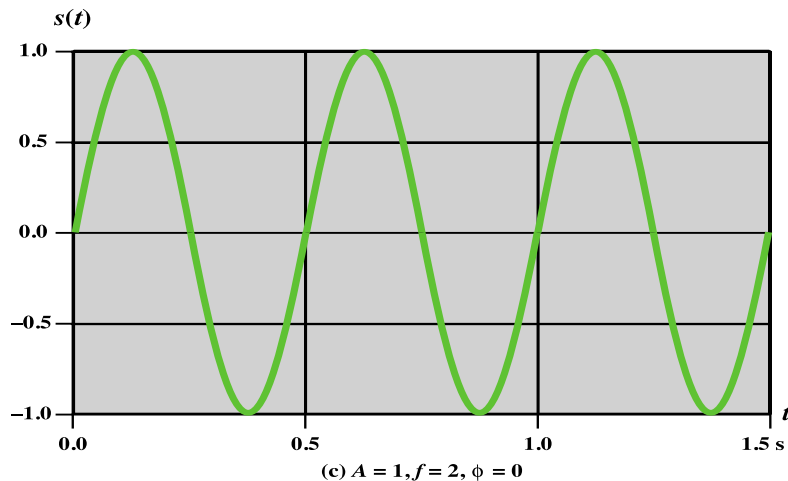
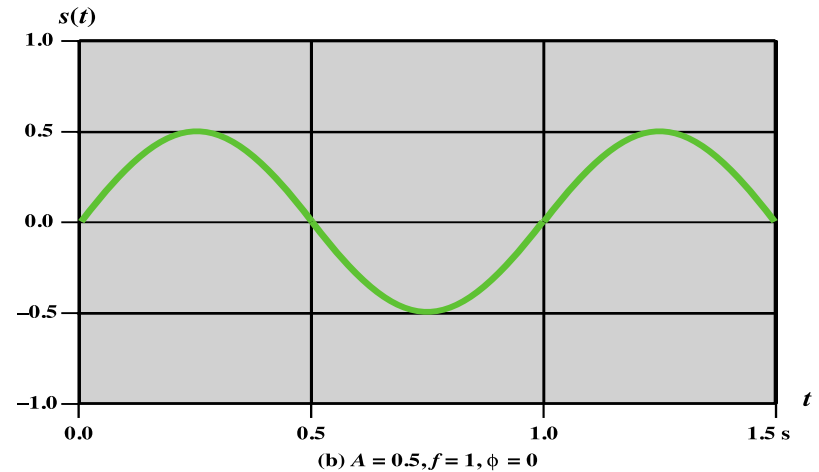
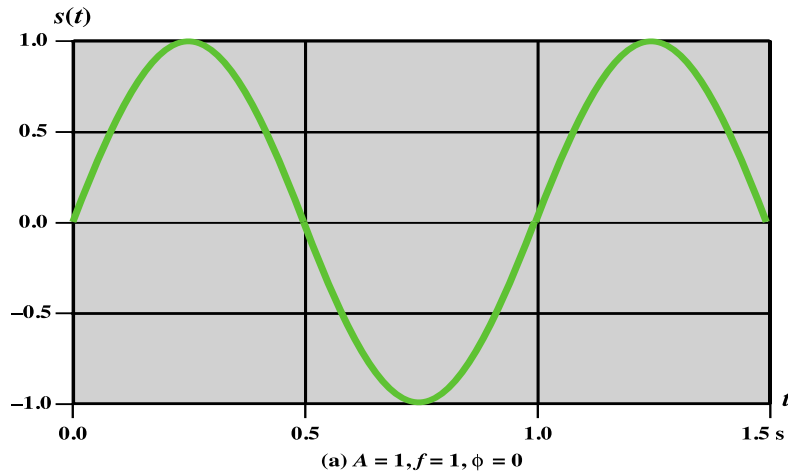




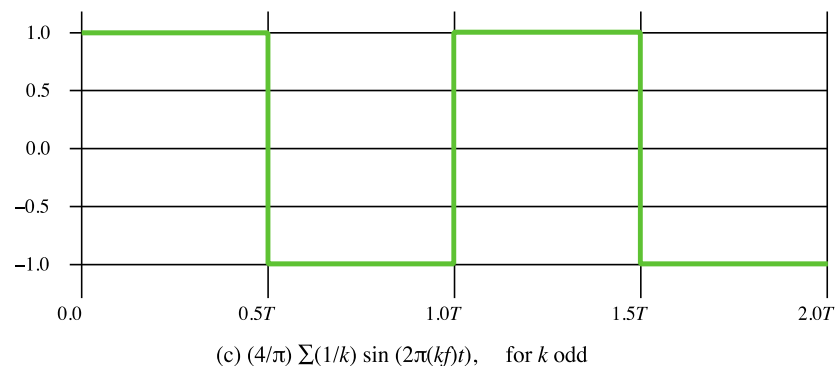
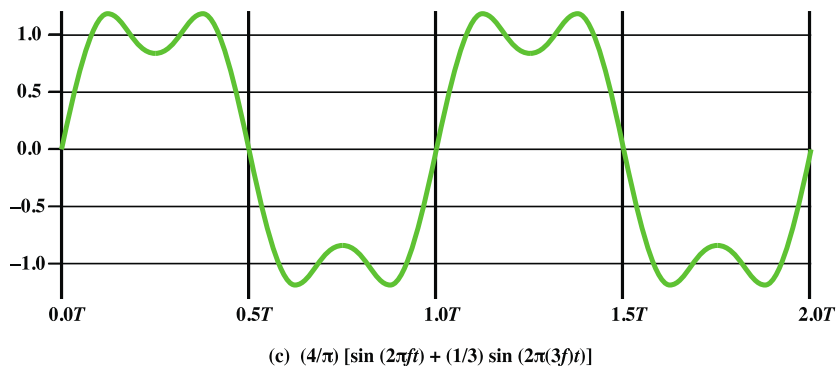
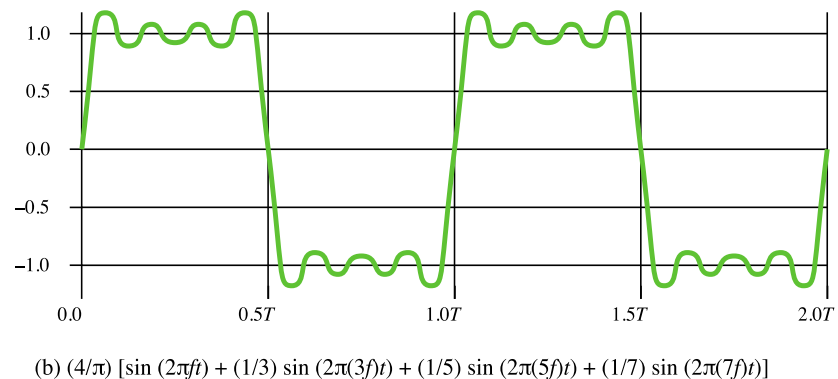
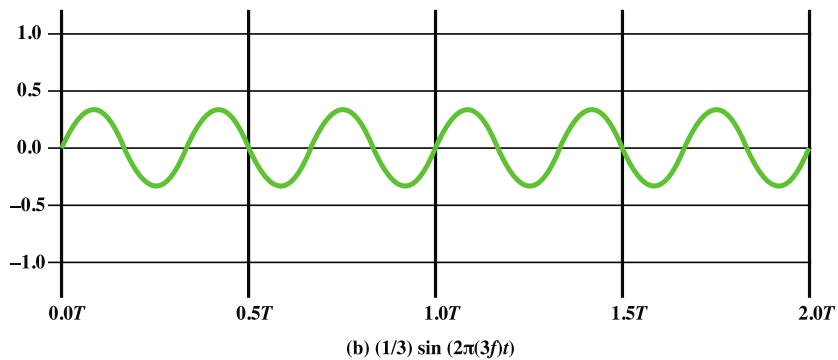
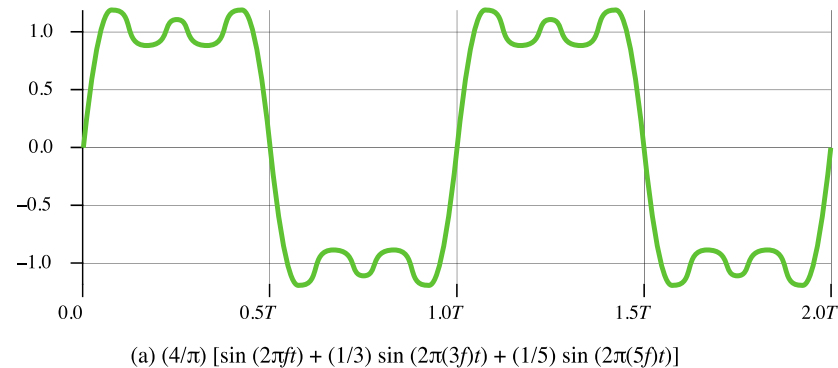
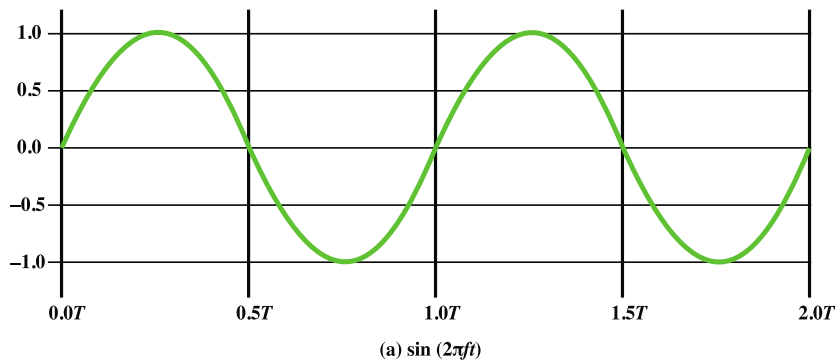
# Frequency Spectrum Bandwidth

# Sinusoidal Signal

- Fundamental periodic signal
- Can be represented by three parameters (peak amplitude, frequency, phase)



**Figure 3.3**  $s(t) = A \sin(2\pi ft + \phi)$



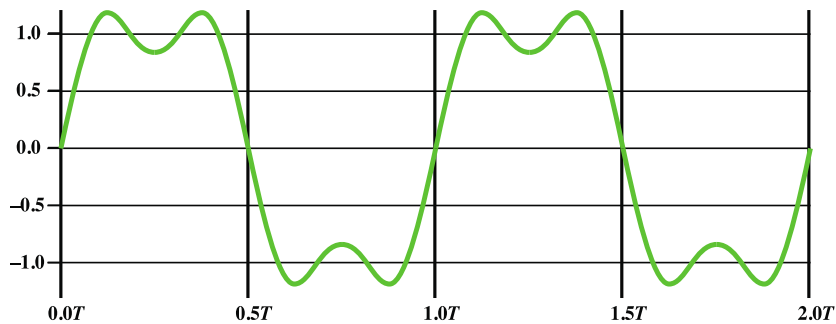
**Figure 3.4** Addition of Frequency Components ( $T = 1/f$ )

**Figure 3.7** Frequency Components of Square Wave ( $T = 1/f$ )

# Frequency Domain Concepts

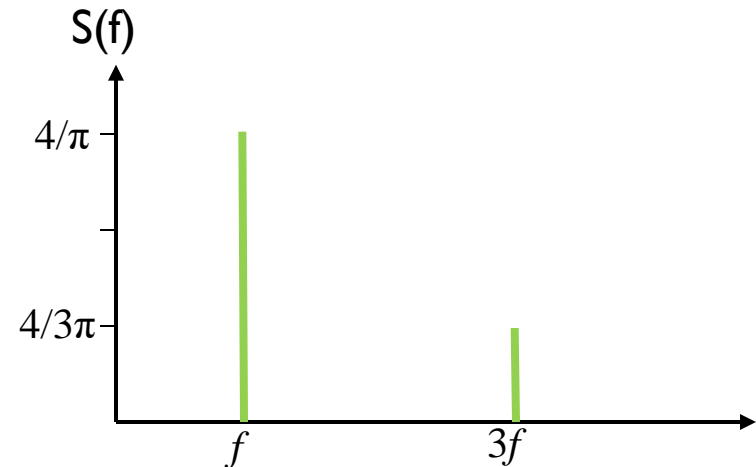
- Signals are made up of many frequencies
- Components are sine waves
- Fourier analysis can show that any signal is made up of components at various frequencies, in which each component is a sinusoid
- Can plot frequency domain functions

## Periodic Signal



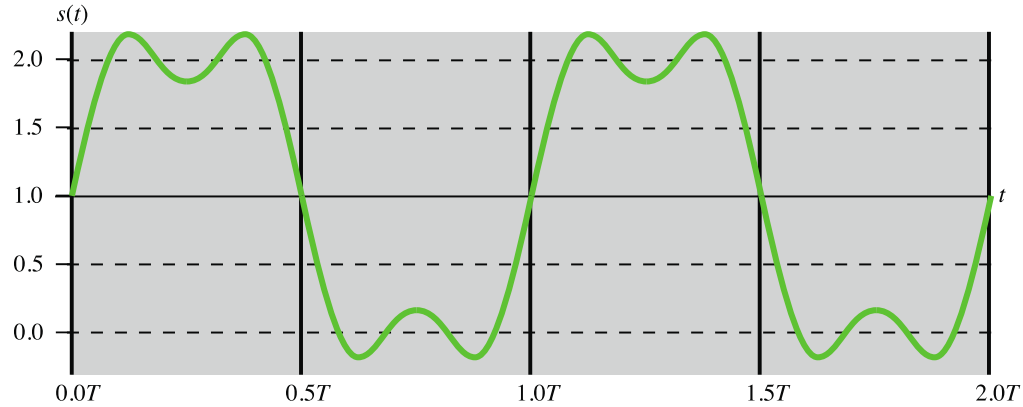
(c)  $(4/\pi) [\sin(2\pi ft) + (1/3) \sin(2\pi(3f)t)]$

time domain

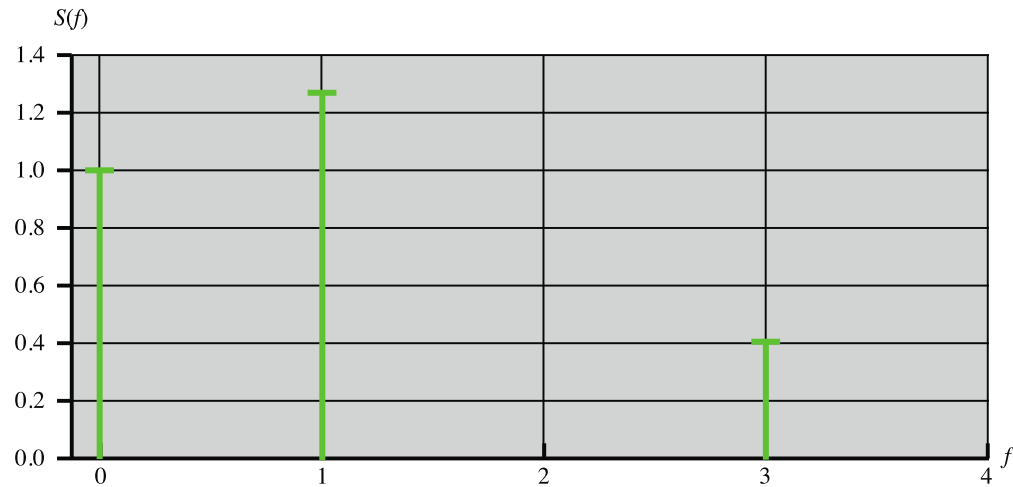


frequency domain

# Signal with dc component



(a)  $s(t) = 1 + (4/\pi) [\sin(2\pi f t) + (1/3) \sin(2\pi(3f)t)]$



(b)  $S(f)$

**Figure 3.6 Signal with dc Component**



# Non-periodic Signal

- Even non-periodic signals can be characterized in the frequency domain using a continuous spectrum of frequency components

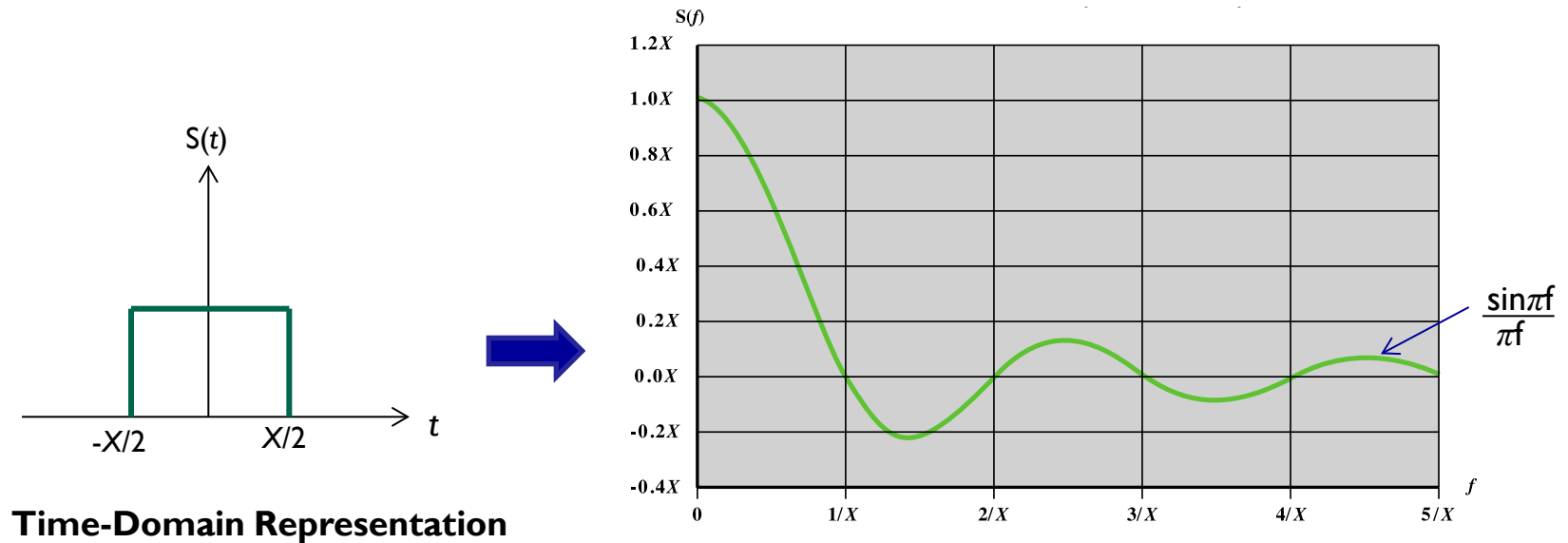
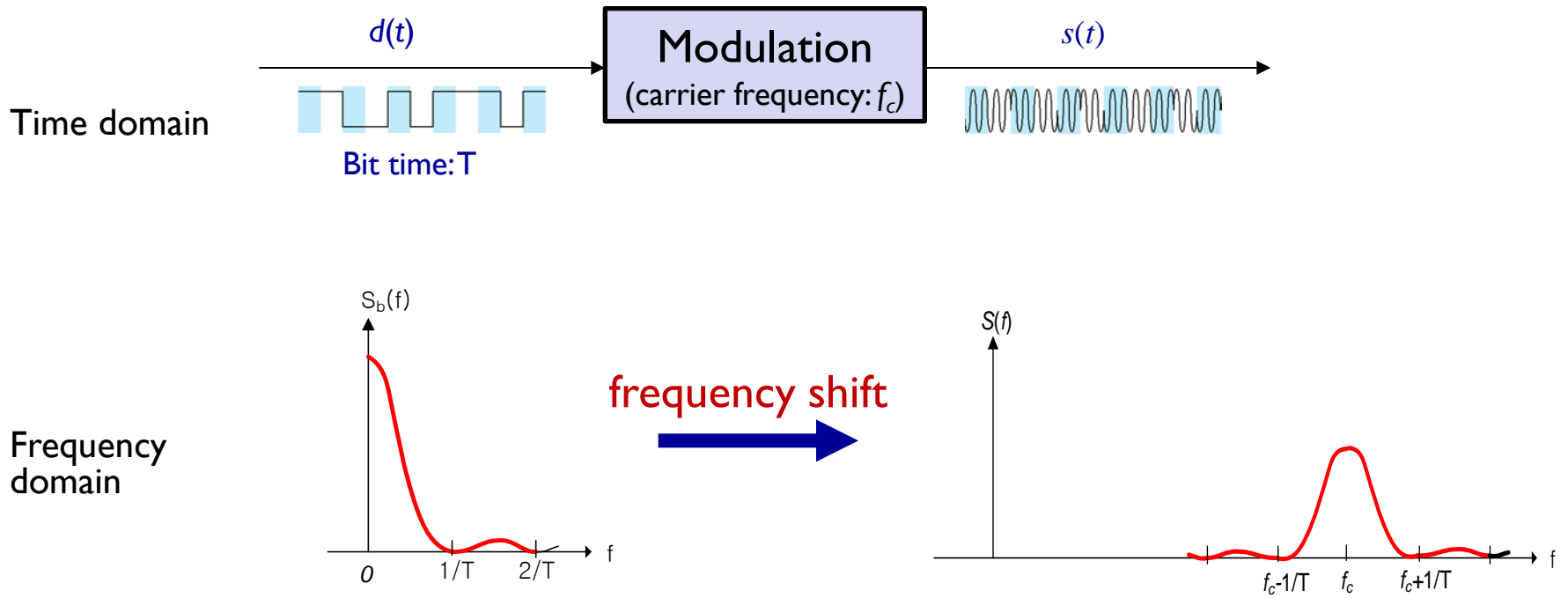


Figure 3.5 Frequency-Domain Representations

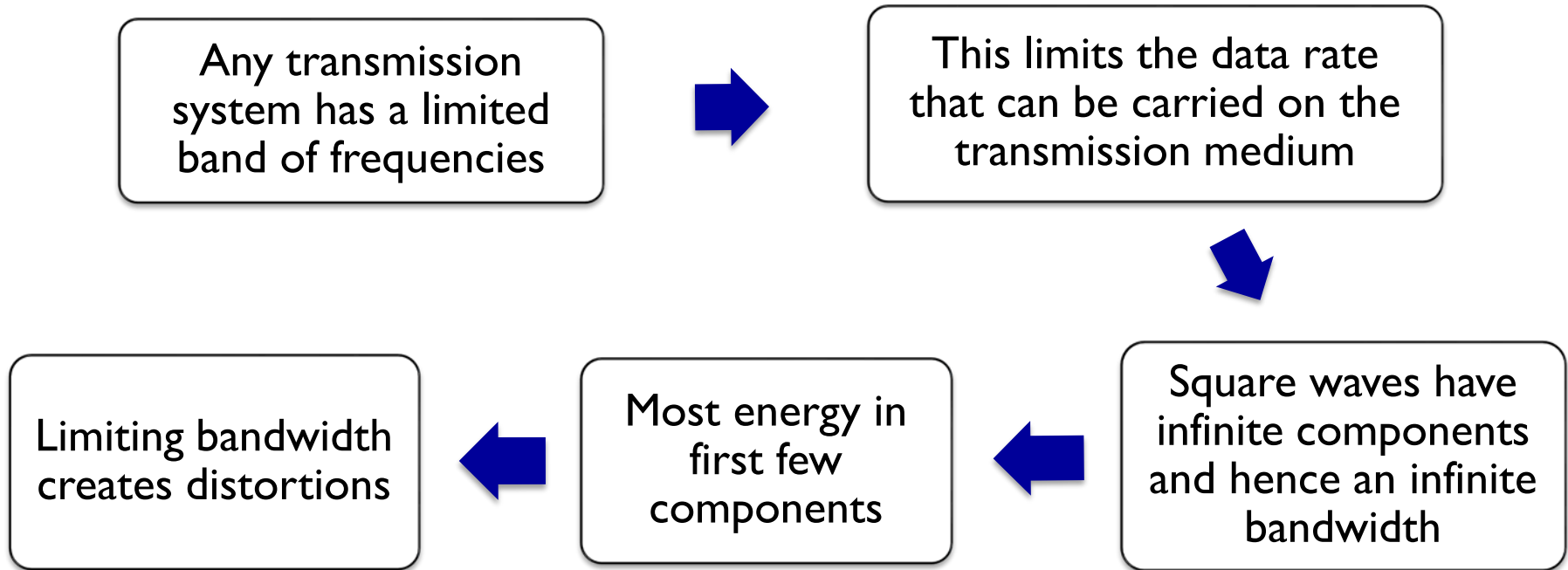
# Modulation



# Terminology

- **Signal Spectrum**
  - Range of frequencies that the signal contains
- **Absolute bandwidth**
  - Width of the spectrum
- **Effective bandwidth**
  - Band of frequencies which contains most of the energy of the signal
- **DC component**
  - When the signal contains zero frequency
  - The mean value of waveform
  - A waveform without dc component is known as a dc-balanced waveform

# Data Rate and Bandwidth



❖ **There is a direct relationship between data rate and bandwidth**

❖ **Shannon Capacity:**

$$C = B \log_2 (1 + \text{SNR})$$

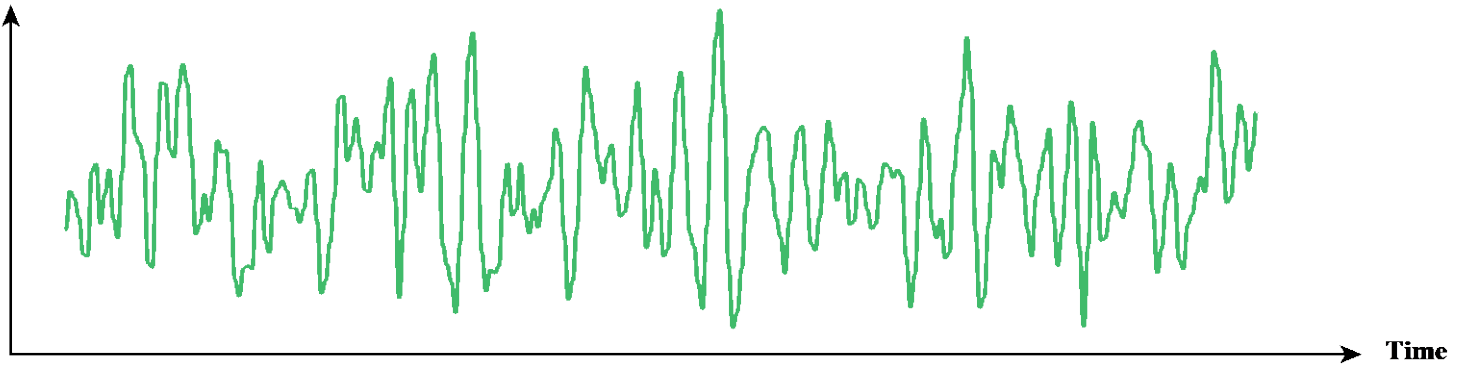
Maximum rate      Bandwidth      Ratio of signal power to noise power

## Chapter 5

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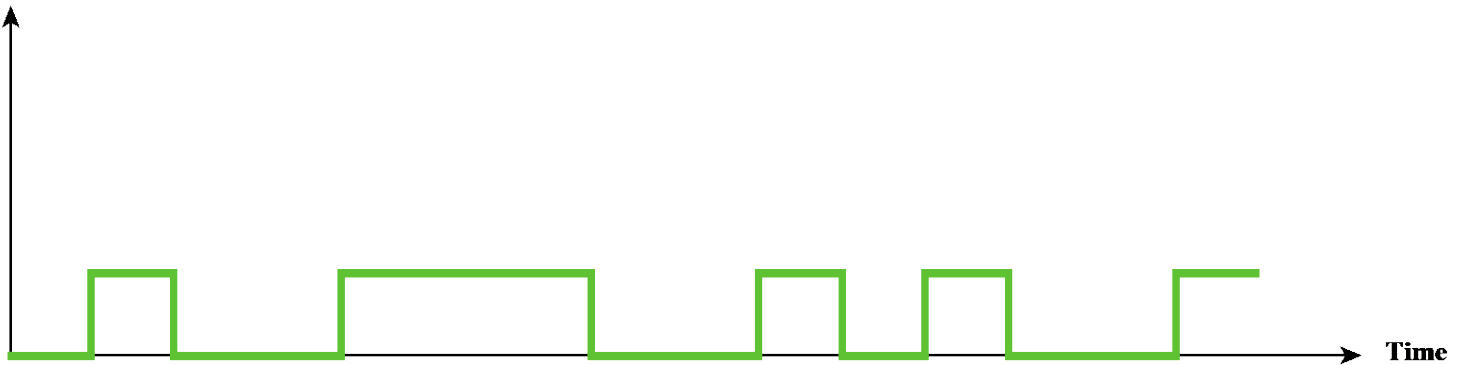
# Signal Encoding Techniques

**Amplitude  
(volts)**



**(a) Analog**

**Amplitude  
(volts)**



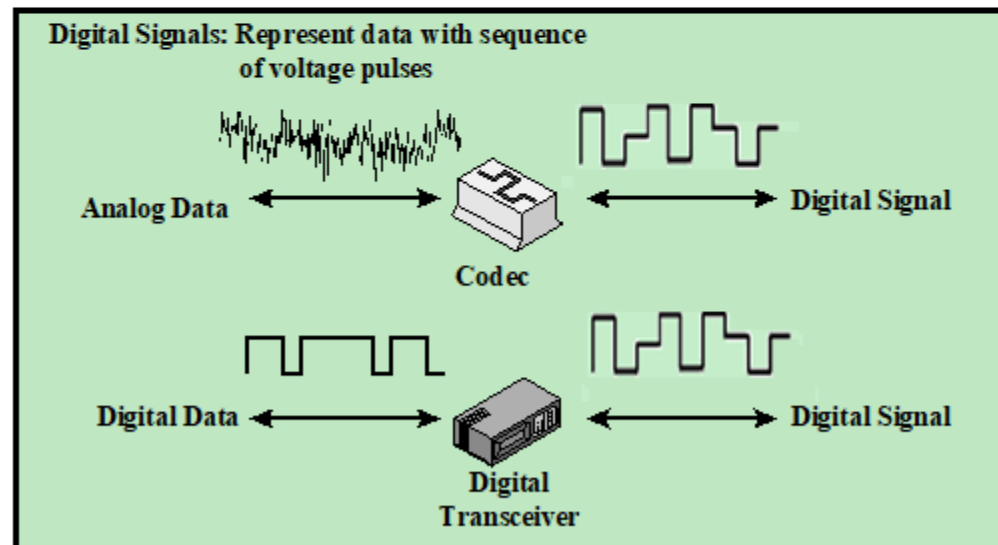
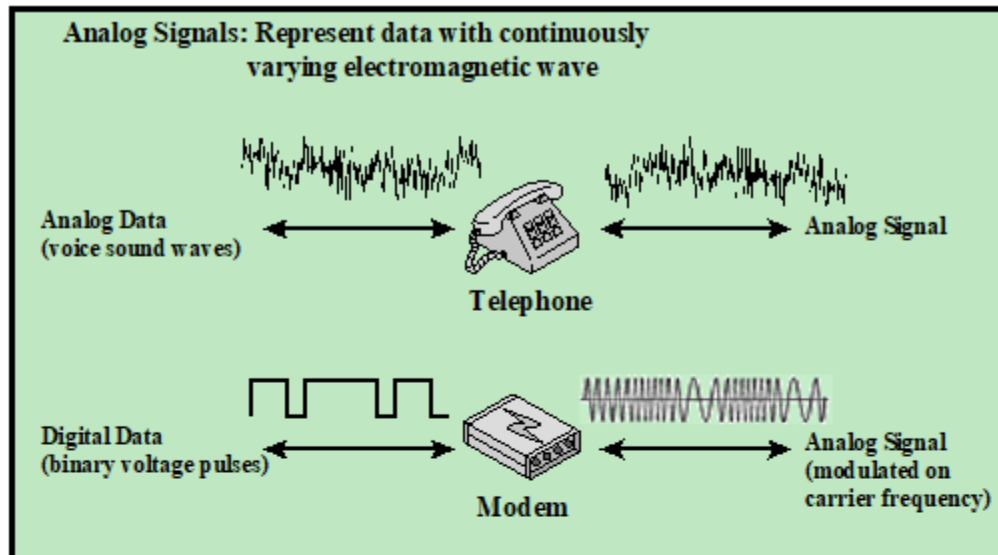
**(b) Digital**

**Figure 3.1 Analog and Digital Waveforms**

# Encoding Technique

---

- Digital Data  $\rightarrow$  Digital Signal
- Digital Data  $\rightarrow$  Analog Signal
- Analog Data
  - $\rightarrow$  Digital Data  $\rightarrow$  Digital Signal
  - $\rightarrow$  Digital Data  $\rightarrow$  Analog Signal
- Analog Data  $\rightarrow$  Analog Signal



**Figure 3.13 Analog and Digital Signaling of Analog and Digital Data**



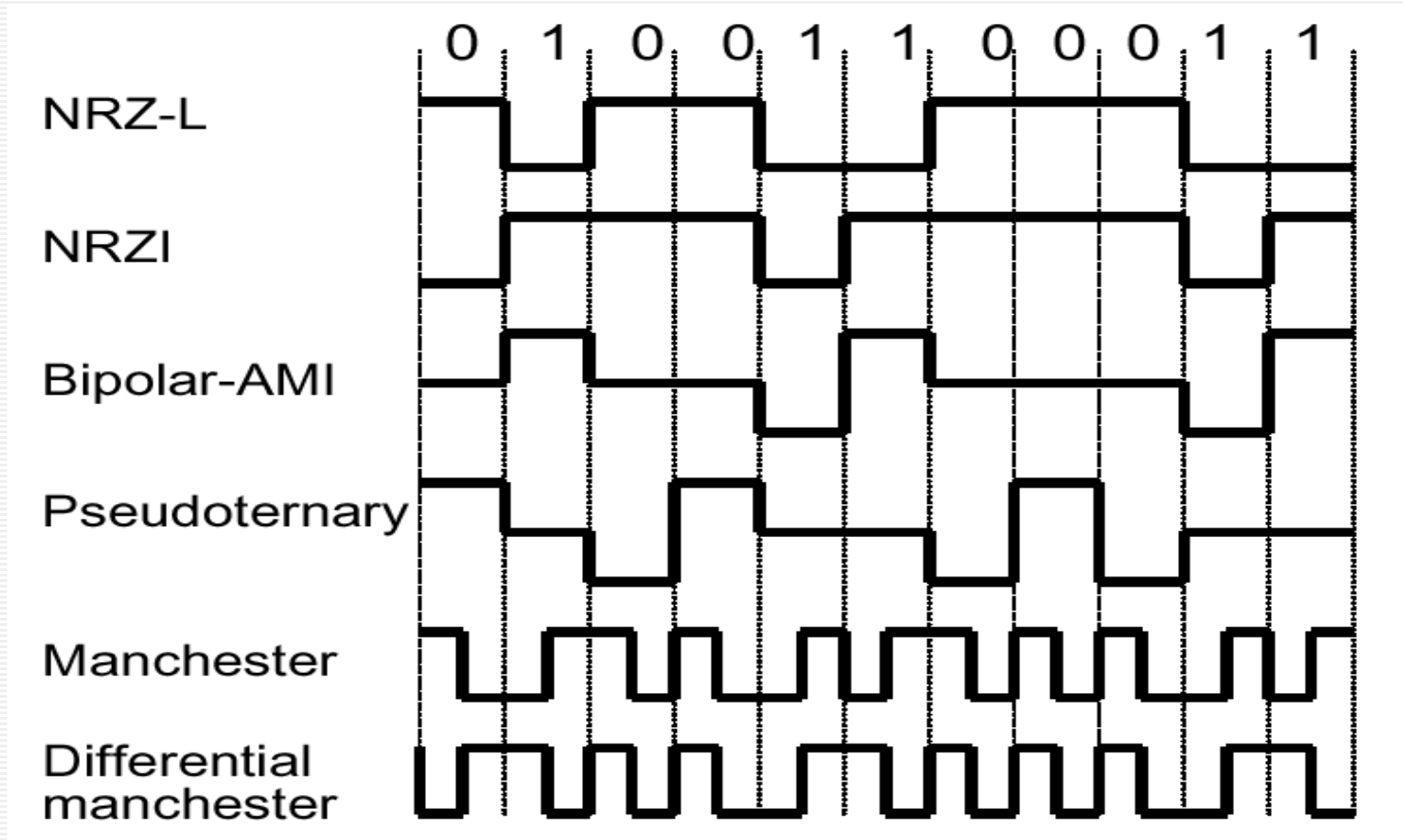
# Digital Data → Digital Signals (1)

---

- A **digital signal** is a sequence of discrete, discontinuous voltage pulses.
- Each pulse is a **signal element**.
- **Encoding Scheme**: mapping a digital data into signal element
- Binary data is transmitted by encoding each data bit into signal elements.

# Digital Data → Digital Signals (2)

- Various encoding schemes

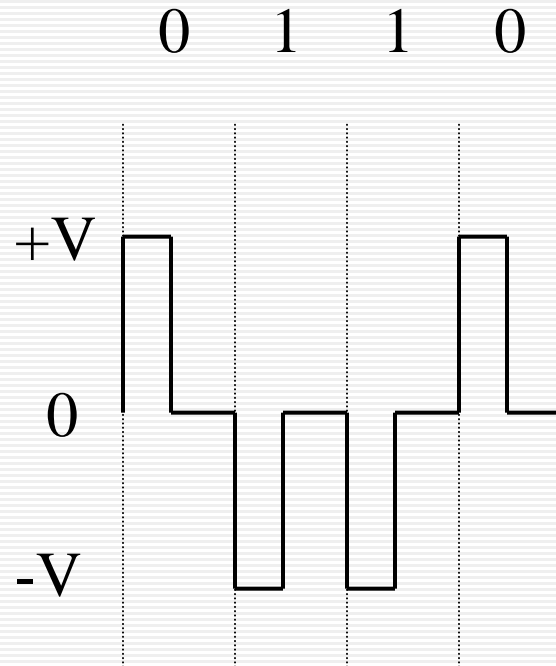


# Digital Data → Digital Signals (3)

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## RZ(Return to Zero)

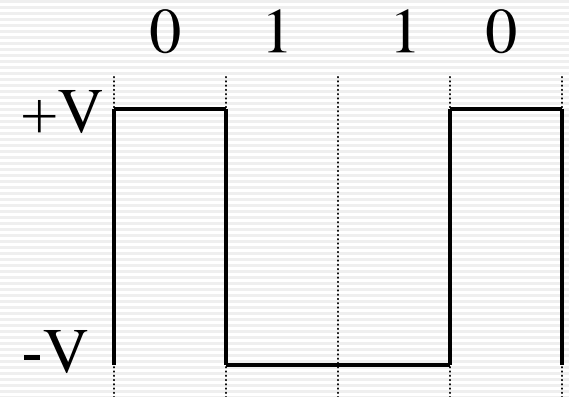
- 0: positive pulse
- 1: negative pulse
- Signal returns to zero after each encoded bit



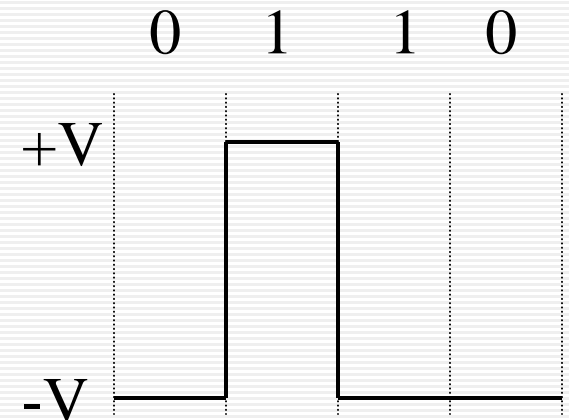
# Digital Data → Digital Signals (4)

## NRZ(Nonreturn to Zero)

- Voltage level is constant during bit interval (no return to a zero voltage level)
- NRZ-L(NRZ Level)
  - 0: positive voltage
  - 1: negative voltage
- NRZ-I(NRZ Inverted)
  - a form of differential encoding
  - 0: no transition
  - 1: transition at the beginning of bit interval
- NRZ strengths
  - simple, and efficiently use bandwidth
- NRZ limitations
  - presence of dc component
  - lack of synchronization capability



NRZ-L



NRZ-I

# Digital Data → Digital Signals (5)

---

## Multilevel Binary

- Bipolar-AMI(Alternate Mark Inversion)
  - three voltage levels(positive, zero, negative)
  - 0: zero voltage
  - 1: alternately by positive and negative voltages
  - better synchronization than NRZ
  - no dc component
  - error detection capability
- Pseudoternary
  - Same as bipolar-AMI, except representation of 0 and 1 is interchanged

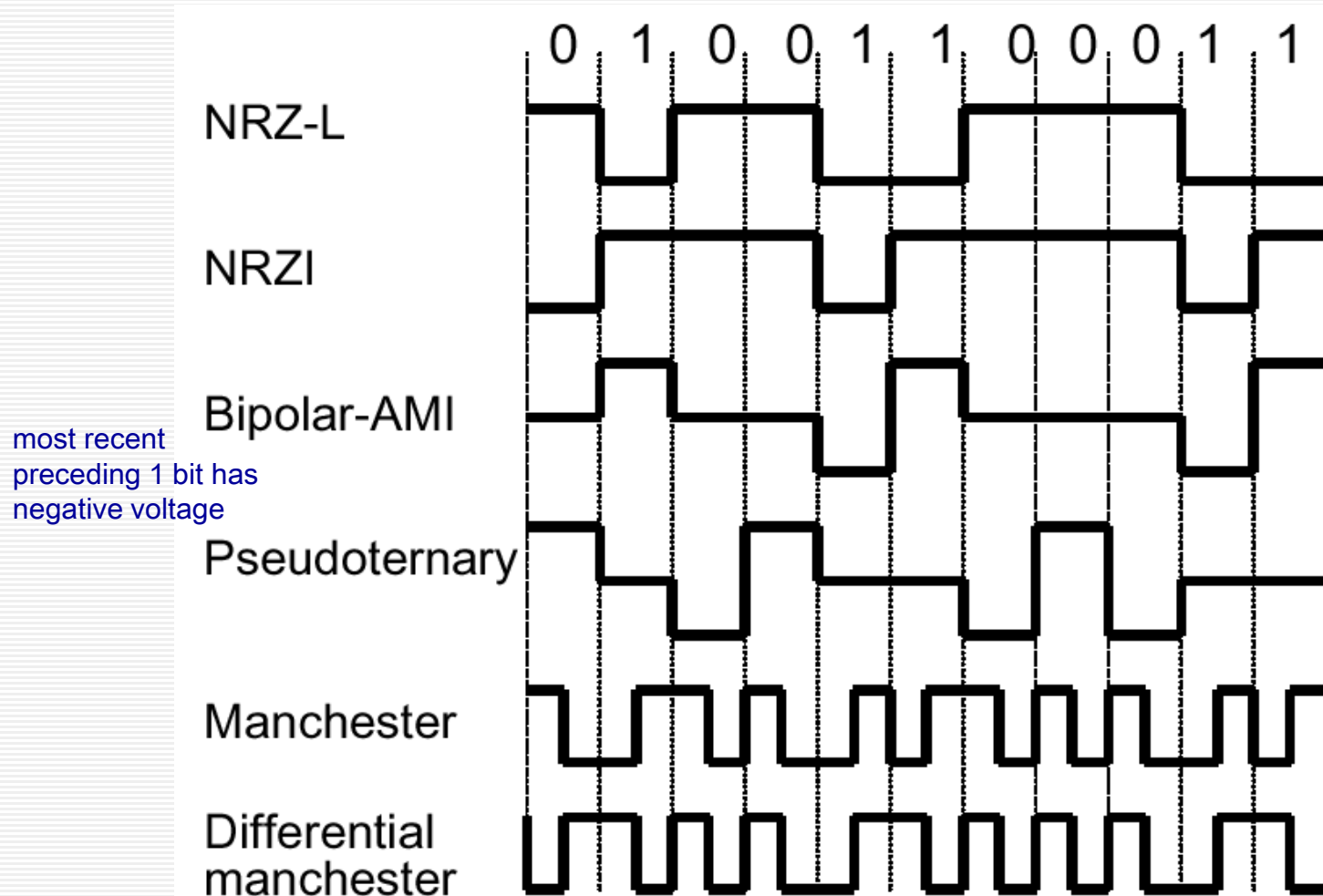
# Digital Data → Digital Signals (6)

---

## Biphase

- Always a transition at the middle of each bit interval
- Manchester
  - 0: high to low transition
  - 1: low to high transition
- Differential Manchester
  - 0: transition at the beginning of bit interval
  - 1: no transition
- Synchronization and error detection capability, and no dc component

# Digital Data → Digital Signals (7)



# Digital Data → Digital Signals (8)

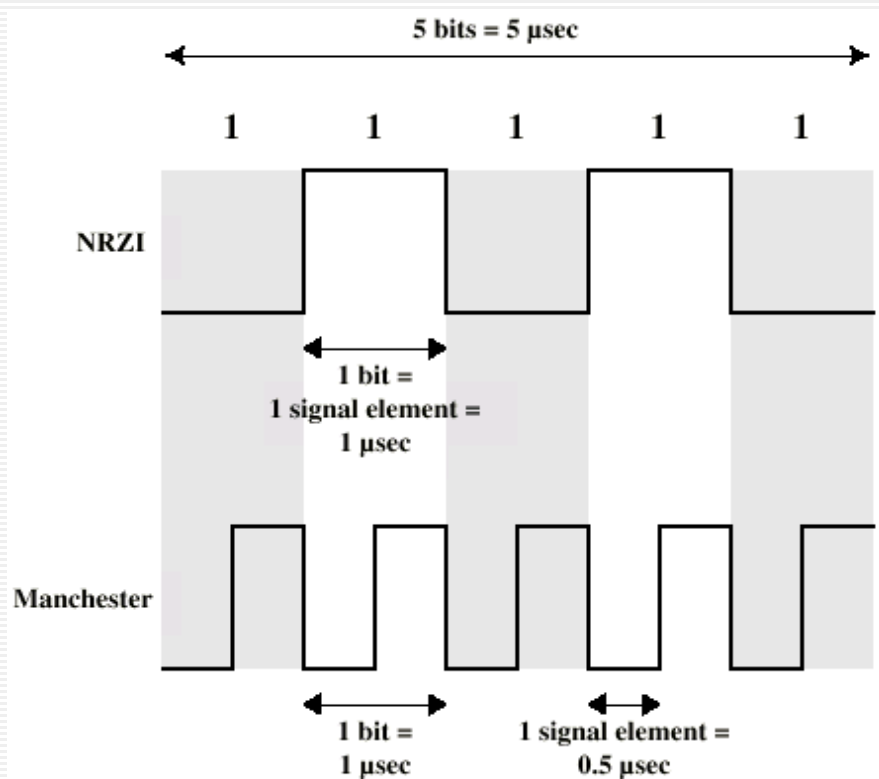
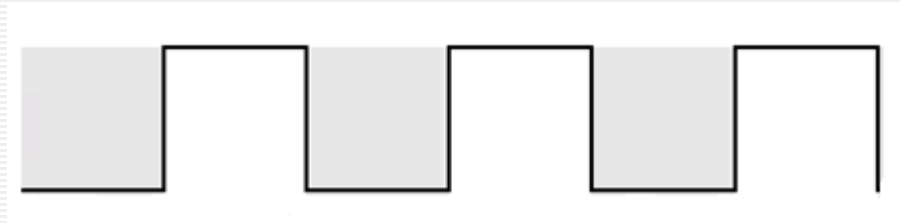


Figure 5.5 A Stream of Binary Ones at 1 Mbps



# Digital Data → Digital Signals (9)

---



NRZI: 6 bits

Manchester: 3 bits ⇒ lower data rate

# Digital Data → Digital Signals (10)

---

## Scrambling Techniques

- For long-distance communications
- No dc component, good synchronization, and error detection capability, without reduction in data rate
- B8ZS(Bipolar with 8-Zeros Substitution)
  - Based on bipolar-AMI
  - 8 consecutive zeros are encoded as either 000+-0-+ or 000-+0+- so that two code violations always occur

# Digital Data → Digital Signals (11)

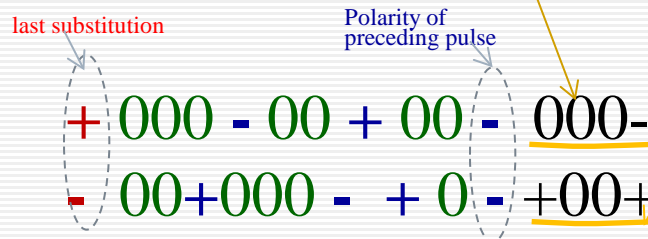
## ■ HDB3(High-Density Bipolar 3-Zeros)

- 4 zeros are encoded as either 000-, 000+, +00+, or -00-
- Substitution rule :
  - the 4th bit is always a code violation,
  - successive violations are of alternate polarity (not to introduce dc component)

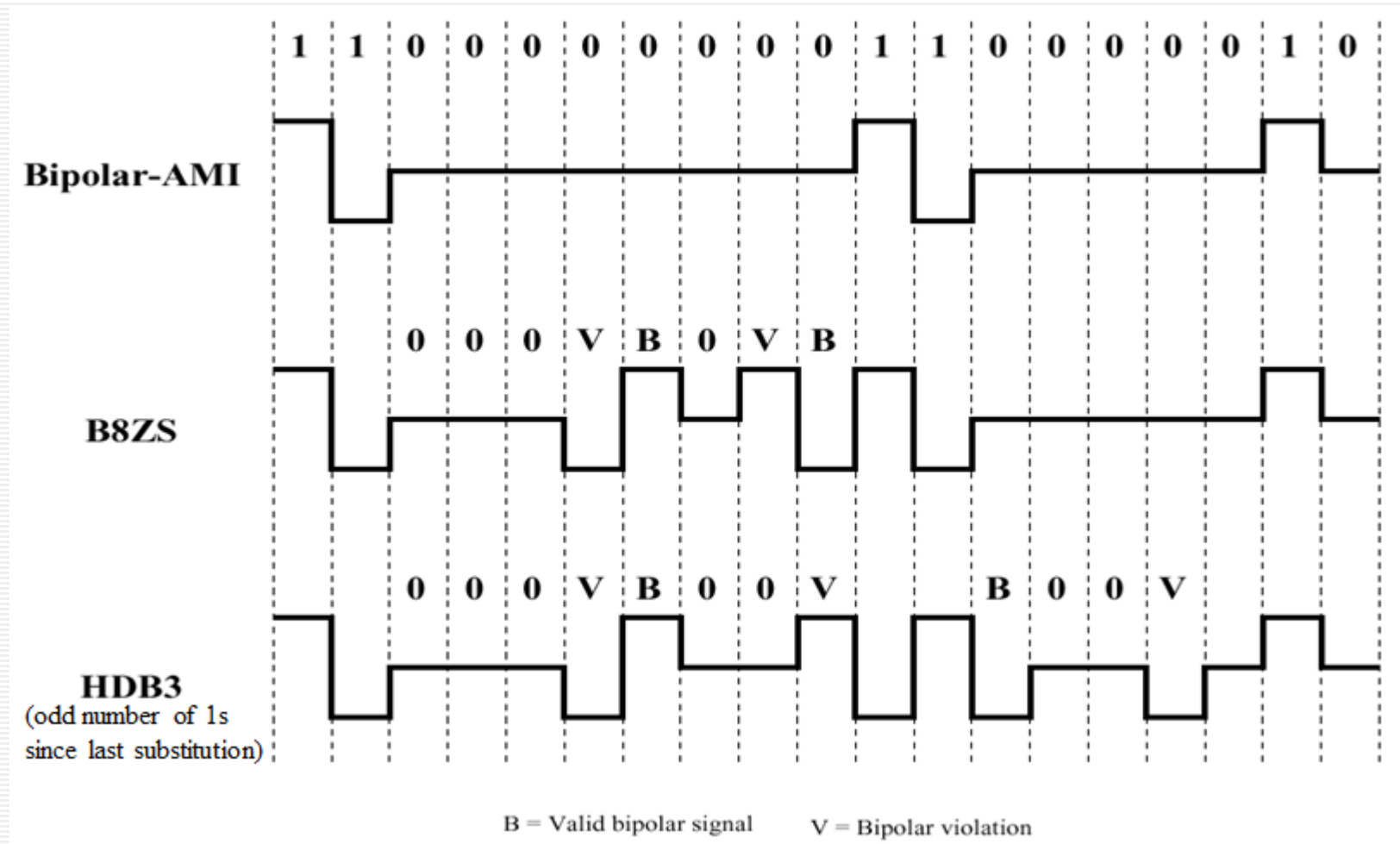
| Polarity of preceding pulse | Number of ones since last substitution |      |
|-----------------------------|--|------|
|                             | odd                                    | even |
| -                           | 000-                                   | +00+ |
| +                           | 000+                                   | -00- |



Example:



# Digital Data → Digital Signals (12)



# Digital Data → Digital Signals (13)

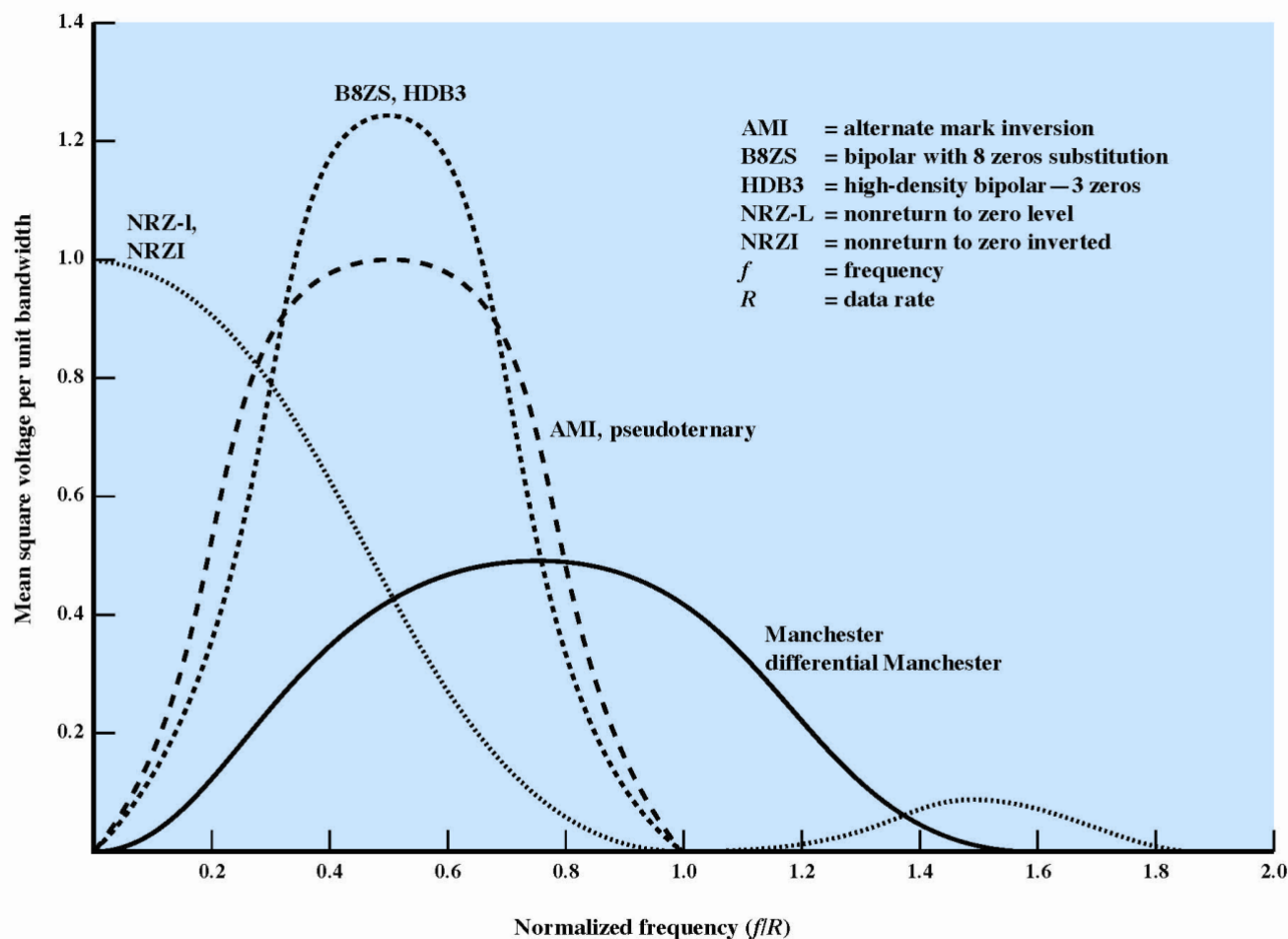


Figure 5.3 Spectral Density of Various Signal Encoding Schemes

# Digital Data → Digital Signals (14)

---

- Evaluation factors for encoding schemes
  - Signal spectrum
    - less bandwidth, no dc component
    - shape of spectrum(better to center in the middle of bandwidth)
  - Self-clocking capability
    - It is desired for synchronization
  - Error detection capability
  - Signal interference and noise immunity
  - Cost and complexity

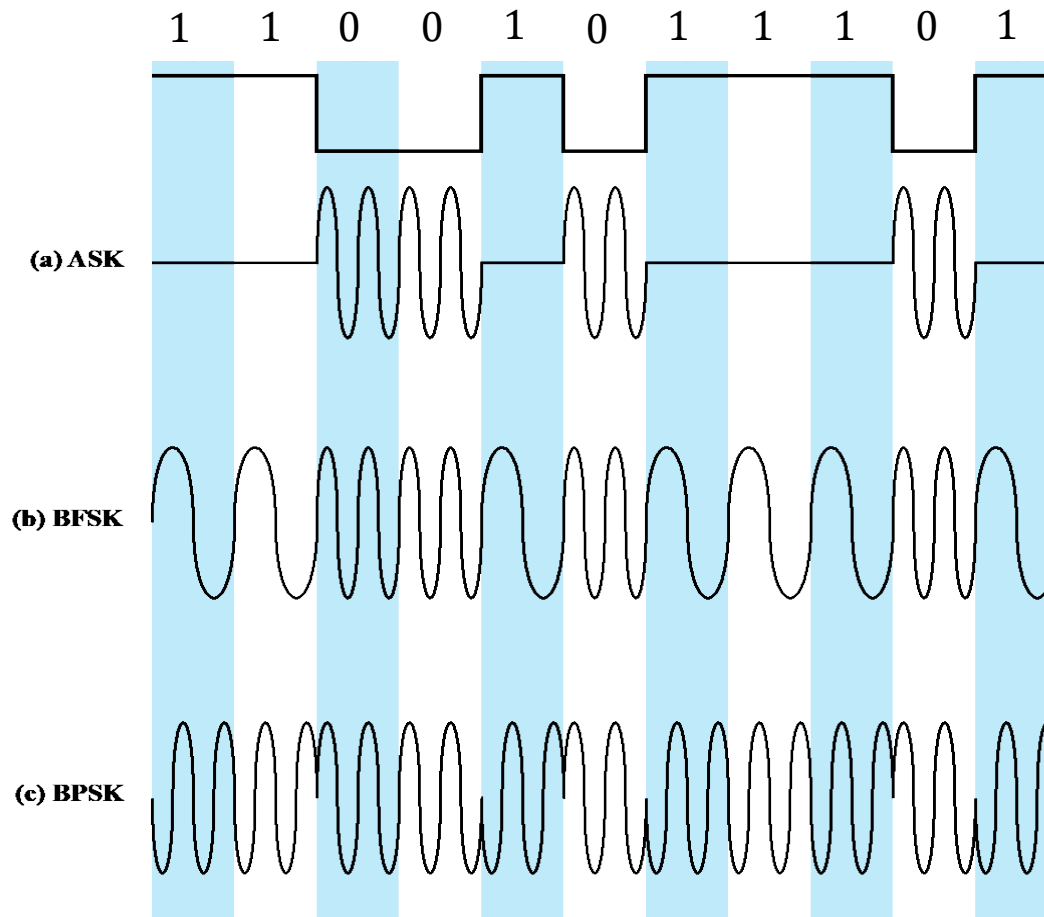
# Digital Data → Analog Signals (1)

---

- Encoding is by modulation of a continuous sinusoidal carrier signal.  $A \cos(2\pi f_c t + \theta_c)$
- This involves alteration of some characteristics of the carrier signal-amplitude, frequency, or phase.
- Various encoding techniques: ASK, FSK, PSK, ...

|          | ASK                             | FSK                             | PSK                              |
|----------|---------------------------------|---------------------------------|----------------------------------|
| Binary 0 | $A \cos(2\pi f_c t + \theta_c)$ | $A \cos(2\pi f_1 t + \theta_c)$ | $A \cos(2\pi f_c t)$             |
| Binary 1 | 0                               | $A \cos(2\pi f_2 t + \theta_c)$ | $A \cos(2\pi f_c t + 180^\circ)$ |

# Digital Data → Analog Signals (2)



**Figure 5.7 Modulation of Analog Signals for Digital Data**



# Digital Data → Analog Signals (3)

## QPSK(Quadrature PSK) modulation

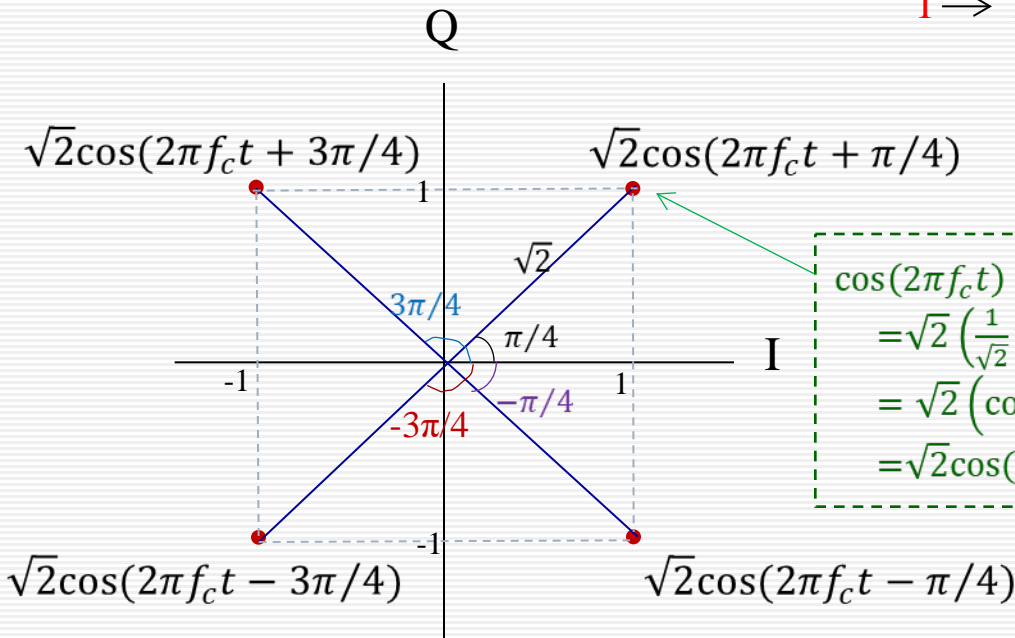
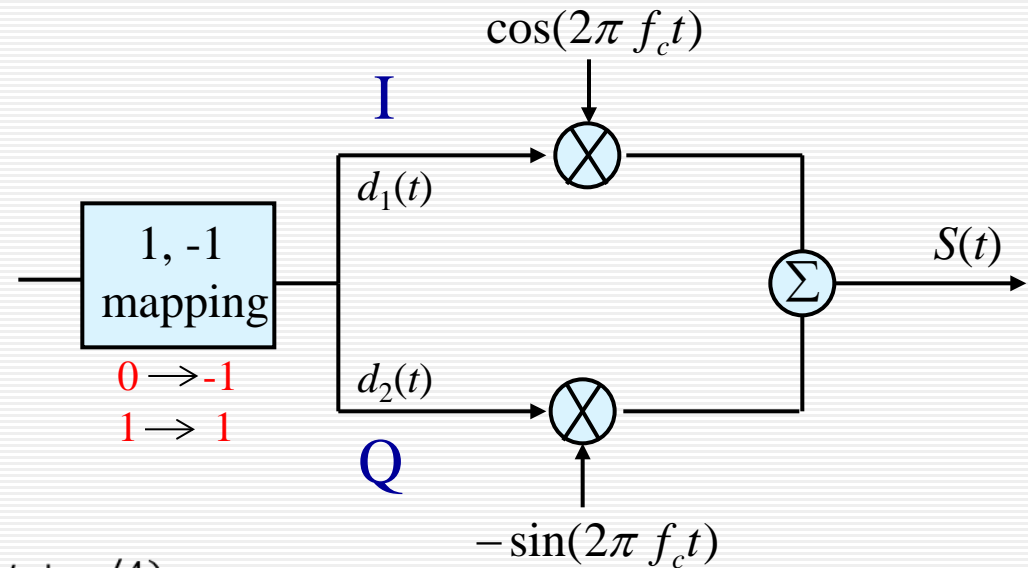
Bit phase IQ

1 1:  $\pi/4$  (+1+j)

0 1:  $3\pi/4$  (-1+j)

0 0:  $-3\pi/4$  (-1-j)

1 0:  $-\pi/4$  (+1-j)

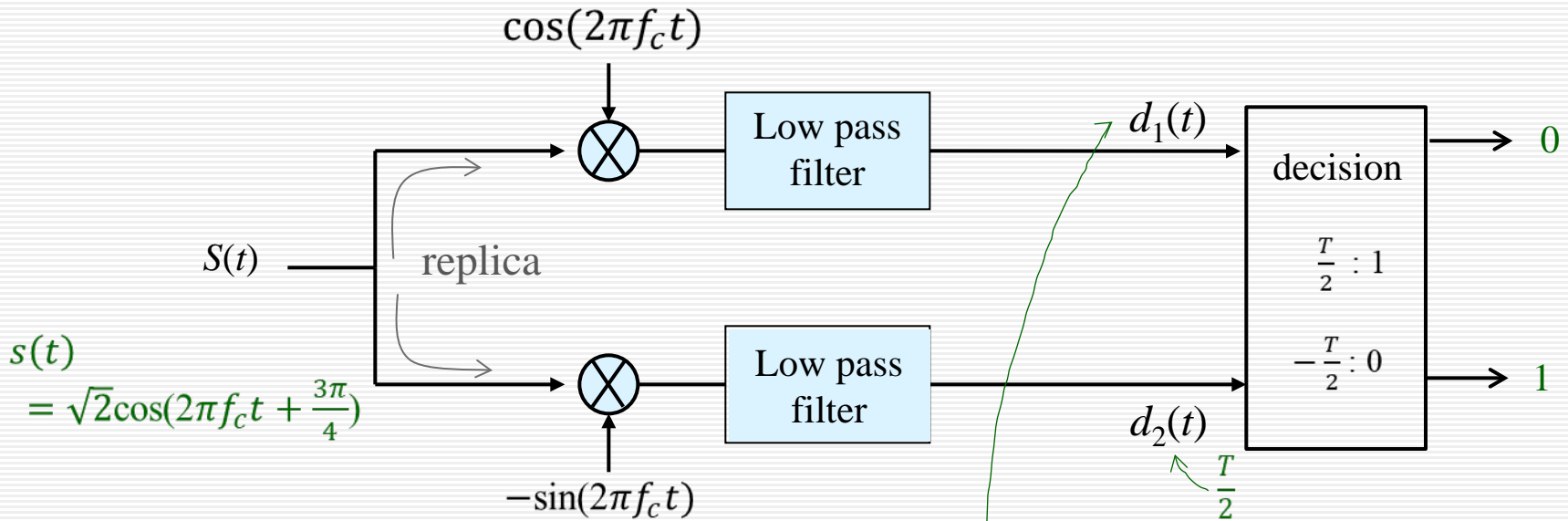


$$\begin{aligned}
 & \cos(2\pi f_c t) - \sin(2\pi f_c t) \\
 &= \sqrt{2} \left( \frac{1}{\sqrt{2}} \cos(2\pi f_c t) - \frac{1}{\sqrt{2}} \sin(2\pi f_c t) \right) \\
 &= \sqrt{2} \left( \cos\left(\frac{\pi}{4}\right) \cos(2\pi f_c t) - \sin\left(\frac{\pi}{4}\right) \sin(2\pi f_c t) \right) \\
 &= \sqrt{2} \cos\left(2\pi f_c t + \frac{\pi}{4}\right)
 \end{aligned}$$

$$\begin{aligned}
 \cos(A + B) &= \cos A \cos B - \sin A \sin B \\
 \sin(A + B) &= \sin A \cos B + \cos A \sin B
 \end{aligned}$$

# Digital Data → Analog Signals (4)

## QPSK demodulation



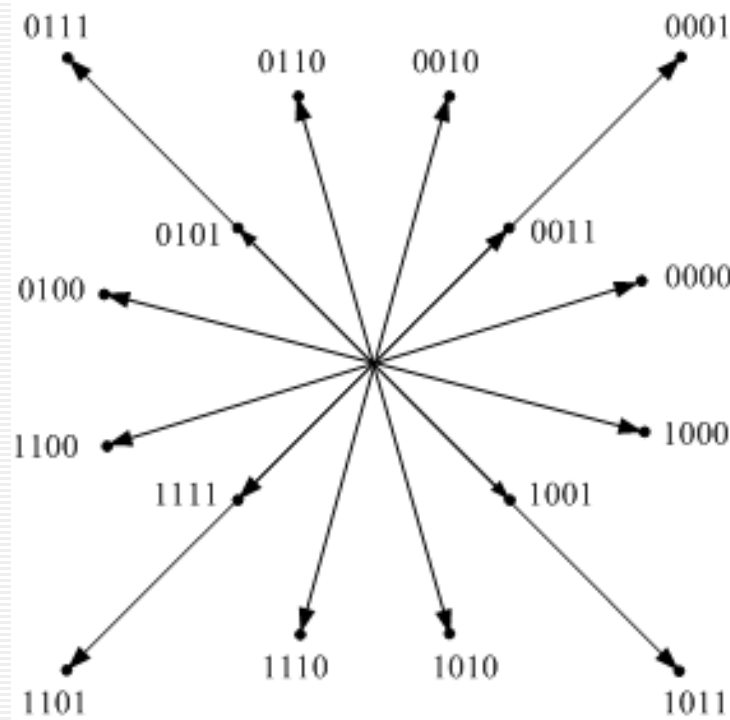
$$\begin{aligned}
 \int_0^T s(t) \cos(2\pi f_c t) dt &= \int_0^T \sqrt{2} \cos(2\pi f_c t + \frac{3\pi}{4}) \cos(2\pi f_c t) dt \\
 &= \sqrt{2} \int_0^T \cos(2\pi f_c t) \cos \frac{3\pi}{4} \cos(2\pi f_c t) dt - \sqrt{2} \int_0^T \sin(2\pi f_c t) \sin \frac{3\pi}{4} \cos(2\pi f_c t) dt \\
 &= -\int_0^T \cos(2\pi f_c t) \cos(2\pi f_c t) dt - \int_0^T \sin(2\pi f_c t) \cos(2\pi f_c t) dt \\
 &= -\frac{1}{2} \int_0^T (\cos(4\pi f_c t) + \cos 0) dt - \frac{1}{2} \int_0^T (\sin(4\pi f_c t) + \sin 0) dt \\
 &= -\frac{1}{2} \int_0^T \cos(4\pi f_c t) dt - \frac{1}{2} \int_0^T 1 dt - \int_0^T \sin(4\pi f_c t) dt = -\frac{T}{2}
 \end{aligned}$$

$$\begin{aligned}
 \cos(A + B) &= \cos A \cos B - \sin A \sin B \\
 \cos(A - B) &= \cos A \cos B + \sin A \sin B \\
 \sin(A + B) &= \sin A \cos B + \cos A \sin B \\
 \sin(A - B) &= \sin A \cos B - \cos A \sin B
 \end{aligned}$$

# Digital Data → Analog Signals (5)

---

## Constellation Pattern



- PSK using 12 angles and three amplitudes  
- 9,600 bps modem (2,400 baud)

# Digital Data → Analog Signals (6)

---

- Modulation Rate

- Data rate (in bps) = modulation rate (in baud) x the number of bits per signal element

- Constellation Pattern

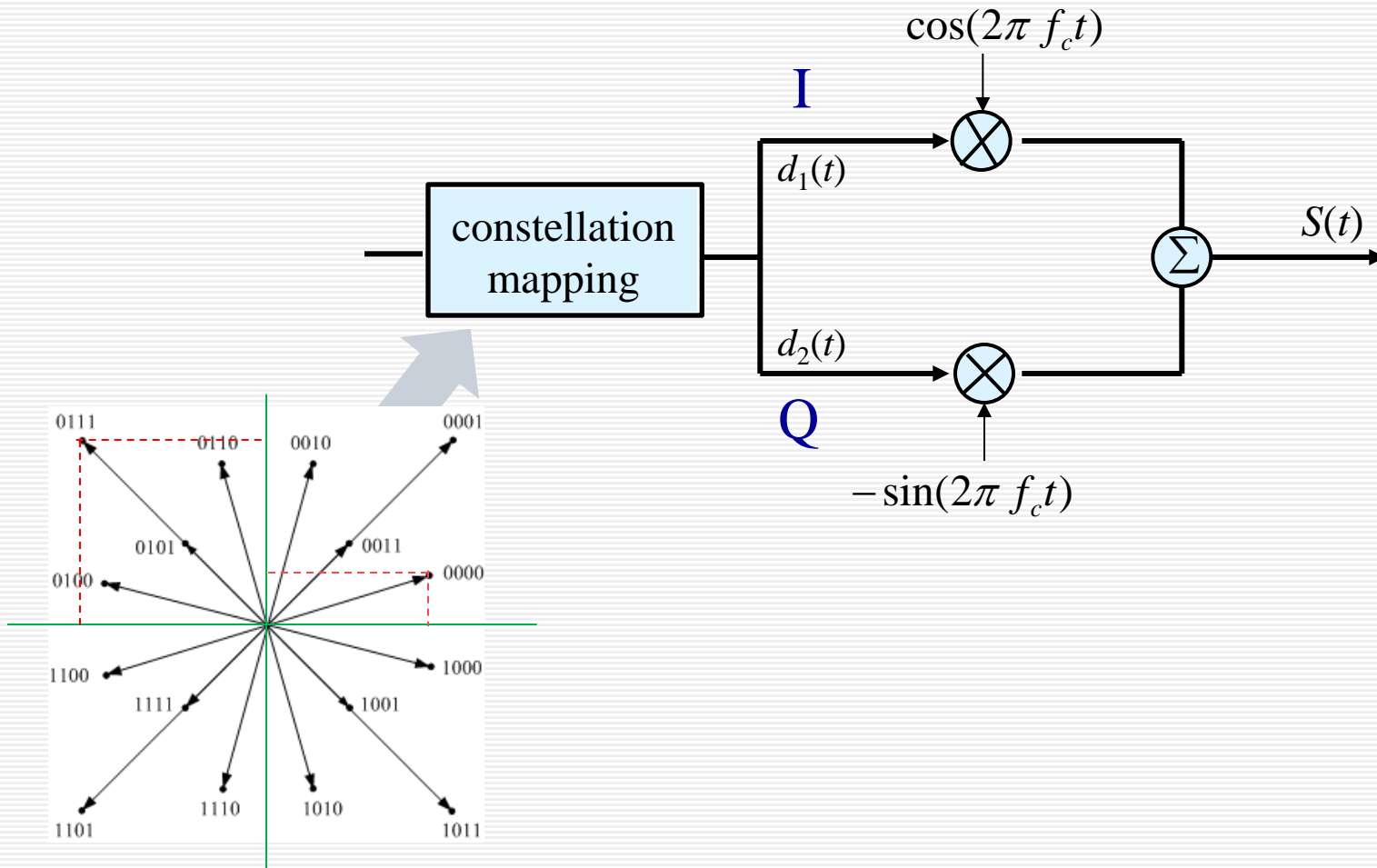
- The patterns showing legal combinations of amplitude and phase

- Example (2400 baud)

- 64 points constellation  $\Rightarrow$  14,400 bps modem
- 128 points constellation  $\Rightarrow$  16,800 bps modem

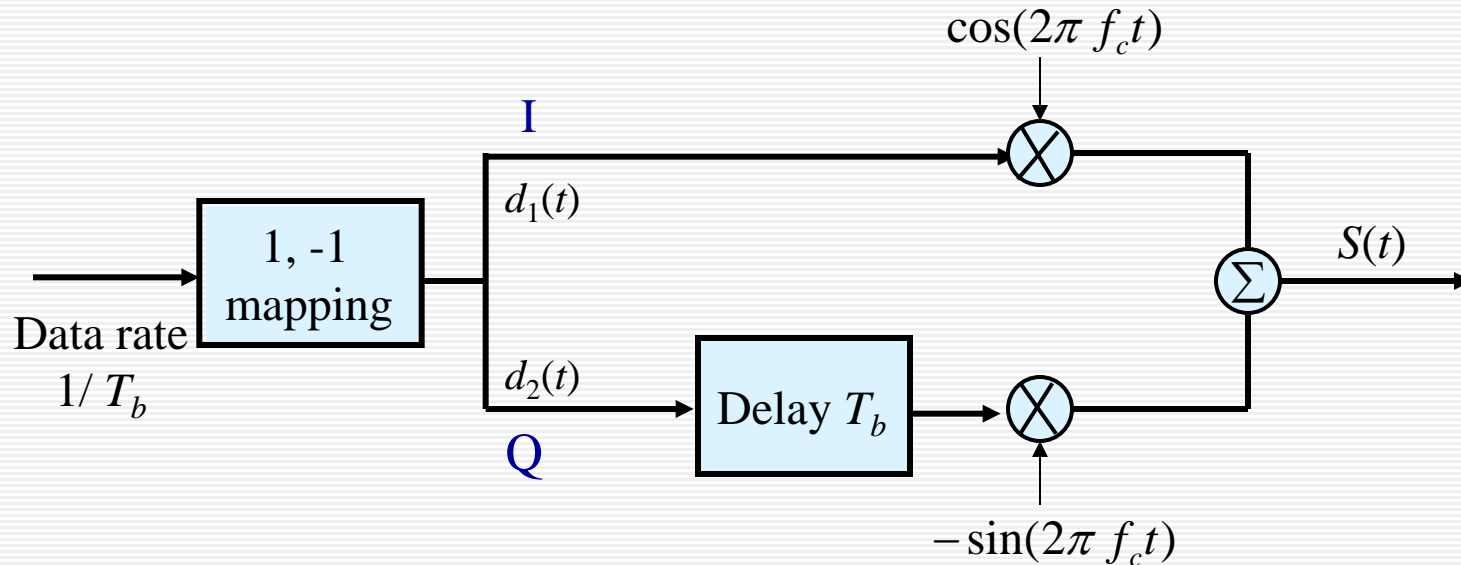
# Digital Data $\rightarrow$ Analog Signals (7)

- QAM modulator

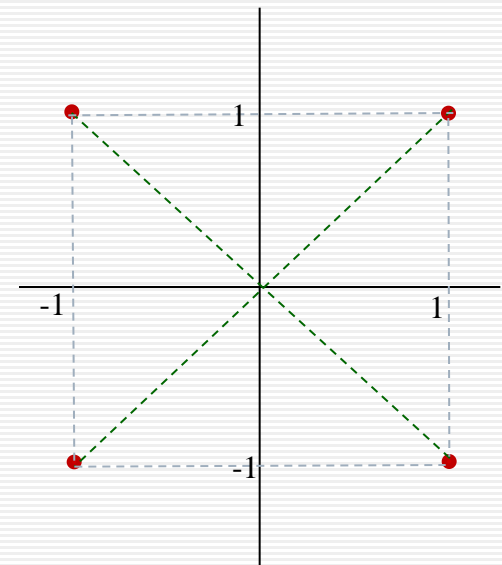
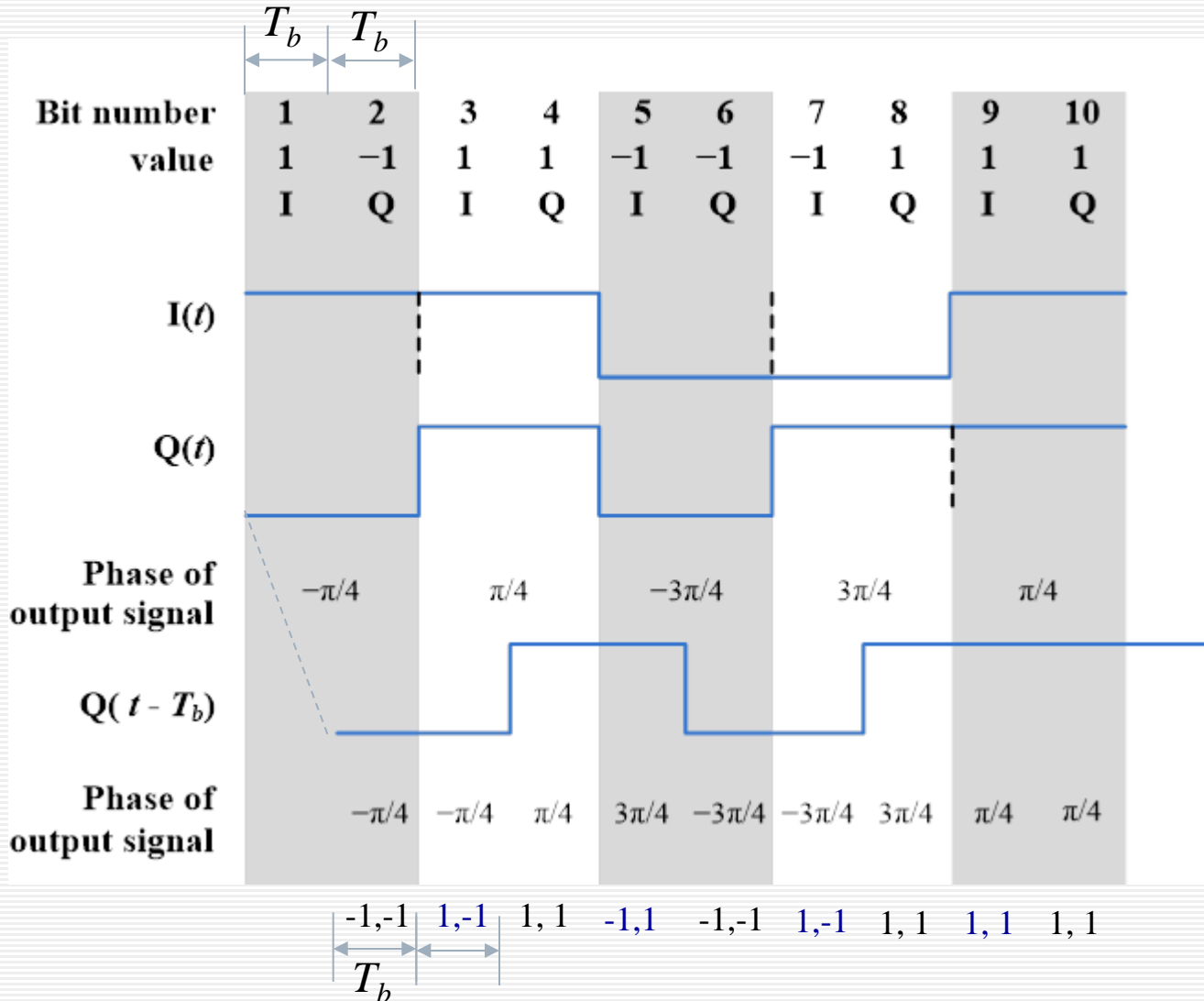


# Digital Data → Analog Signals (8)

- Offset-QPSK modulation



# Offset-QPSK



# Digital data transmission terminology

---

| Term                                     | Description   |
|--|---|
| <b>Data element</b>                      | bits (binary one or zero)   |
| <b>Data rate</b>                         | bits per second(bps)<br>rate at which data elements are transmitted   |
| <b>Signal element</b>                    | <b>Digital:</b> a voltage pulse of constant amplitude<br><b>Analog:</b> a pulse of constant frequency, phase, and amplitude |
| <b>Signaling rate or modulation rate</b> | signal elements per second (baud)<br>rate at which signal elements are transmitted  |



# Analog Data → Digital Data (1)

---

## Digitization

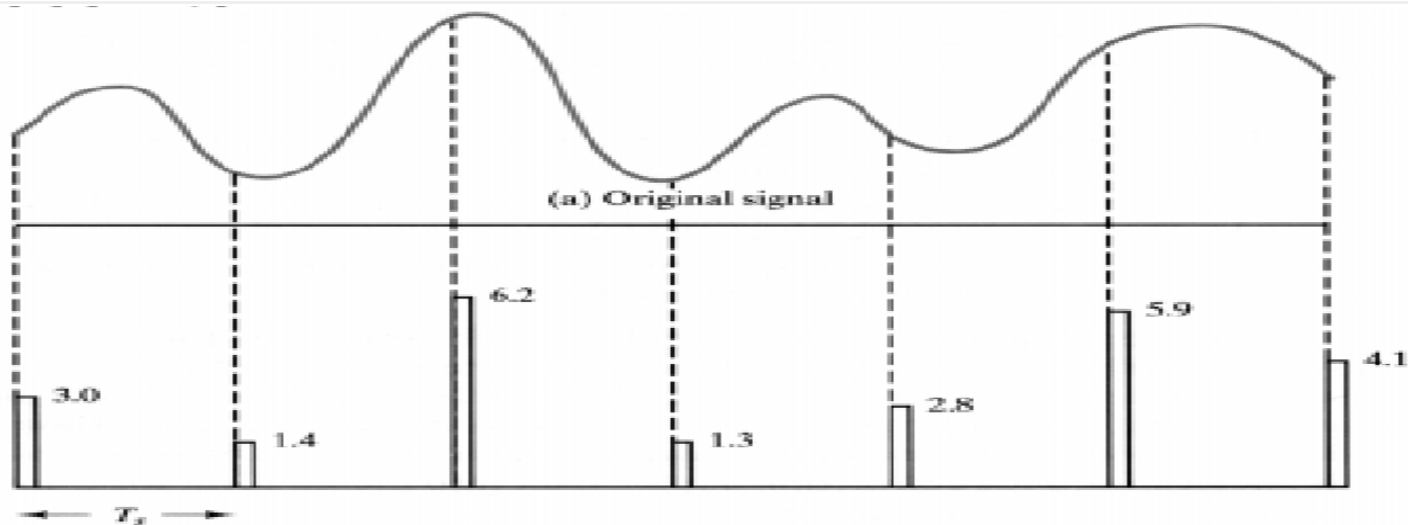
- Conversion of analog data into digital data
  - Digital data can then be transmitted using NRZ-L
  - Digital data can then be transmitted using code other than NRZ-L
  - Digital data can be converted to analog signal
- Analog to digital conversion done using a codec
- Pulse code modulation
- Delta modulation

# Analog Data → Digital Data (2)

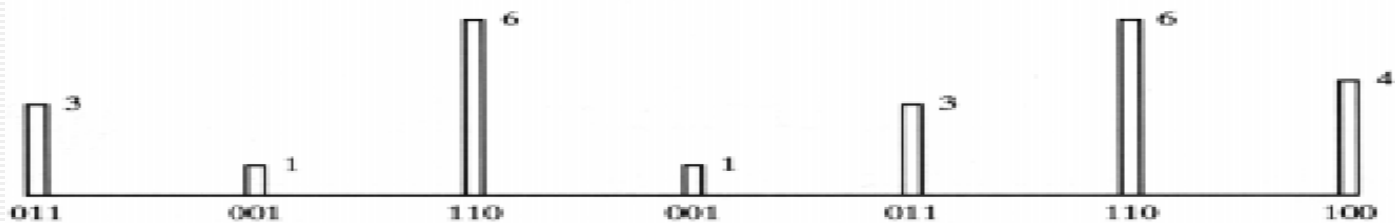
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- **PCM(Pulse Code Modulation)**
  - Based on the Nyquist's Sampling Theorem
- **Nyquist's sampling Theorem**
  - If a signal is sampled at regular intervals at a rate higher than twice the highest signal frequency, the samples contain all the information of the original signal
  - example (voice)
    - sampling rate = 8000 sampling/sec (1 sampling/125 $\mu$ s)
    - 1 sample = 8 bits
    - 64kbps

# Analog Data → Digital Data (3)



(b) PAM pulses



(c) PCM pulses

011001110001011110100

(d) PCM output

# Analog Data → Digital Data (4)

---

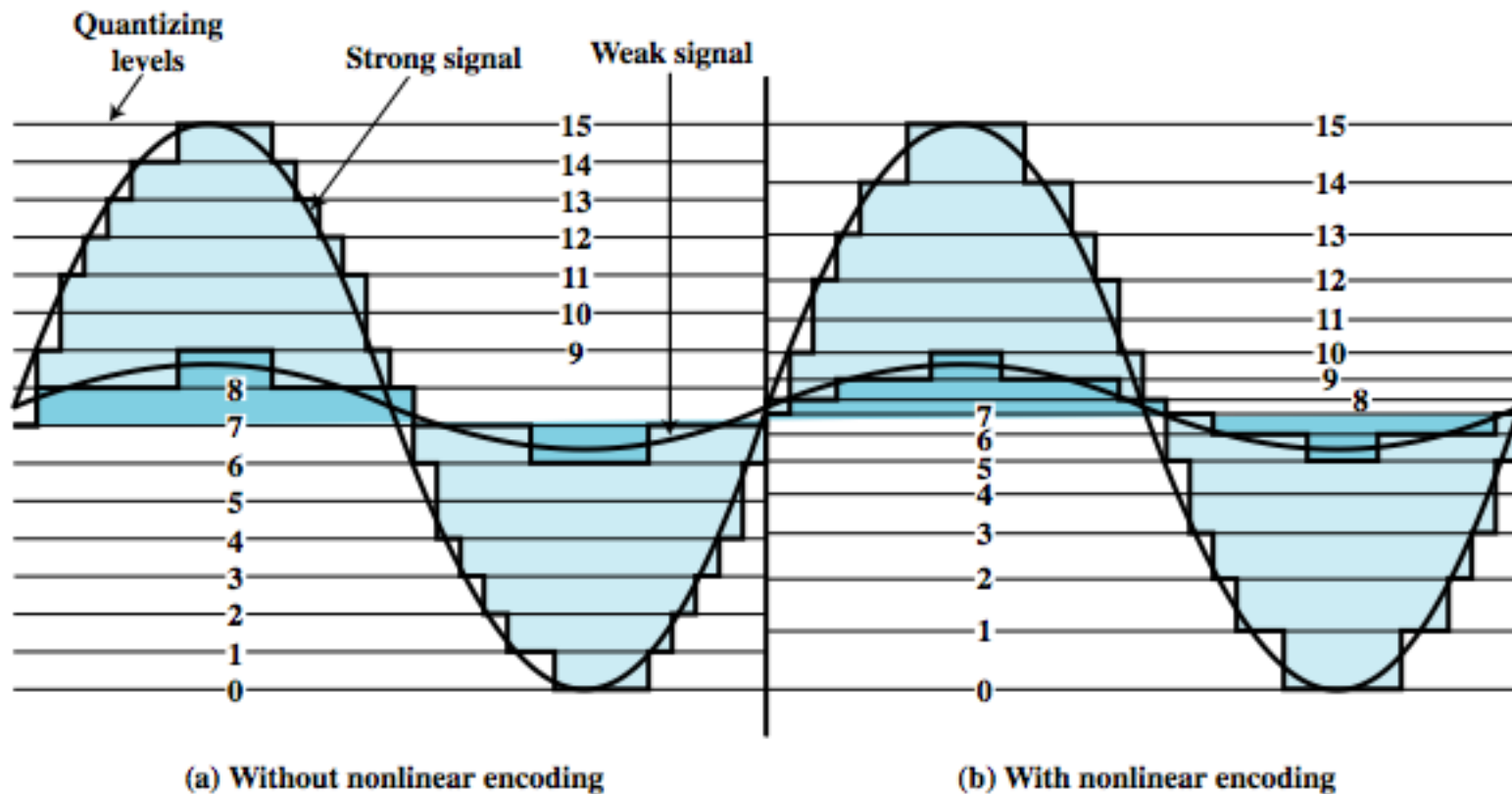
- Analog-to-Digital Conversion



- PCM

- To reduce the information loss due to quantization
  - a large  $n$ , where  $n$  is the number of bits per sample
  - nonlinear coding or companding+linear coding

# Analog Data → Digital Data (5)



**Effect of nonlinear coding**

# Companing (compressing-expanding)

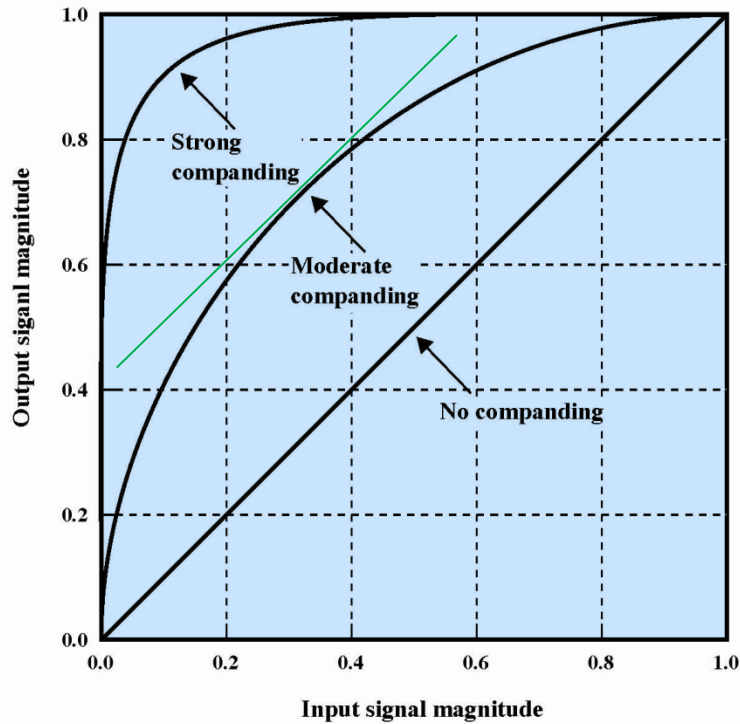


Figure 5.19 Typical Companding Functions

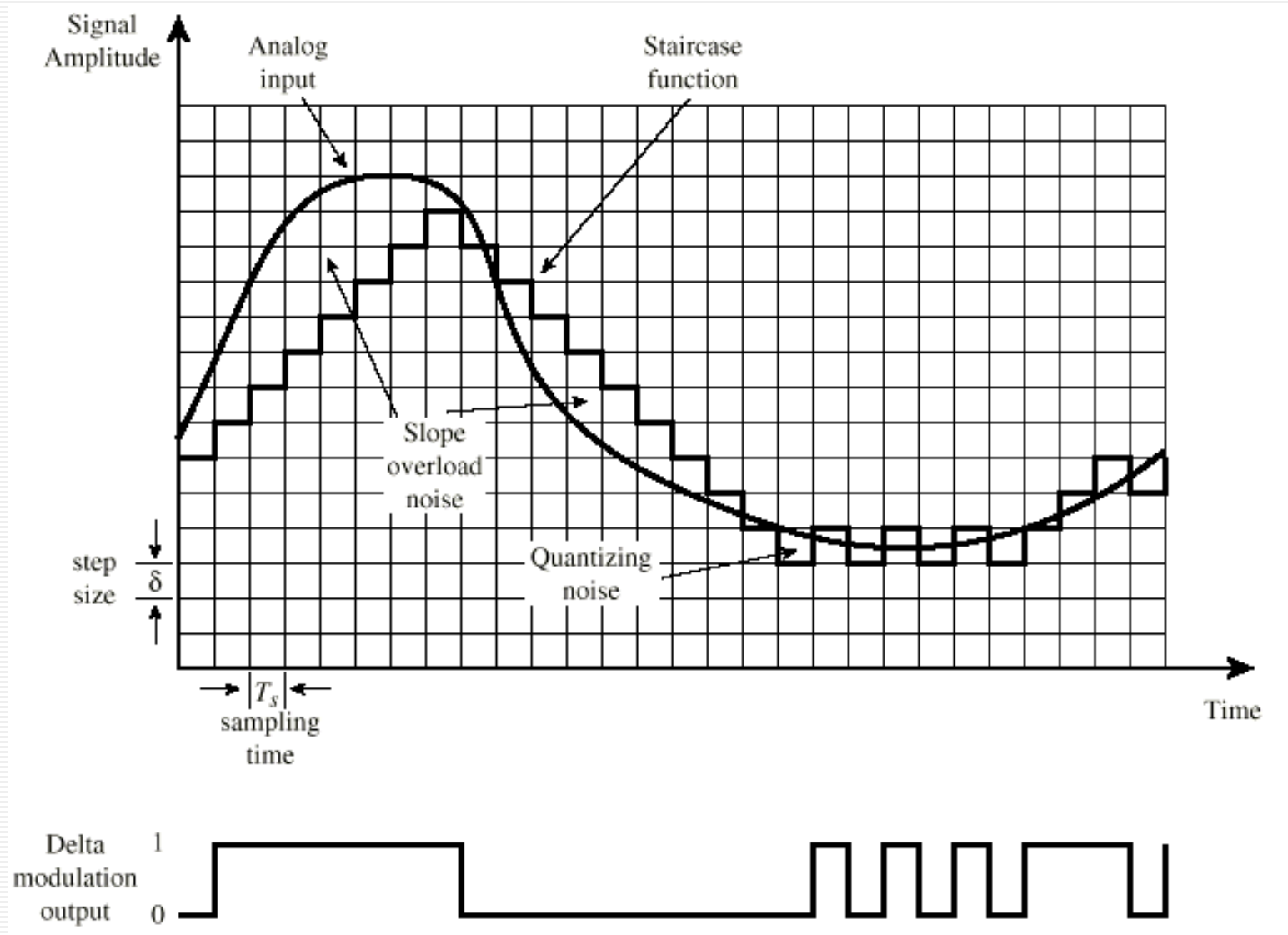
# Analog Data → Digital Data (6)

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## DM(Delta Modulation)

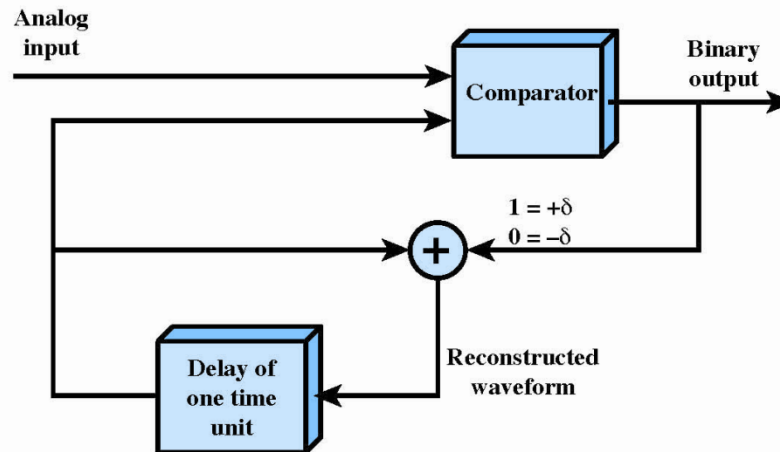
- This uses  $n=1$ 
  - Higher sampling rate than PCM(Nyquist's rate) is needed, but each sample uses only 1 bit instead of  $n$ .
- Analog input is approximated by a staircase function
  - Move up or down by  $\delta$  at each sampling time
- binary digital signal is produced;
  - “0” stands for change of  $-\delta$
  - “1” for change of  $+\delta$ .
- Implementation much simpler than PCM.

# Delta Modulation Example

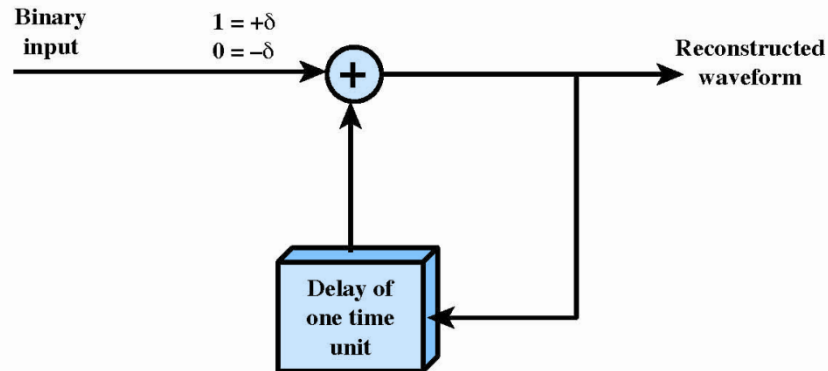




# Delta Modulation



(a) Transmission



(b) Reception